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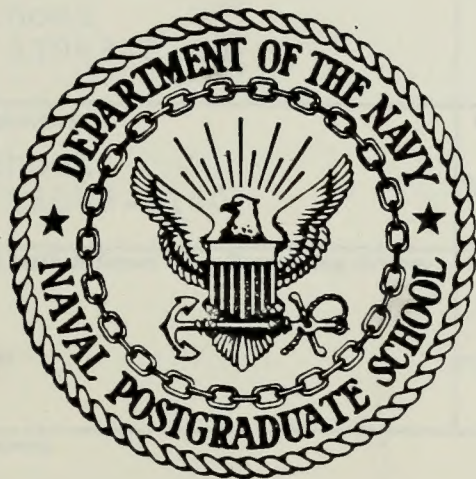




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# NAVAL POSTGRADUATE SCHOOL

Monterey, California



## THESIS

COMPUTER SIMULATION OF  
DIGITAL SIGNAL MODULATION TECHNIQUES IN  
SATELLITE COMMUNICATIONS

by

Craig Dean Carlson

September 1985

Thesis Advisor:

James L. Wayman

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of the FFT's was conducted to determine if there is any relationship between the components of the FFT of the different signals. The statistic used to investigate this possible relationship was the F-distribution. The computer simulation was written and conducted in the FORTRAN programming language. A copy of the program, results of the simulation and the statistical analysis conducted are included in the appendices.



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Computer Simulation of  
Digital Signal Modulation Techniques in  
Satellite Communications

by

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Submitted in partial fulfillment of the  
requirements for the degree of

MASTER OF SCIENCE IN SYSTEMS TECHNOLOGY  
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September 1985

## ABSTRACT

This thesis is a tutorial on digital signal modulation techniques used in satellite communications and includes computer simulations of those digital signal modulation techniques introduced. The purpose of the thesis is to introduce digital signal modulation techniques and through the use of computer simulation, generate statistics which represent the characteristics of the FFT for the respective signal type. Further, an analysis of the statistics of the FFT's was conducted to determine if there is any relationship between the components of the FFT of the different signals. The statistic used to investigate this possible relationship was the F-distribution. The computer simulation was written and conducted in the FORTRAN programming language. A copy of the program, results of the simulations and the statistical analysis conducted are included in the appendices.



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## I. INTRODUCTION

### A. BACKGROUND

Since the introduction of the sampling theorem and the matched filter, digital communications techniques have developed into a highly proficient discipline. The marriage of this discipline with the rapidly expanding space program has resulted in communication satellites employing a multitude of digital signal modulation techniques. A modulation technique is a method of transmitting the information contained in a message by varying or modulating the characteristics of a carrier waveform. These methods offer a range of advantages and disadvantages depending on the specific characteristics of the modulation technique employed. The applications of the various signal modulation techniques likewise vary. In order to understand this exciting new field it is necessary to look at some of the aspects of satellite communications systems in general. Then an investigation will be made into the general attributes of digital communications and their relationship to the satellite system.

The ability to understand these digital signal modulation techniques is the first step in being able to intercept, identify and demodulate unknown digital signals. These digital signals, transmitted from unknown sources, are believed to display frequency characteristics peculiar to the modulation technique employed in the encoding process. The digital computer offers unique opportunities in simulating these modulation techniques, in signal processing and in decoding of an intercepted signal.

## B. SPECIFIC GOALS

It is the specific goal of this thesis to investigate the open literature on the topic of digital signal modulation techniques in satellite communications. This includes a basic understanding of the communications satellite system as well as the specific techniques employed in signal modulation. Additionally, once a basic understanding of these digital signal modulation techniques is achieved, computer code in the FORTRAN programming language will be developed which simulates these modulated signals. The statistics of the time-varying Fast Fourier Transforms (FFT) of these simulated signals will be investigated. The purpose of this analysis will be to lead to follow-on research in the area of signal analysis of intercepted digital signals whereby they can be classified by their FFT as employing a specific digital signal modulation technique.

## C. SCOPE OF THE PROJECT

Although an indepth analysis of all the digital signal modulation techniques which will be introduced involves considerable higher level mathematics, it is not within the scope of this project to delve heavily into the mathematics. It would be advantageous, however, for the reader to have had integral and differential calculus and an introduction to statistics. Also a course in electrical engineering may prove helpful but is not essential since in actuality it is the electrical engineering aspects of satellite communications that this tutorial is attempting to present. The computer programming will be accomplished utilizing the techniques of software engineering and top down modular design. Existing blocks of code will be used as modules whenever appropriate and available. Again it is the



ultimate purpose of this project to examine a variety of digital signal modulation techniques and develop computer code that simulates common signals used in satellite communications that have been produced by one of the many digital signal modulation techniques to be investigated.

## II. UNDERSTANDING SATELLITE COMMUNICATIONS

### A. HISTORY AND APPLICATIONS

Forty years ago, in 1945, Arthur Clarke first envisioned the use of space stations placed in geosynchronous orbit for communicating to different points on the earth [Ref. 1]. Nine years later in 1954, J.R. Pierce of Bell Laboratories performed a system analysis on such a communications system [Ref. 2]. In 1957, the launch of Sputnik demonstrated the feasibility of using a satellite for just such an application. However, by 1961 the only satellite communications technologies which had been demonstrated were the Courier 1B satellite, a short-life active retransmission teletype communications satellite in a 1000 km orbit, and the Echo I balloon in a 1600 km orbit which demonstrated passive reflection of powerful microwave signals from one earth station to another. Active microwave communications demonstrated in Projects Telstar and Relay were years away [Ref. 3]. The first geostationary orbit was achieved by Project Syncom in 1963 [Ref. 4].

In 1964, communication organizations from several countries joined together to form the international organization of INTELSAT (International Telecommunications Satellite Organization). INTELSAT's purpose was to develop a satellite network which would provide truly global communications capabilities. This resulted in the launch in 1965 of the world's first commercial communications satellite, INTELSAT I, also known as "Early Bird". With "Early Bird", telecommunications utilizing satellite relay were established between the United States and Europe. [Ref. 3]

The reliability of these satellite systems has improved dramatically since INTELSAT I in 1965. Reliability of individual links in the system approach 99.99 percent or higher. Total system reliability exceeding 99.9 percent is common [Ref. 5], making satellite communications more reliable than many other modes of communications. In this sense, reliability is a measure of the probability that no failure will occur in a respective channel or in the system during the design life of the satellite.

In order for the operability, capability and reliability of these systems to have developed at this pace, it was necessary for the technologies associated with them to develop as rapidly or even more rapidly than the systems themselves. Engineering sciences and specifically the fields of aerospace and electrical engineering historically have required between 7 and 10 years to take a concept from operational requirement to full scale operation. This was not the case with the concept of communications satellites which has seen four generations of INTELSAT satellites within a decade's time. [Ref. 3]

The "Early Bird" satellite weighed approximately 38 kilograms and possessed limited power and bandwidth capacity enabling it to carry only 240 two-way telephone conversations. Today's communications satellites represent order-of-magnitude improvements in many important operating parameters such as power and bandwidth. Additionally, increased effective radiated power from the techniques of stabilized earth pointing antennas have greatly increased the capacity of later generation communications satellites. INTELSAT V, the current generation of communications satellites, is a three-axis stabilized platform using not only the 6/4 GHz frequency band (6 GHz receive/4 GHz transmit) as in earlier generation satellites but also a 500 MHz bandwidth available in the 14/11 GHz band. The



separation in the receive and transmit frequencies is necessary to prevent interference during simultaneous operation of the receiver and transmitter. Another factor increasing the capacity of present generation satellites is the increase in primary power available in the satellite. [Ref. 3]

The technologies which have contributed to the evolution of communications satellites come from two primary sources, namely technology from the space program of the 1960's supported by NASA and DoD and communications technology due largely to commercial and private sector contributions. [Ref. 3]

Electronic devices and components have contributed significantly to the rapid development of satellite systems. These electronic devices range from something which is now considered basic, i.e., the transistor, to devices such as Traveling Wave Tube amplifiers, lightweight antennas and antenna feed systems. There have been major improvements in satellite power sources including more efficient solar cells and high storage capacity/lightweight batteries. Additionally, two significant developments in electrical engineering have made modern day digital signal processing a reality. They are the sampling theorem and the matched filter. [Ref. 3]

Communications satellites of the future are likely to utilize onboard signal processing. Signal processing functions such as signal reshaping, switching and/or multiplexing and compression could soon take place on the satellite due to the reduction in size and weight of the necessary hardware. Operation at over 100 megabits per second are envisioned. Onboard signal processing will greatly reduce the expense and complexity of earth stations thereby making services of a satellite available to a wider range of small and geographically disperse users. [Ref. 3]

Satellite communications hold great promise to provide service to a multitude of users over a wide area. Applications lie not solely in retransmission but also in data collection from that same large geographic area. Military users have definite applications in these areas when warning, intelligence and surveillance systems provide digital inputs to a central source. Although most of the present applications for satellite communications are still provided by the government, commercial applications have seen tremendous increases in the last several years. [Ref. 3]

## B. ORBITS AND LIMITATIONS

Satellite orbits are generally categorized as either equatorial (0 degrees inclination), polar (90 degrees inclination) or inclined at some angle other than 0 or 90 degrees relative to the spin axis of the earth as illustrated in Figure 2.1 [Ref. 6]. Each satellite orbit has a characteristic velocity which is dependent on the height of the orbit and the orbit's eccentricity. Eccentricity is a measure of the degree to which the orbit approximates a circle. A circle has eccentricity equal to zero since the focus is located at the center. See equation 2.1 and Figure 2.2.

A communications satellite in elliptical orbit about the earth obeys Kepler's second law of planetary motion. That is, a satellite's constant angular momentum about the earth means that its areal velocity also remains constant. See Figure 2.3. Of particular interest is the satellite velocity at apogee ( $V_a$ ) and perigee ( $V_p$ ) given by equations 2.2 and 2.3.

Note that for a circle,  $e = 0$  and  $R_a = R_p$ . Therefore a satellite in circular orbit about the earth has an orbital velocity as given in equation 2.4.

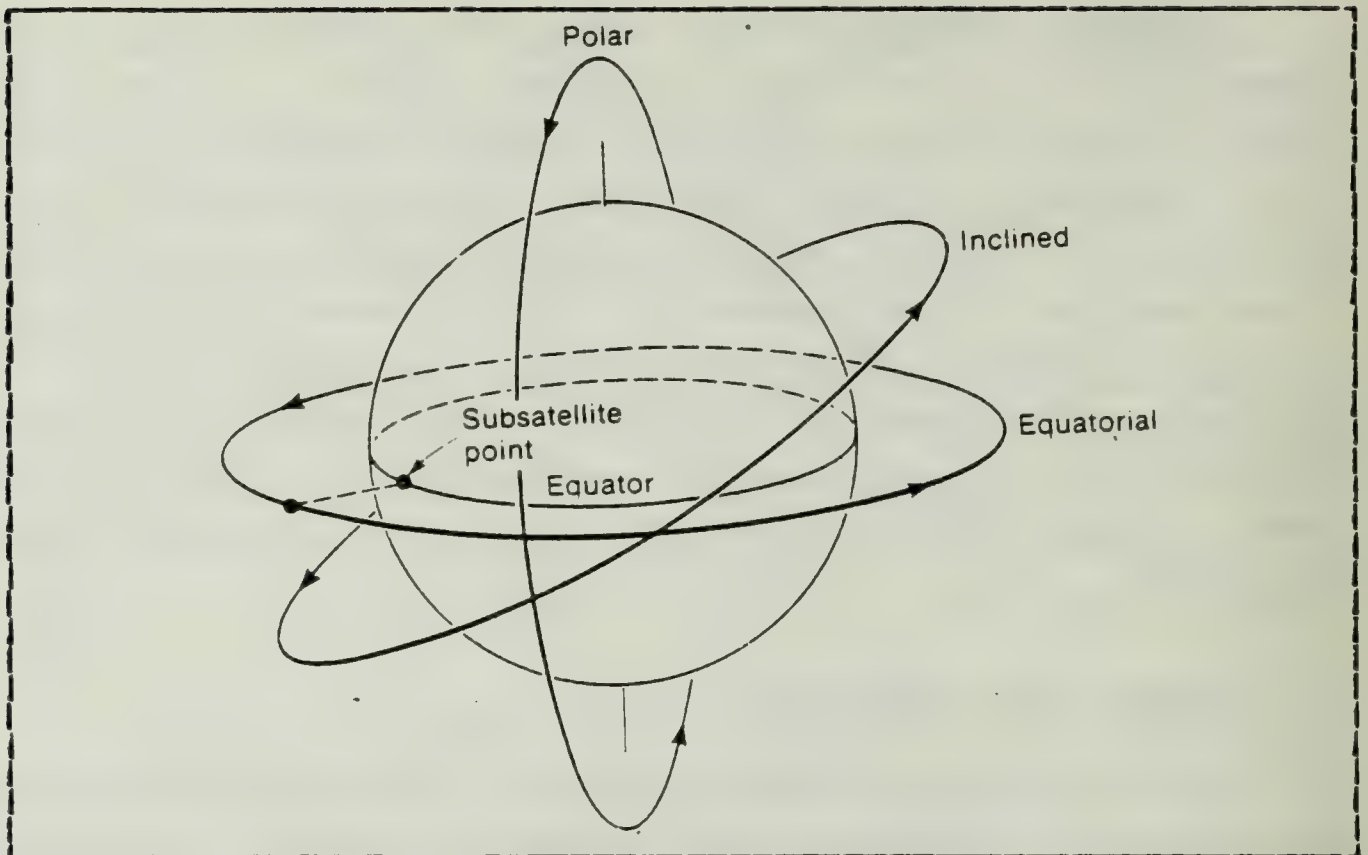


Figure 2.1 Satellite Orbits

Most communications satellites are placed in a circular equatorial orbit where the desire is to stabilize them over a fixed point on the surface of the earth called the subsatellite point. This type of orbit is called geostationary or stationary. Any satellite which has an orbital period equal to the period of rotation of the earth is called synchronous or geosynchronous. These terms are used almost interchangeably in most literature. As mentioned for a geosynchronous communications satellite, the orbital period of the satellite,  $T$ , must be equal to the period of rotation of the earth. Kepler's third law of planetary motion can be rewritten to yield equation 2.5.

The period of revolution of the earth for a sidereal day is 23 hr 56 min 4 sec. For that given period there is only one satellite altitude as expressed by Kepler's third law.



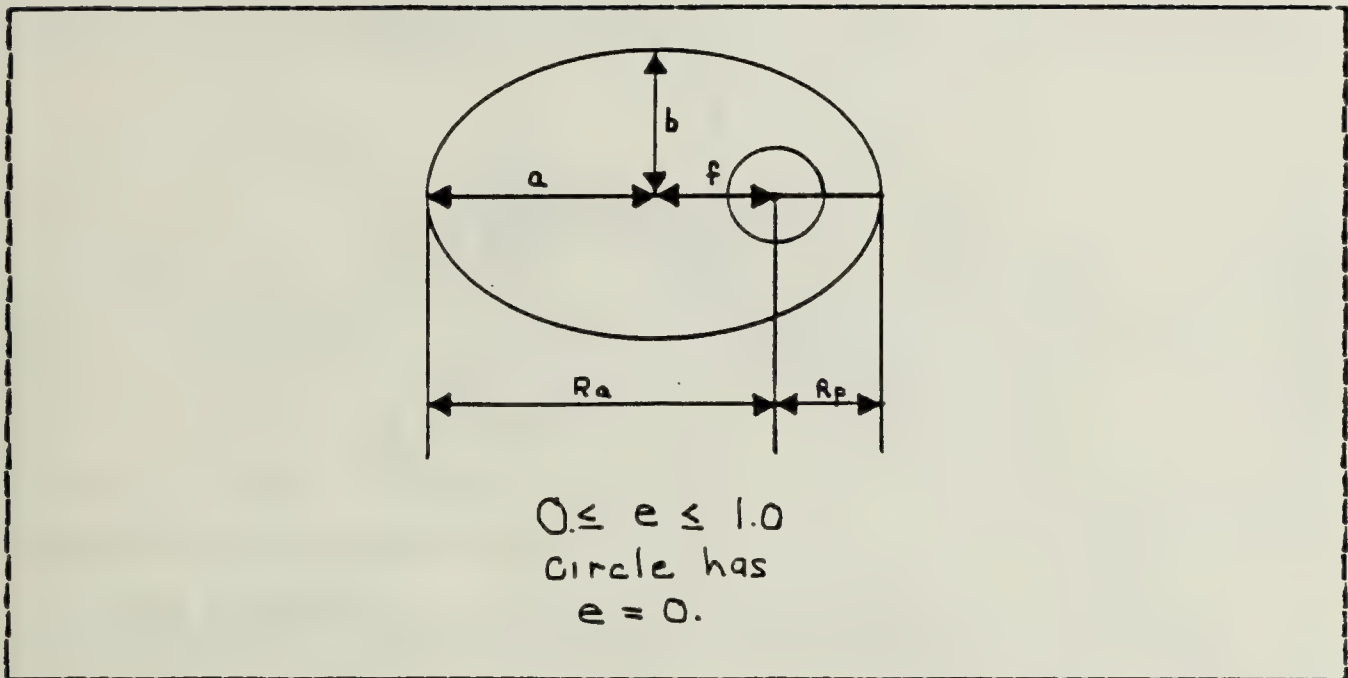


Figure 2.2 Eccentricity

eccentricity (e) =  $c/a$  (eqn 2.1)

c = distance from focus

a = semi-major axis

b = semi-minor axis

Ra = radius of apogee

Rp = radius of perigee

By rearranging terms that altitude is given in equation 2.6. Since the orbit is circular,  $a = R_a = R_p$  and the height of the orbit above the surface of the earth is  $h = a - R_e$ ; or 35,804 km.

One disadvantage of a geosynchronous communications satellite is the lack of global coverage. These satellites provide excellent coverage of the subsatellite point and

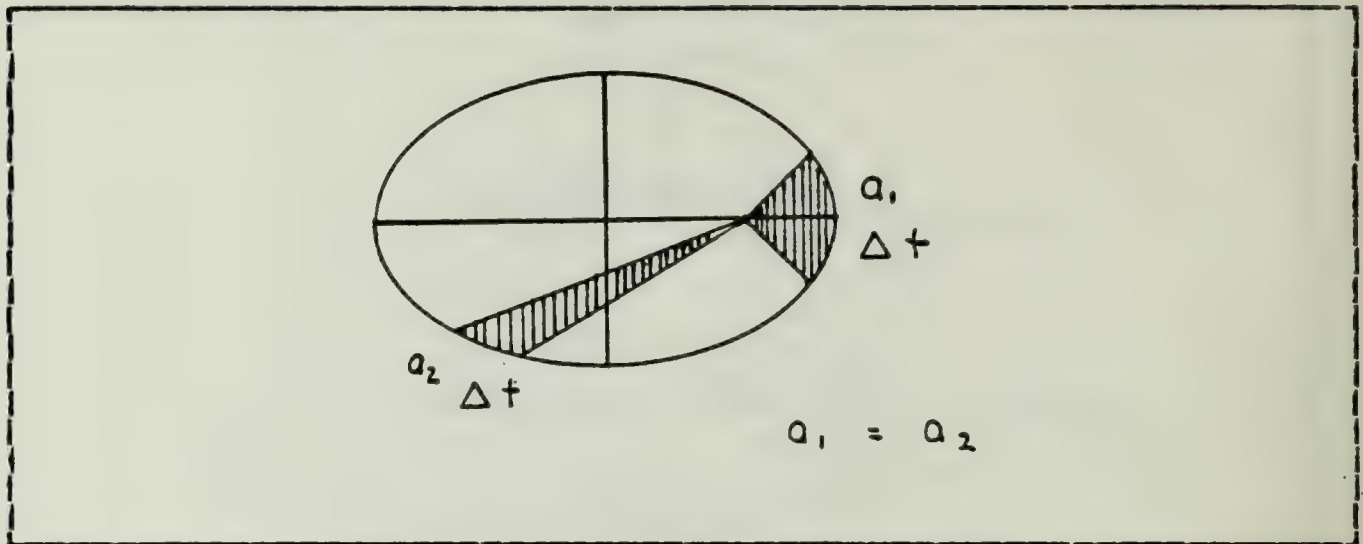


Figure 2.3 Kepler's Second Law of Planetary Motion

$$V_a = (k/R_a(1-e))^{.5} \quad (\text{eqn 2.2})$$

$$V_p = (k/R_p(1+e))^{.5} \quad (\text{eqn 2.3})$$

$$k = G M_e = 3.98866 \times 10^{-11} \text{ m}^3/\text{s}^2$$

G = universal gravitational constant

$$G = 6.67 \times 10^{-11} \text{ N m}^2/\text{kg}^2$$

N = newtons = m kg/s<sup>2</sup>

$$M_e = \text{mass of the earth} = 5.98 \times 10^{24} \text{ kg}$$

$$V_c = (k/R_c)^{.5} \quad (\text{eqn 2.4})$$

$$R_c = R_e + h$$

$$R_e = \text{radius of the earth} = 6.37 \times 10^6 \text{ m}$$

h = satellite orbital altitude

$$T = 2 \pi a^{1.5} / k^{.5} \quad (\text{eqn 2.5})$$

T = period of rotation

a = semi-major axis

$$a = \sqrt[3]{(T / 2 \pi) (k)}$$

(eqn 2.6)

$$a = 42,173 \text{ km}$$

laterally to a latitude of about  $\pm 80^\circ$  [Ref. 6]. This is generally no problem for commercial applications since the polar regions do not require a significant degree of access. The military, on the other hand, does have an interest in communications in the polar region and therefore has several communications satellites that have orbits inclined at various angles. This type of orbit generally has disadvantages of lack of continuous coverage and a much more complicated system of ground tracking and receiving stations.

### C. FREQUENCY BAND CONSIDERATIONS

Although there appears to be an infinite number of frequencies available for communications, restrictions do exist as to those which are practicable. Limitations on available frequency bands for satellite communications are due to the need to select segments of the electromagnetic spectrum which reduce noise and interference and are most favorable in terms of power efficiency and propagation distortions. Trade offs must be made to arrive at the optimum frequency for a particular application since single frequencies seldom offer the best performance for all variables. The problems arise when consideration is given to the number of users requiring the same frequency bands including terrestrial communications networks. Since the problem of interference is global, a worldwide organization has been established to assign frequency bands for various applications. This organization is called WARC, World Administrative Radio Conference [Ref. 6]. Illustrations of



the portions of the electromagnetic spectrum in question and current allocations of satellite frequency bands are shown in Figures 2.4 and 2.5 [Ref. 6].

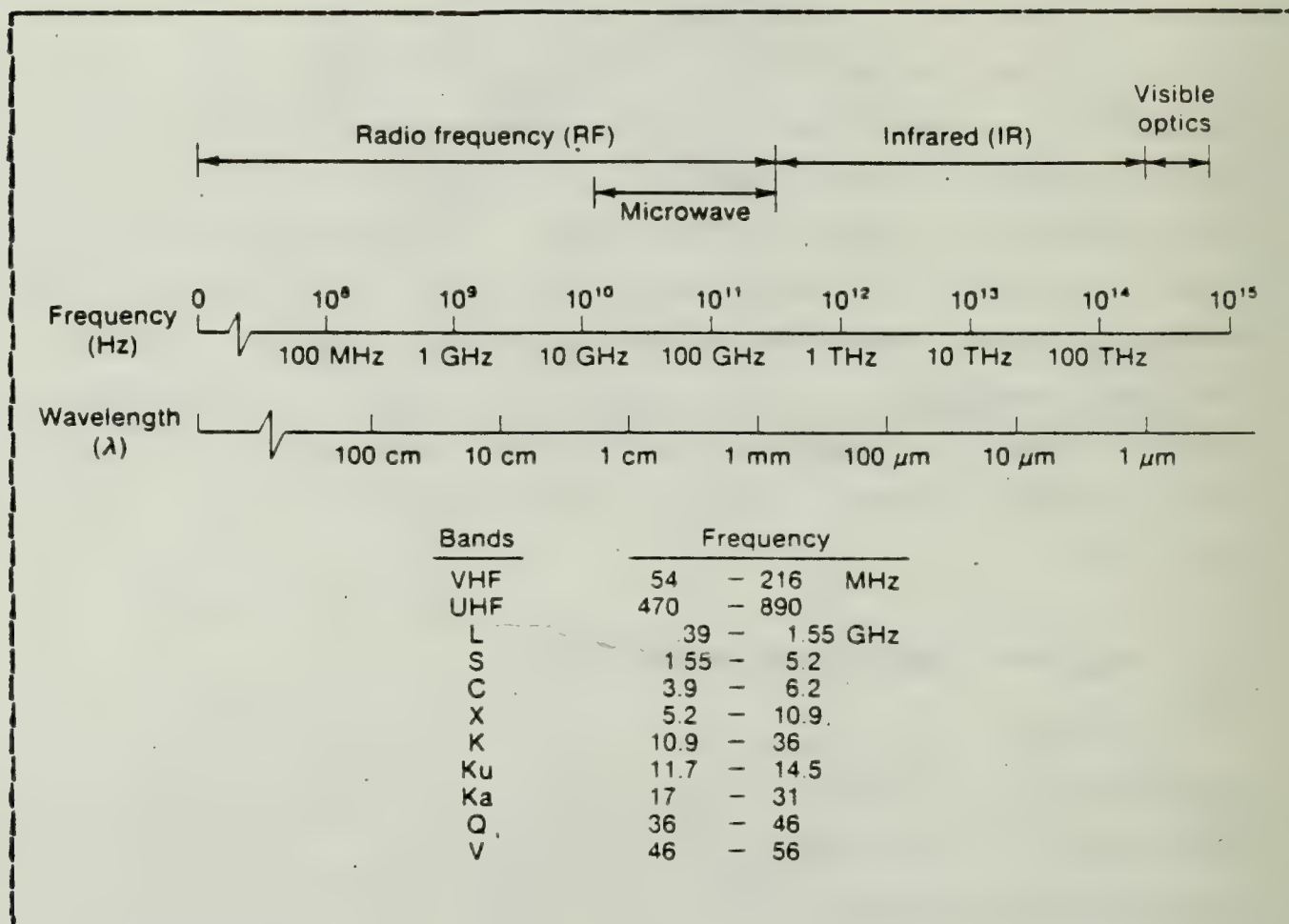


Figure 2.4 Electromagnetic Spectrum

In general, as long as the hardware and technology will support it, higher frequencies are more desirable and more in demand because they offer higher theoretical capacity. This is due to the fact that only a percentage of the carrier frequency is capable of actually transmitting a signal of a given bandwidth. Higher frequencies would also experience less interference with existing land and satellite systems. A further discussion of the bandwidths available as a function of frequency will be included at a

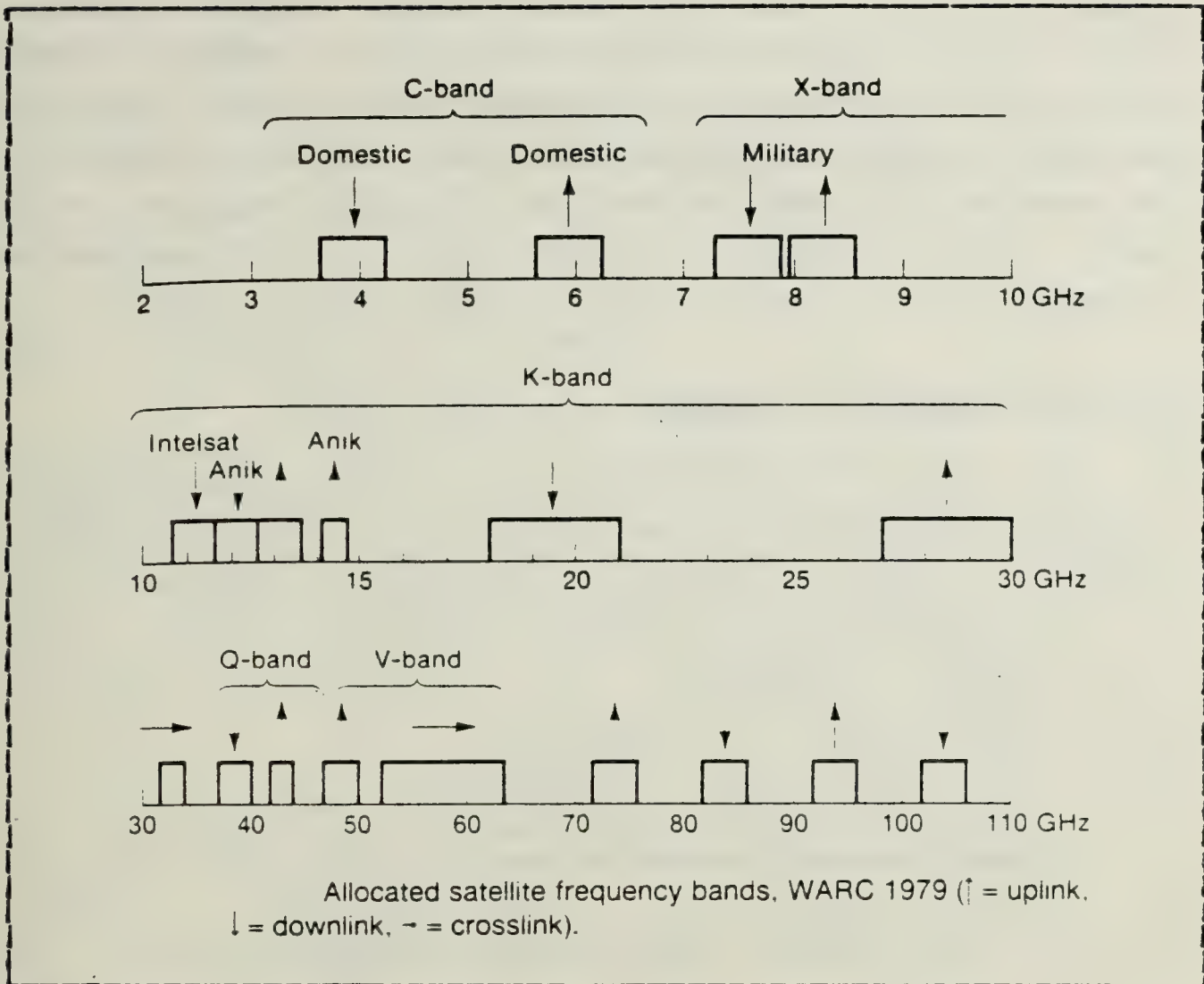


Figure 2.5 Satellite Frequencies

later point under the topic of actual digital signal techniques.

As an additional note, in order to assist in the ability to handle more signals in an obviously limited electromagnetic spectrum, several multiple access schemes have been developed. The most common techniques involve frequency-division multiple access (FDMA) where the allocated satellite frequency band is divided among the users into specific uplink and downlink frequencies. Next is time-division multiple access (TDMA) where the satellite frequency band is shared by all users by carefully dividing

the time any one user has access. Finally there is code-division multiple access (CDMA) which involves modulating a specific address code which has been superimposed on the signal directly onto the carrier. In this manner all users share the satellite frequency band and only those receiving stations that can demodulate the address code can receive the specific signal. [Ref. 6]

#### D. WHY DIGITAL MODULATION

In general, digital signal modulation techniques are superior to analog signal modulation techniques used in satellite communications for the following reasons:

1. Compatibility with digital computers
2. Economic advantages
3. High degree of flexibility
4. Less susceptible to interference
5. Quality of signal independent of transmission distance and network makeup

##### 1. Compatibility with Digital Computers

Clearly one of the most distinctive advantages of digital techniques over analog techniques involves computer applications. Properly formatted digital signals can be used to represent any analog signal (more on this under the discussion of the sampling theorem). Once in the digital format these signals can be easily manipulated within the digital computer. Arithmetic operations can be applied as well as logical operations. The signal can be stored without alteration or delayed and therefore can be used to "simulate" real physical situations.

The hardware associated with digital circuits is free from drift or aging and does not require calibration. Additionally digital circuitry is compatible with present day integrated circuit technology allowing a standardized



building block construction approach. The operating characteristics of these systems employing a digital computer can be changed by altering the software rather than the hardware as is the case for analog systems. Finally, the use of digital computer technology allows time multiplexing.

## 2. Flexibility and Economy

The flexibility and economy of digital satellite communications comes from the fact that more and more processing can be done onboard. This allows the uplink and downlink to be completely separated. This regenerative nature makes it possible for low error rates and high reliability through the use of digital techniques not available to analog systems. Because digital signal processing or multiplexing is less costly than for analog signals, simpler and cheaper interfaces between earth stations and terrestrial communications networks are possible. Additionally, there are reduced production costs and increased capacity associated with digital circuits. [Ref. 7]

## 3. Quality and Interference

The capability of digital systems to regenerate the signal and allow for multiple switching and signal processing without degradation in signal quality makes the digital signal basically independent of transmission distance. Multiple hops from satellite to earth station or satellite to satellite are possible without accumulation of the noise characteristic of analog systems. Additionally, digital systems are capable of operating at a signal to noise ratio of 20 dB to 30 dB as compared to analog systems requiring a much more powerful signal. [Ref. 8]

### III. INTRODUCTION TO THE FOURIER TRANSFORM, SAMPLING THEOREM,

#### AUTOCORRELATION FUNCTION AND THE MATCHED FILTER

Essential to the understanding of digital signal modulation techniques are a few basic tools of the electrical engineer and the mathematician. These tools will be developed and elaborated on to the extent necessary to understand their applicability to the subject of digital communications. The description is not meant to be a detailed investigation of the respective topics. Where relevant, the application of the concept being described will be mentioned.

#### A. FOURIER TRANSFORM

In mathematical terms, voltages can be expressed as functions of time or of frequency. It is more common to see voltages represented as a function of time as in equation 3.1.

$$v(t) = A \cos(\omega t) \quad (\text{eqn 3.1})$$

A = amplitude

$\omega$  = angular frequency =  $2 \pi f$

f = frequency =  $1/T$

T = period

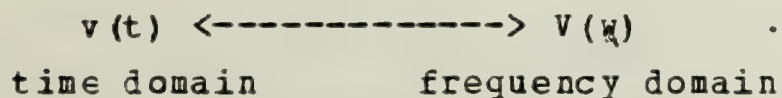
It is important to note that both the time and frequency functions are representations of voltage and as such may be used interchangeably. The Fourier representation  $[v(t)] = V(f)$  is a voltage descriptor in the frequency domain (a

function of frequency) while  $v(t)$  is a voltage descriptor in the time domain (a function of time). They are different but equivalent and either voltage descriptor may be used, depending on the concept being explored, to best represent the voltage within context.

The Fourier transform is of principle interest rather than the Fourier series since the latter is applicable only to periodic voltages. The Fourier transform is applicable to voltage pulses, random voltages and other non-periodic voltages.

DEFINITION:

$$\mathcal{F}[v(t)] = V(\omega) = \int_{-\infty}^{\infty} v(t) e^{-j\omega t} dt \quad (\text{eqn 3.2})$$



Remembering Euler's formula

$$e^{-j\omega t} = \cos(\omega t) - j \sin(\omega t) \quad (\text{eqn 3.3})$$

the Fourier transform becomes

$$\mathcal{F}[v(t)] = V(\omega) = \int_{-\infty}^{\infty} v(t) [\cos(\omega t) - j \sin(\omega t)] \quad (\text{eqn 3.4})$$

$$V(\omega) = \int_{-\infty}^{\infty} v(t) \cos(\omega t) - j \int_{-\infty}^{\infty} v(t) \sin(\omega t) \quad (\text{eqn 3.5})$$

Since it will be seen that digital communications deals primarily with pulses of finite duration (expressed as period,  $T$ ), it is worthwhile to examine the Fourier transform of a pulse of amplitude  $A$  and duration  $T$ .

$$\text{let } v(t) = \begin{cases} A; & -T/2 < t < T/2 \\ 0; & \text{elsewhere} \end{cases}$$



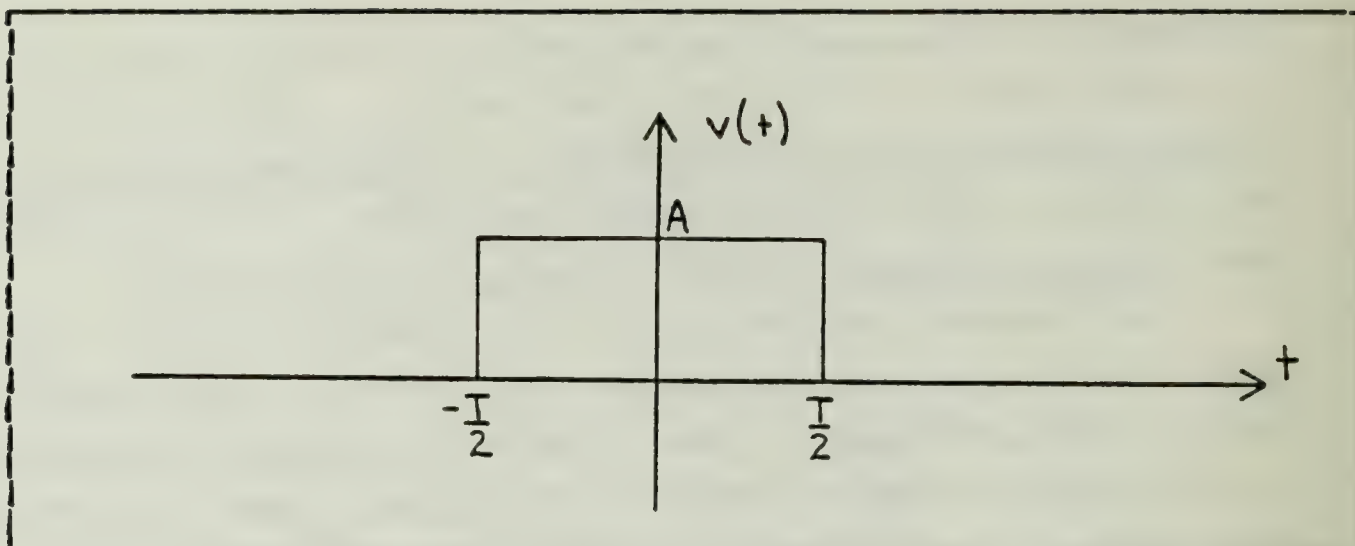


Figure 3.1 Square Pulse

Figure 3.1 is a representation of  $v(t)$  or the voltage expressed in the time domain. The position of  $v(t)$  on the  $t$ -axis was chosen for convenience of integration but could have been situated anywhere on the time line.

$$\mathcal{F}[v(t)] = V(\omega) = \int_{-\infty}^{\infty} v(t) e^{** -j\omega t} dt \quad (\text{eqn 3.6})$$

$$V(\omega) = \int_{-\infty}^{-T/2} 0 \cdot e^{** -j\omega t} dt + \int_{-T/2}^{T/2} A e^{** -j\omega t} dt + \int_{T/2}^{\infty} 0 \cdot e^{** -j\omega t} dt \quad (\text{eqn 3.7})$$

$$V(\omega) = \int_{-T/2}^{T/2} A e^{** -j\omega t} dt \quad (\text{eqn 3.8})$$

The integration above may be attacked either head on or by substituting  $\cos(\omega t) - j \sin(\omega t)$  for  $e^{** -j\omega t}$ . The direct approach is illustrated due to the relative simplicity of the integrand.

$$V(\omega) = \int_{-T/2}^{T/2} A e^{** -j\omega t} dt \quad (\text{eqn 3.9})$$

$$V(\omega) = -A/j\omega [e^{** -j\omega t}] \Big|_{-T/2}^{T/2} \quad (\text{eqn 3.10})$$

$$V(\omega) = -A/j\omega[e^{-j\omega T/2} - e^{j\omega T/2}] \quad (\text{eqn 3.11})$$

$$V(\omega) = A/j\omega[e^{j\omega T/2} - e^{-j\omega T/2}] \quad (\text{eqn 3.12})$$

By substituting  $2\pi f = \omega$ , the following result is obtained:

$$V(f) = A/j2\pi f[e^{j2\pi fT/2} - e^{-j2\pi fT/2}] \quad (\text{eqn 3.13})$$

$$V(f) = A/j2\pi f[e^{j\pi fT} - e^{-j\pi fT}]$$

Using Euler's formula, i.e.,  $\sin \theta = (e^{j\theta} - e^{-j\theta})/2j$ :

$$V(f) = 2jA/j2\pi f[(e^{j\pi fT} - e^{-j\pi fT})/2j] \quad (\text{eqn 3.14})$$

$$V(f) = A/\pi f[\sin(\pi fT)]$$

Knowing that  $\sin x / x = \text{sinc } x$

$$V(f) = AT/\pi fT[\sin(\pi fT)] \quad (\text{eqn 3.15})$$

$$V(f) = AT \text{sinc}(\pi fT)$$

The sinc function is common in digital electronics and plots as the product of  $\sin(\pi fT)$  and  $1/(\pi fT)$  as in Figure 3.2. In Figure 3.2,  $V(f)$  in the frequency domain is equivalent to  $v(t)$  in the time domain. Note that as the pulse,  $T$ , gets longer,  $1/T$  gets smaller or the first zero crossing of the sinc function occurs at a lower and lower frequency.

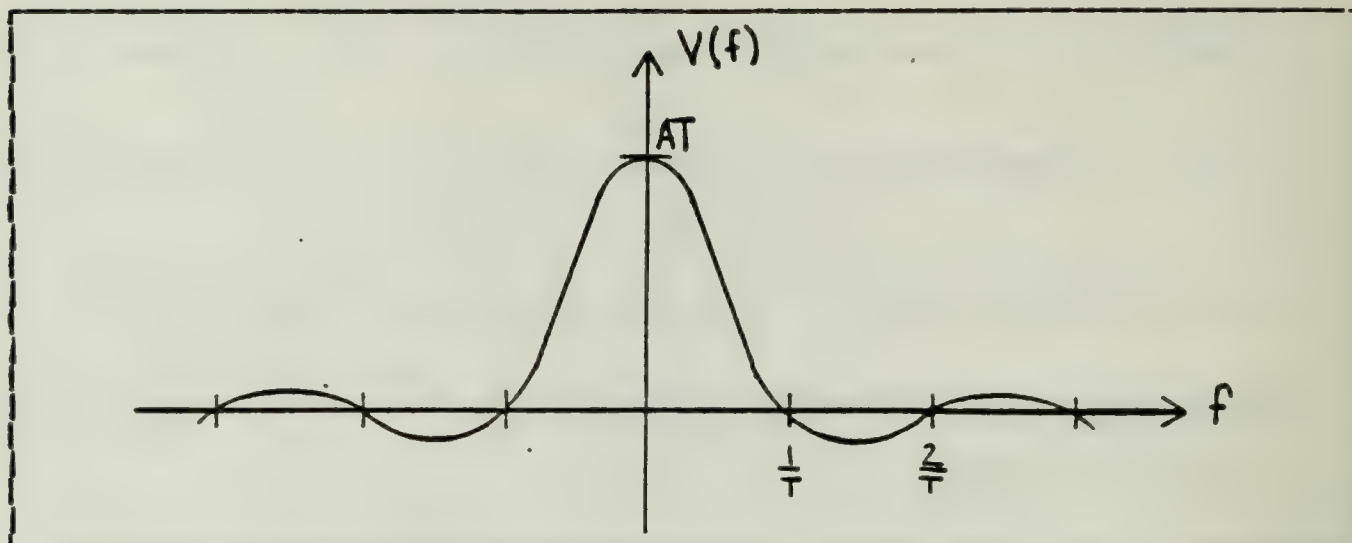


Figure 3.2 Plot of Sinc Function

1. Amplitude and Phase Spectrum

Recall that  $V(\omega)$  can be represented as in equation 3.16. The cosine term is the real part while the sine term is the imaginary part. By referencing Figure 3.3, a brief review of the complex plane is accomplished and its relationship to the amplitude and phase spectrum of a given voltage is represented. See equations 3.17 through 3.20.

$$V(\omega) = \int_{-\infty}^{\infty} v(t) \cos(\omega t) dt - j \int_{-\infty}^{\infty} v(t) \sin(\omega t) dt \quad (\text{eqn 3.16})$$

real
imaginary

$|V(f)|$  is called the amplitude spectrum of the given voltage. The amplitude spectrum can also be calculated by using the complex conjugate of the Fourier transform and is always positive as shown in equation 3.21. The phase spectrum,  $\theta$ , of the function in question is represented by the  $\arctan[\text{imaginary}/\text{real}]$  and is illustrated in Figure 3.5.



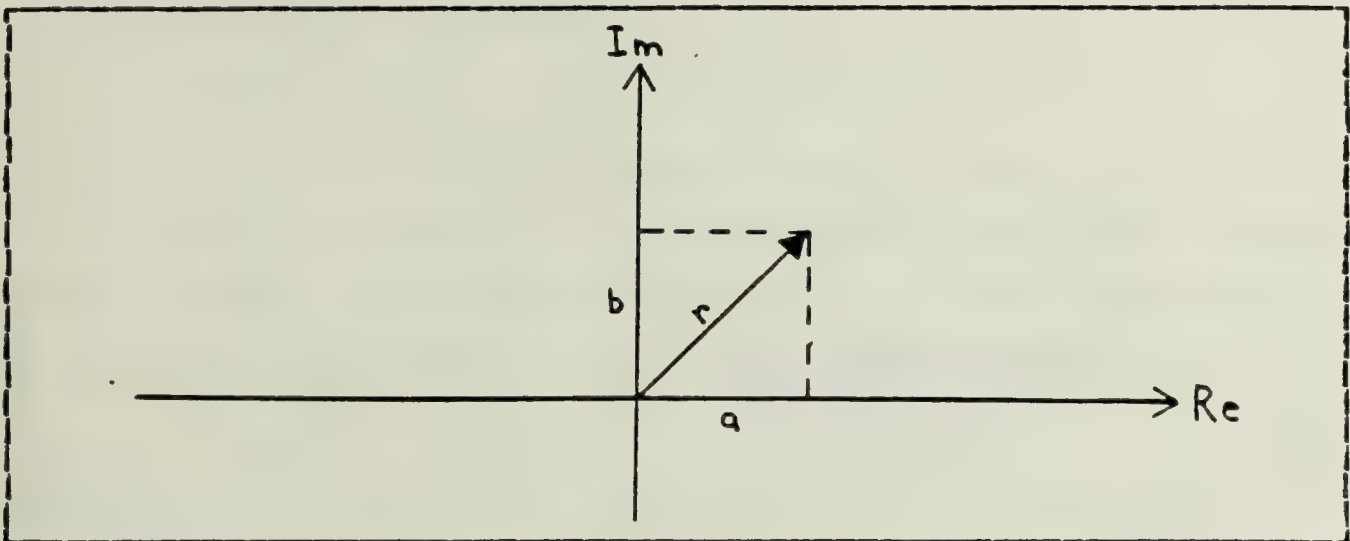


Figure 3.3 Phasor Diagram

$$r = [a^2 + b^2] \quad (\text{eqn 3.17})$$

$$a + jb = [a^2 + b^2]^{.5} e^{j\theta} \quad (\text{eqn 3.18})$$

$$V(f) = [\text{real}^2 + \text{imaginary}^2]^{.5} e^{j\theta} \quad (\text{eqn 3.19})$$

$$|V(f)| = [\text{real}^2 + \text{imaginary}^2]^{.5} \quad (\text{eqn 3.20})$$

since  $|e^{j\theta}| = 1$

$$|V(f)| = [V(f) \cdot V^*(f)]^{.5} \quad (\text{eqn 3.21})$$

## 2. Properties of the Fourier Transform

Several properties of the Fourier transform are useful in the study of digital signals. They are represented here without proof and without a great deal of detail.

### a. Linearity

if  $v_1(t) \leftrightarrow V_1(f)$  and

if  $v_2(t) \leftrightarrow V_2(f)$  then

$$\mathcal{F}[av_1(t) + bv_2(t)] = aV_1(f) + bV_2(f)$$

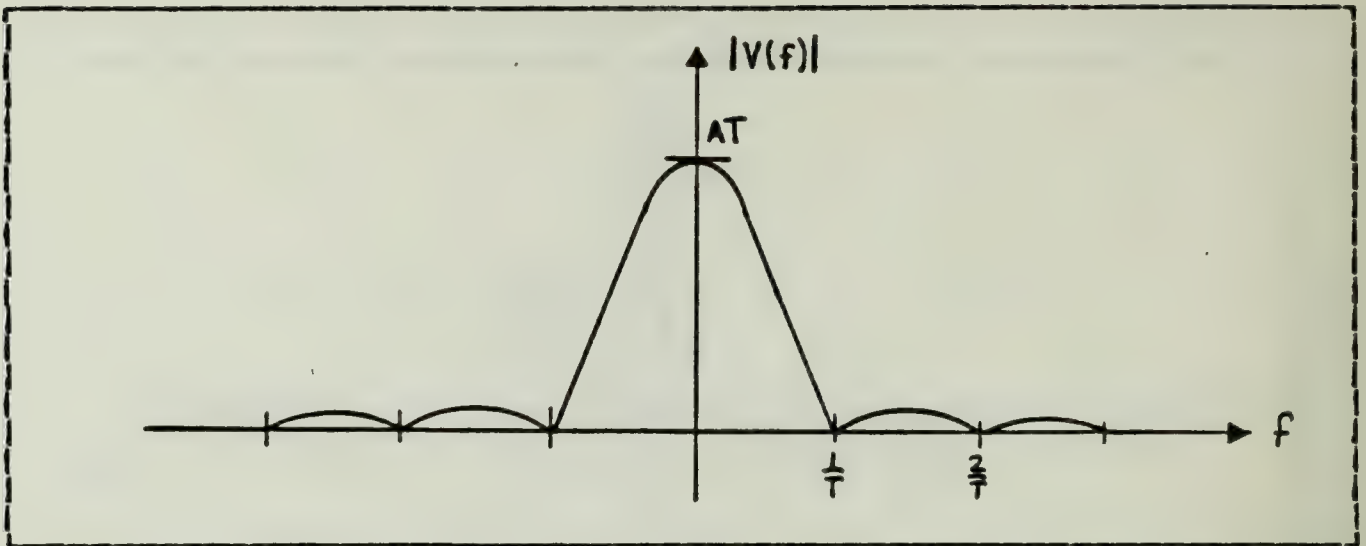


Figure 3.4 Amplitude Spectrum of Square Wave

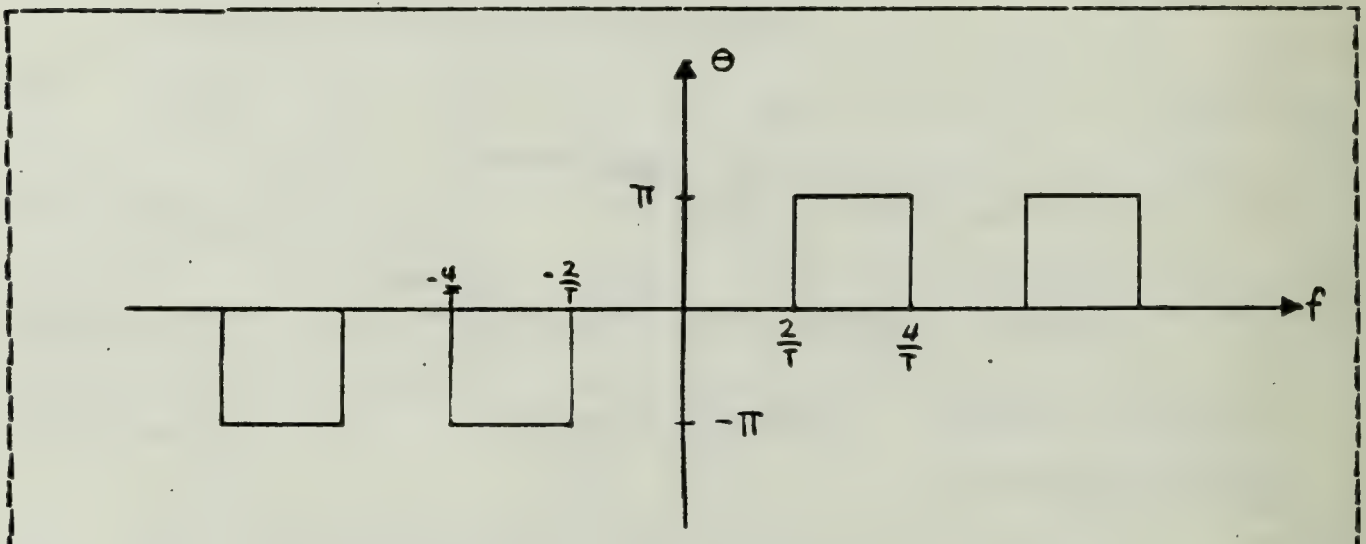


Figure 3.5 Phase Spectrum of Square Wave

b. Time-delay

$$\mathcal{F}[v(t-t_0)] = V(f) e^{-j\omega t_0}$$

Note that the amplitude spectrum of the delayed version is the same as the amplitude spectrum of the undelayed version. It is comforting to note that the Fourier transform of a pulse is the same tomorrow as it is today.

c. Scale change

$$\mathcal{F}[v(at)] = 1/|a| V(f/a)$$

d. Frequency translation

$$v(t) \cos(2 \pi f_c t) \leftrightarrow \frac{1}{2}[V(f+f_c) + V(f-f_c)]$$

$v(t)$  is any voltage

$\cos(2 \pi f_c t)$  is a carrier wave of frequency  $f_c$

Since understanding of this very important property of the Fourier transform is essential to the understanding of digital signal modulation it is expanded slightly here.

As we have seen, the Fourier representation of a voltage pulse of amplitude  $A$  and duration  $T$  is the sinc function of amplitude  $AT$  as illustrated in Figure 3.6.

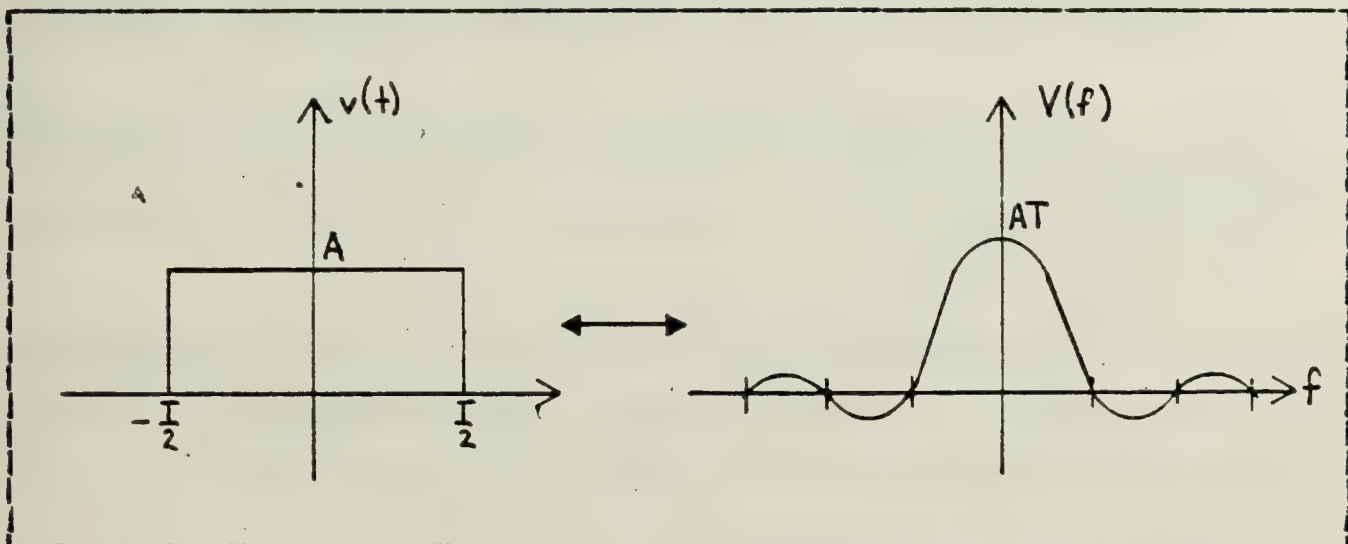


Figure 3.6 Square Wave in Time/Sinc Function in Frequency

The translation is the product of  $v(t)$  and in this case  $\cos(2 \pi f_c t)$ . In the time domain this product only exists in the interval between  $-T/2$  and  $T/2$  as in Figure 3.7. The significance of this translation and its relationship to the bandwidth of the voltage will be discussed further in the section dealing with bandwidth.

e. Differentiation

$$d v(t)/dt \leftrightarrow j \omega V(f)$$

A differentiator could be used as a clock for timing but would never be used in the presence of noise since the



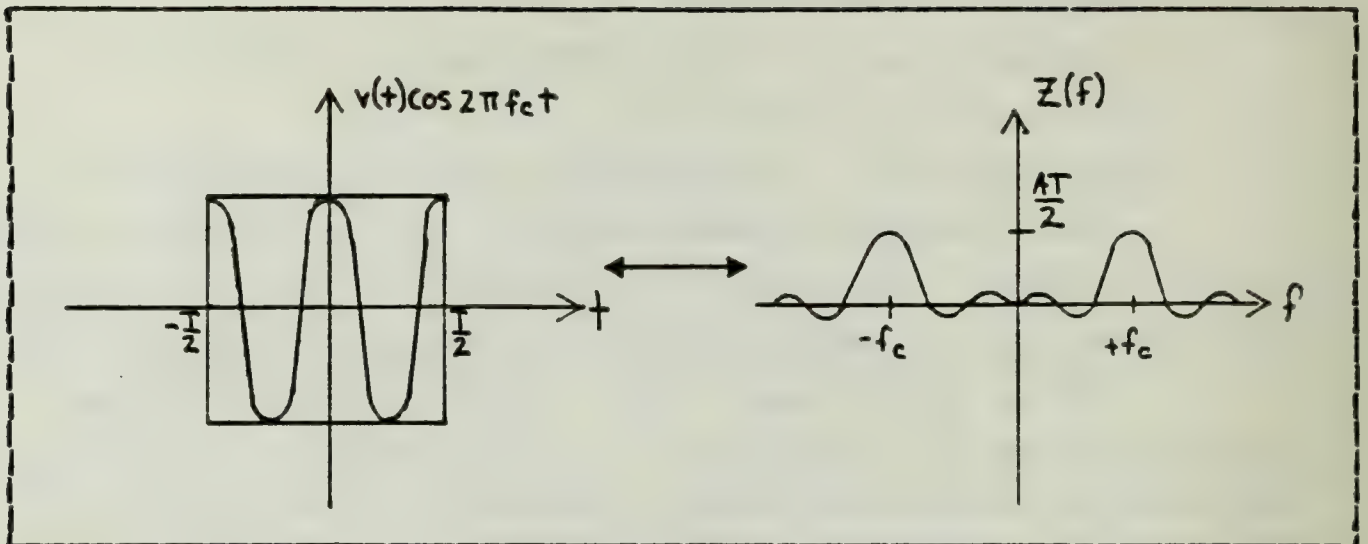


Figure 3.7 Multiplication and Translation Diagrams

result is exaggerated in the presence of high frequencies because  $\omega = 2\pi f$ .

f. Integration

$$v(t) dt \leftrightarrow 1/\omega V(f)$$

An integrator could be used to reduce the effects of noise since  $\omega = 2\pi f$  is in the denominator tending to deemphasize the presence of high frequency noise.

B. THE SAMPLING THEOREM

Essential to the understanding of digital communications is the sampling theorem which was first introduced by Nyquist in 1928 [Ref. 9], and later by Shannon in 1948 [Ref. 10]. The sampling theorem states that any voltage can be uniquely represented by appropriately spaced sample values of the original voltage. More correctly stated, the sampling theorem places limits on the accuracy with which a signal can be represented.

The implication is that an analog signal can be represented digitally or by a set of numbers, i.e., samples. A description of the sampling theorem follows.

Given any analog voltage,  $v(t)$  as in Figure 3.8, the sampling theorem says that the entire analog signal is not required to accurately represent the voltage but only samples of it, call them  $v_s(t)$ . Figure 3.9 shows samples of a representative analog voltage. Samples can be taken of  $v(t)$  at every  $T$  seconds for a period of seconds. This can be accomplished by the use of a voltage clock, call it  $v_c(t)$ . The sampling can be viewed graphically as a block diagram representing an analog voltage multiplier as shown in Figure 3.10. To be of further use in the understanding of digital communications we are interested in a frequency description of the sample voltage,  $v_s(t)$ . Note that the system which describes the obtaining of  $v_s(t)$  involves a voltage multiplication. It was demonstrated in the proceeding section that voltage multiplication amounted to frequency translation, a property of the Fourier transform.

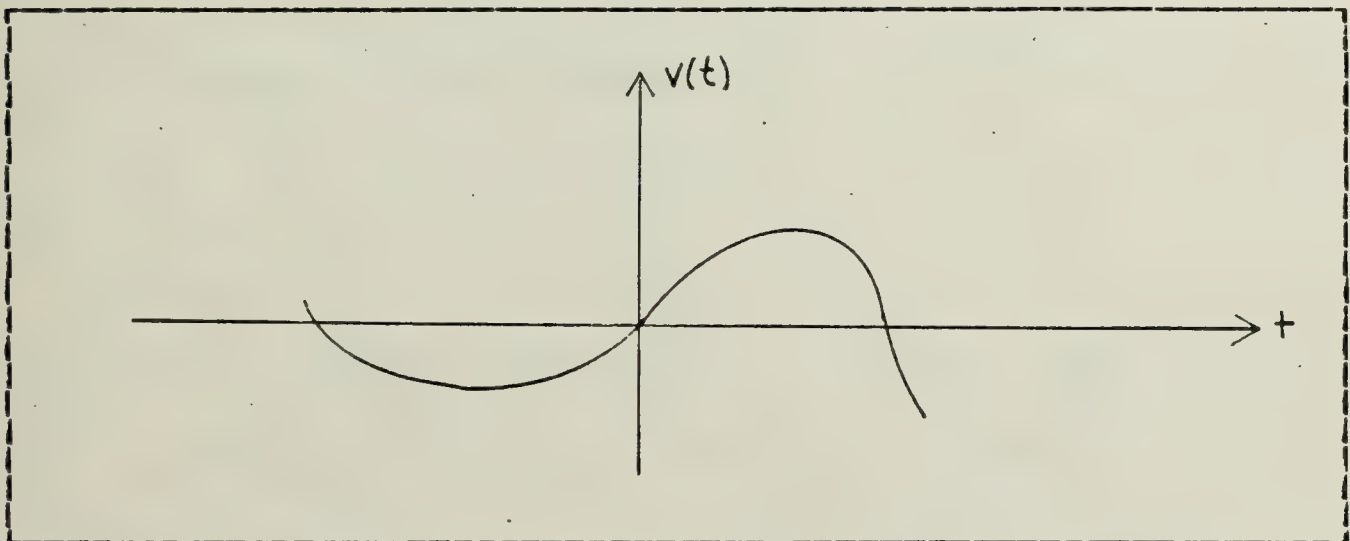


Figure 3.8 Representative Analog Voltage

First it is necessary to find the frequency description of the clock,  $v_c(t)$ , Let  $v_c(t)$  be a periodic square wave of height 1 and duration  $d$  as in Figure 3.11. Since  $v_c(t)$  is periodic, it can be shown that the Fourier series representing  $v_c(t)$  is given by equation 3.22.

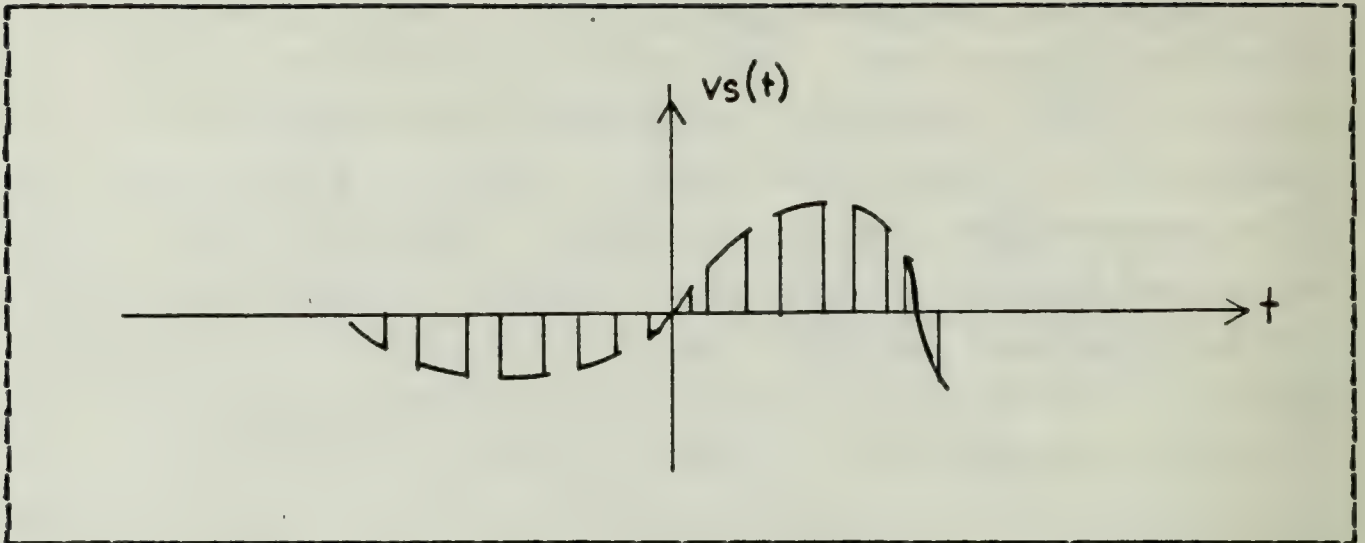


Figure 3.9 Samples of a Representative Voltage

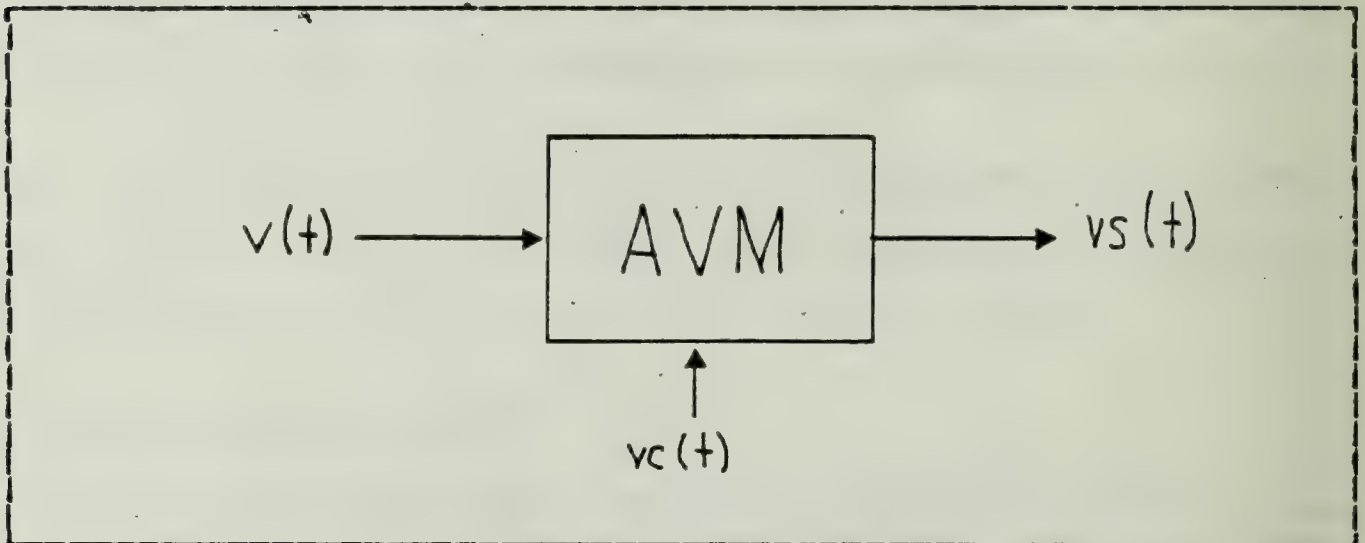


Figure 3.10 Analog Voltage Multiplier

Again it is noted that  $v_s(t)$ , the sample voltage, is the product of  $v(t)$ , the original analog voltage, times  $v_c(t)$ , the clock voltage. In other words  $v_s(t) = v(t)$  times a series of cosine terms.

$$v_c(t) = a_0 + \sum_{n=1}^{\infty} a_n \cos(2 \pi n f_c t) \quad (\text{eqn 3.22})$$

$a_0$  and  $a_n$  are left unevaluated



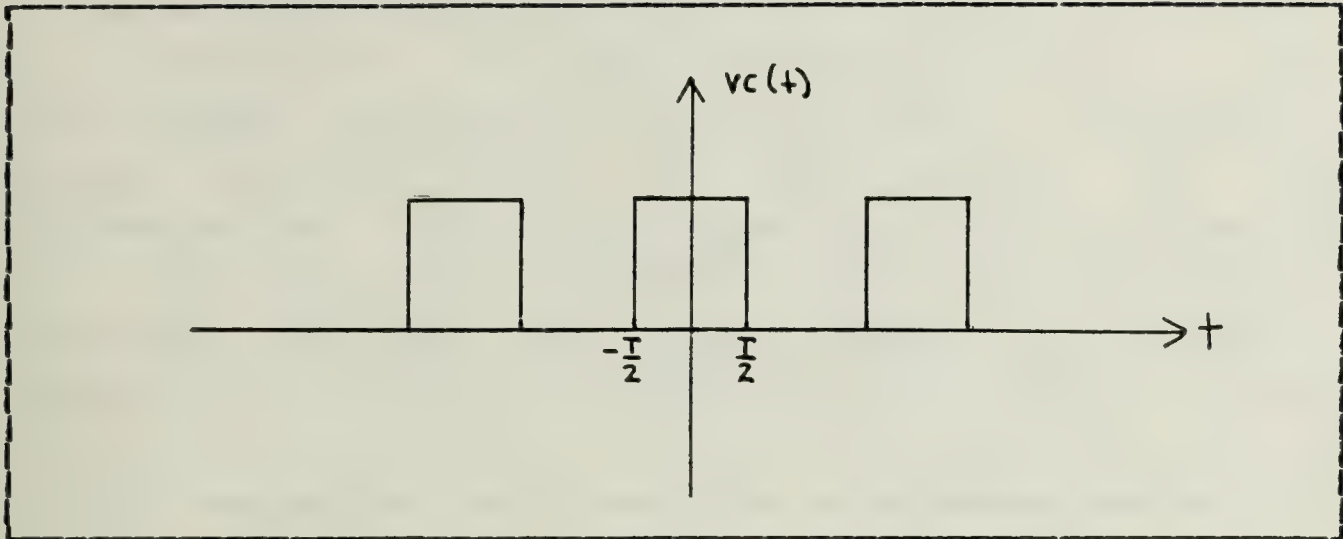


Figure 3.11 A Representative Voltage Clock

$$v_s(t) = v(t) \left[ a_0 + \sum_{n=1}^{\infty} a_n \cos(2 \pi n f_c t) \right] \quad (\text{eqn 3.23})$$

Now, assuming any form of the Fourier transform of  $v(t)$ , as in Figure 3.12 and since  $v_s(t)$  is a product and by utilizing the frequency translation property of the Fourier transform, the following is obtained as a representation of the frequency spectrum of  $v_s(t)$ . See Figure 3.13.

Remember that the curve highlighted in the box in Figure 3.13 is the Fourier representation of  $v(t)$ , the original signal. This original signal can now be recovered by filtering with an appropriate low pass filter of bandwidth equal to or greater than  $B$ . This low pass filter would only permit the reception of that portion of the signal which represents the original voltage.

The only question left to resolve is how often to take a sample. Again referring to the diagram in Figure 3.13, it is noted that in order to prevent any overlap of successive translations (aliasing)

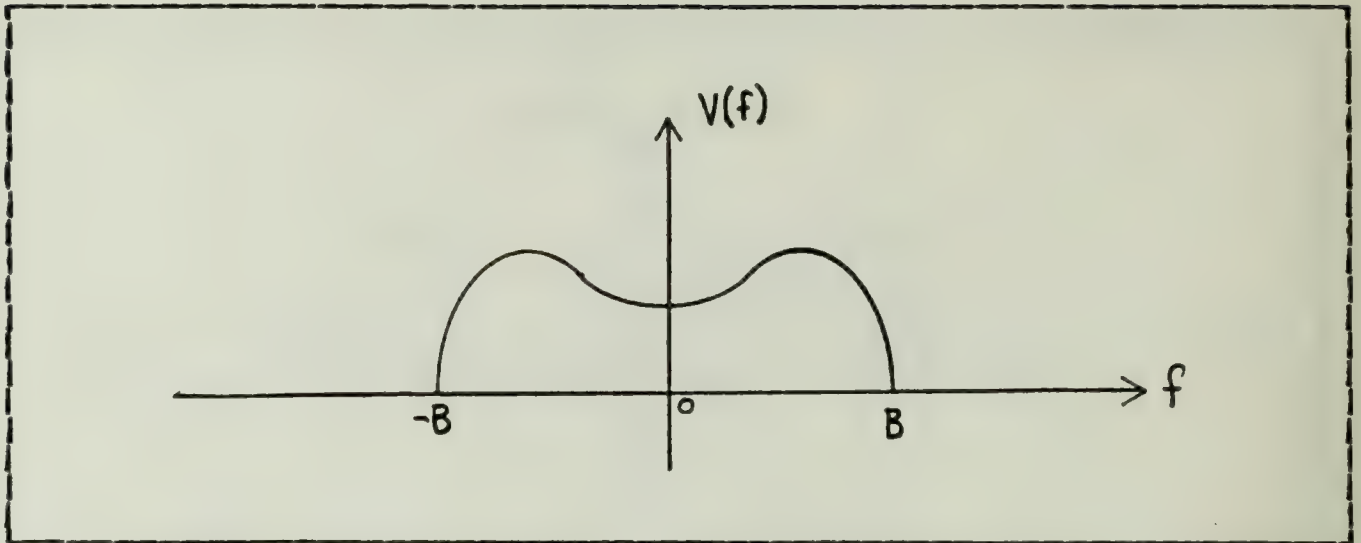


Figure 3.12 Fourier Transform of  $v(t)$

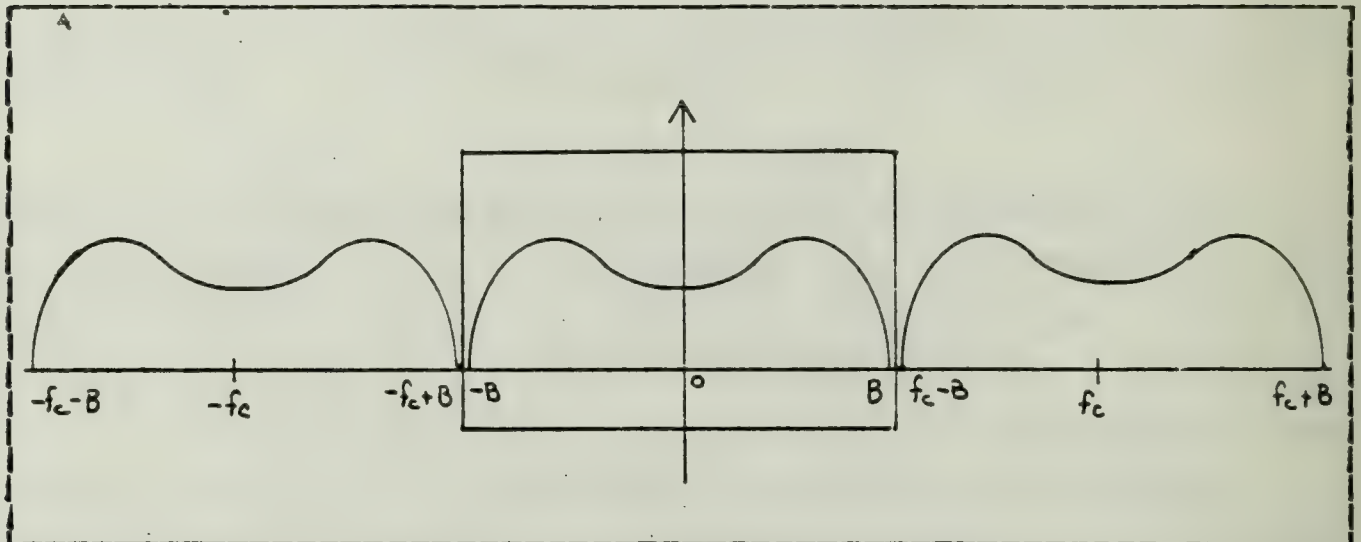


Figure 3.13 Fourier Transform of  $v_s(t)$

$$f_0 - B \gg B$$

$$f_0 \gg 2B$$

or the frequency of the clock,  $f_c$  (the sampling rate =  $1/T$ ) must be strictly greater than 2 times the largest significant Fourier frequency component present. Since  $B$  will vary with different  $v(t)$ , the sampling rate necessary to uniquely recover that  $v(t)$  also varies.

In summary on the sampling theorem, it can be said that any analog signal can be represented as certain sample values spaced the appropriate distance apart. The sampling theorem places limits on the accuracy of this representation. These sample values can then be transmitted from one position to another using analog to digital conversion and any one of a variety of modulation techniques. It has been shown that the Fourier frequency representation of the sample is not equivalent to the Fourier frequency representation of the source voltage; however, the source voltage can be recovered at the receiving end by filtering. A block representation of the system is shown in Figure 3.14.

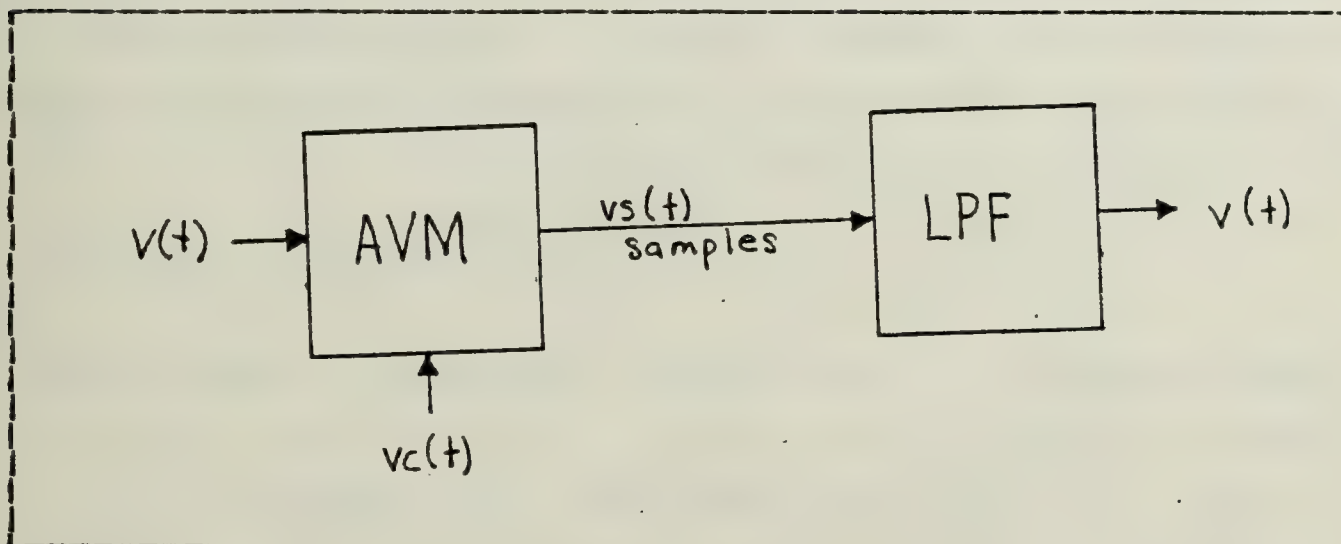


Figure 3.14 Analog Voltage Multiplier and Low Pass Filter

### C. THE AUTOCORRELATION FUNCTION

The above concept certainly seems simple enough. If a signal is present and it is desired to transmit it from point A to point B, all that has to be done is to take appropriately spaced samples of the signal, encode them,



somehow transmit them to point B, filter the received signal and the uniqueness of the original voltage has been reproduced. Unfortunately it isn't quite that easy. The reason it is not, is due to the presence of noise. The sources of noise will not be discussed here, however, the effects of noise in general terms and how a faint signal can be recovered in the presence of noise will be discussed.

The time varying descriptor of voltage and the frequency varying descriptor of voltage have been introduced. Both are precise mathematical descriptions of something (voltage) that is deterministic. That is, it can be described in mathematical or graphical terms during any period of time. Such is not the case for noise corrupted voltages which are entirely random. Therefore other descriptors of voltages must be employed in the presence of noise. They are partial descriptors of voltage since a random signal cannot be described with precision. These partial descriptors are the

1. Autocorrelation function and the
2. Probability density function (p.d.f.)

It is important here to note the difference between the source voltage at the receiver and the sample voltage. The source voltage at the receiver,  $v_{sr}(t)$ , is the digitally converted sample voltage,  $v_s(t)$ , modulated onto a carrier by one of the digital modulation techniques yet to be discussed. Understanding the autocorrelation function is the basis for understanding how a decision is made that  $v_{sr}(t)$  is present at the receiver in the presence of noise. It should be remembered that the exact form of the carrier is known at both the transmitter and receiver.

First of all, an additive noise model is assumed where the signal at the receiver,  $v_r(t)$ , is equal to the signal which is being transmitted,  $v_{sr}(t)$ , (i.e., the digital samples of  $v(t)$  modulated onto the carrier), plus random noise,  $v_n(t)$

$$v_r(t) = v_{sr}(t) + v_n(t) \quad (\text{eqn 3.24})$$

$v_r(t)$  = voltage at the receiver

$v_{sr}(t)$  = signal voltage at the receiver

$v_n(t)$  = noise voltage at the receiver

If it is realized that most receivers operate on the principle of detection of DC voltage, the problem becomes one of rectification of the received signal and determination if the DC voltage, characteristic of the transmitted signal is present.

A common type of rectification of an analog voltage involves squaring the input waveform. If the given signal at the receiver is  $v_r(t)$  as in equation 3.24 above, squaring the waveform introduces the square of not only the desired signal but also the square of the noise term. If instead of squaring  $v_r(t)$  and introducing or at least not eliminating the noise,  $v_r(t)$  is multiplied by  $v_{sr}(t)$ , the results in equation 3.25 are obtained.

$$v_r(t) \cdot v_{sr}(t) = v_{sr}^2(t) + v_{sr}(t)v_n(t) \quad (\text{eqn 3.25})$$

By averaging, all that remains is  $\overline{v_{sr}^2(t)}$  since the average of  $v_{sr}(t)v_n(t)$  is zero because  $\overline{v_n(t)}$  is random and the average of any random voltage is zero. The DC component of  $v_r(t) \cdot v_{sr}(t)$  is  $\overline{v_{sr}^2(t)}$ . Any remaining AC component can be removed by a low pass filter. Graphically in block diagram form this receiver looks like Figure 3.15 illustrated below. If  $v_{sr}^2(t) > 0$  then  $v_{sr}(t)$  is present in the signal  $v_r(t)$ . If  $v_r(t)$  is pure noise or some other signal is present exclusively, then the output of the receiver will be 0.

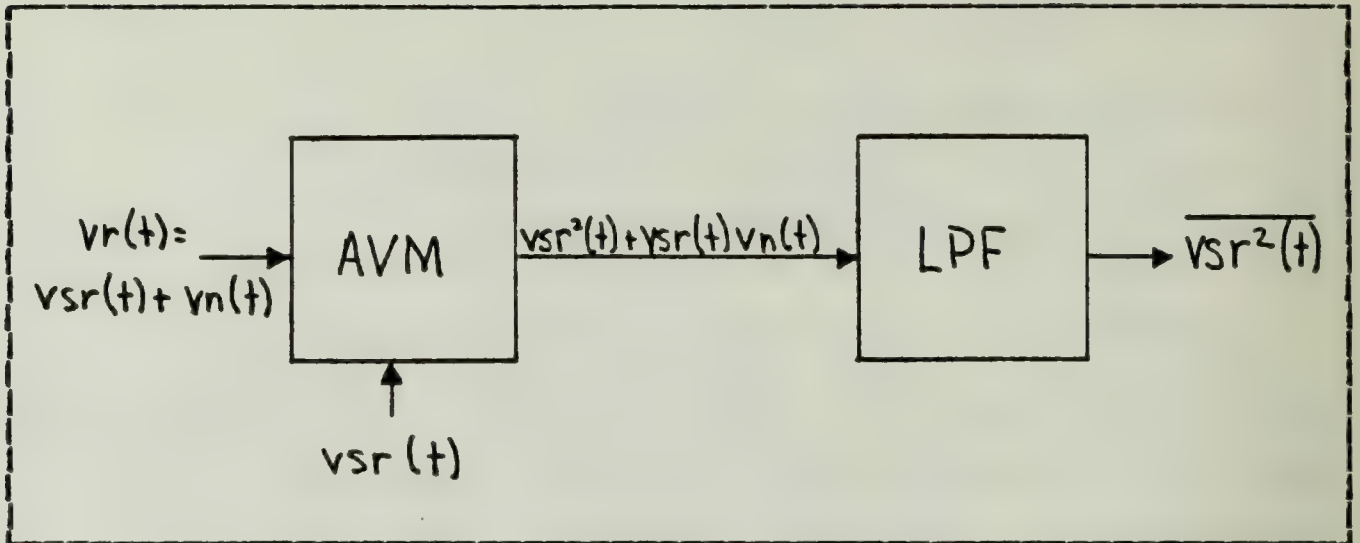


Figure 3.15 Rectification through an AVM and LPF

In practice, however, the exact waveform of  $v_{sr}(t)$  may not be known or it may be a time delayed or distorted version of the original signal when it arrives at the receiver. If the distortion or delay is significant enough there will be little agreement or "correlation" in the receiver.

The solution is to multiply the received signal by a series of time delayed approximations of the transmitted signal. The signal is present whenever the output of the series of voltage multipliers or correlators exceeds a certain threshold approaching  $v_{sr}^2(t)$ . The concept is illustrated in Figure 3.16.

For the concept illustrated in Figure 3.16 to work, the average value of  $v_r(t) \cdot v_{sr}(t-p)$  must have a DC component.

$$v_r(t) \cdot v_{sr}(t-p) = v_{sr}(t)v_{sr}(t-p) + v_n(t)v_{sr}(t-p) \quad (\text{eqn 3.26})$$

$$\overline{v_r(t) \cdot v_{sr}(t-p)} = \overline{v_{sr}(t)v_{sr}(t-p)} + \overline{v_n(t)v_{sr}(t-p)} \quad (\text{eqn 3.27})$$



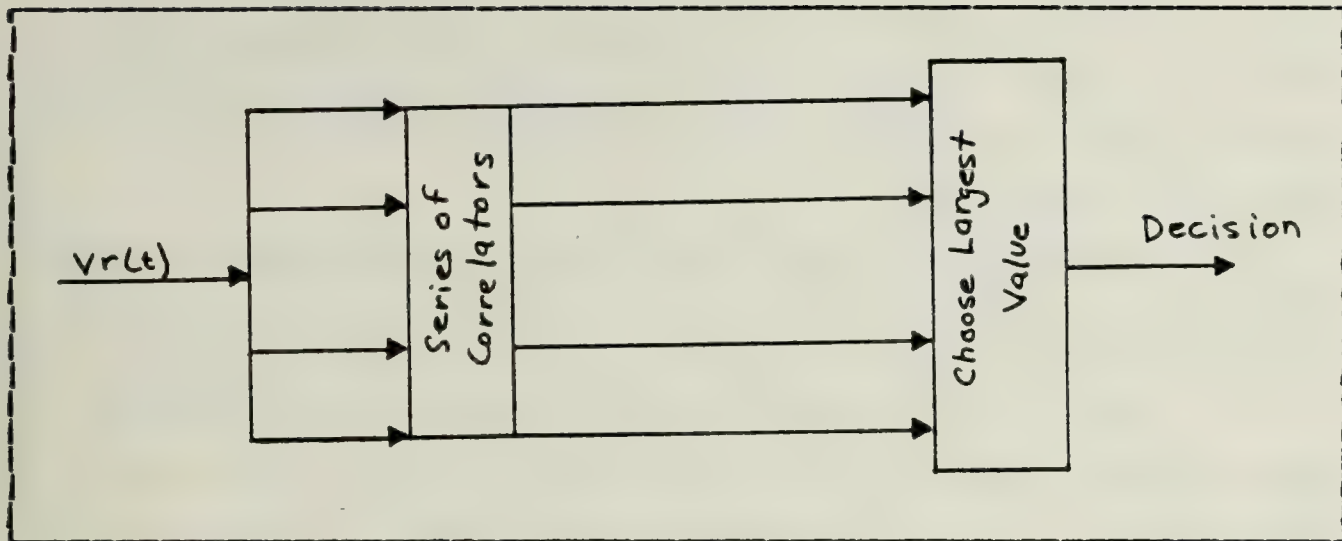


Figure 3.16 Correlation Device

But the average of  $v_n(t)v_{sr}(t-p) = 0$  since  $\overline{v_n(t)}$  is random and the results of equation 3.28 are obtained.

$$\overline{v_r(t) \cdot v_{sr}(t-p)} = \overline{v_{sr}(t)v_{sr}(t-p)} \quad (\text{eqn 3.28})$$

What is of interest is the average value of  $v_{sr}(t)v_{sr}(t-p)$ . The autocorrelation function (ACF) will be defined as that average value of  $v_{sr}(t)v_{sr}(t-p)$ . The mathematical representation of the autocorrelation function for a continuous voltage is represented in equation 3.29 while the autocorrelation function for a pulse voltage is presented as equation 3.30.

$$\text{ACF} = R_{vv}(p) = \frac{1}{2T} \int_{-T}^T v(t)v(t-p) dt \quad (\text{eqn 3.29})$$

$$\text{ACF} = C_{vv}(p) = \int_{-\infty}^{\infty} v(t)v(t-p) dt \quad (\text{eqn 3.30})$$

The autocorrelation function is a measure of the degree to which two identical signals which are corrupted in some manner are alike. The ACF of two signals which are identical, not distorted in any way and occurring at exactly

the same time, i.e.,  $p = 0$  has a maximum value. On the other hand there will be very little correlation between two identical time variant signals when the time difference between them is great. In this case the autocorrelation function is near zero.

A similar concept is used within the context of signal comparison. This concept is the crosscorrelation which is a measure of the degree to which 2 different signals are alike. When the crosscorrelation between voltage  $v_{sr}$  and random noise voltage  $v_n$  is 0 there is no relationship or correlation.

Again let us assume that the signal at the receiver is a noise disrupted version of the signal originally transmitted.

$$v_r(t) = v_{sr}(t) + v_n(t) \quad \text{(eqn 3.31)}$$

In the receiver, the time delayed version of the signal is applied to the incoming signal and the average is formed as before and shown in Figure 3.17.

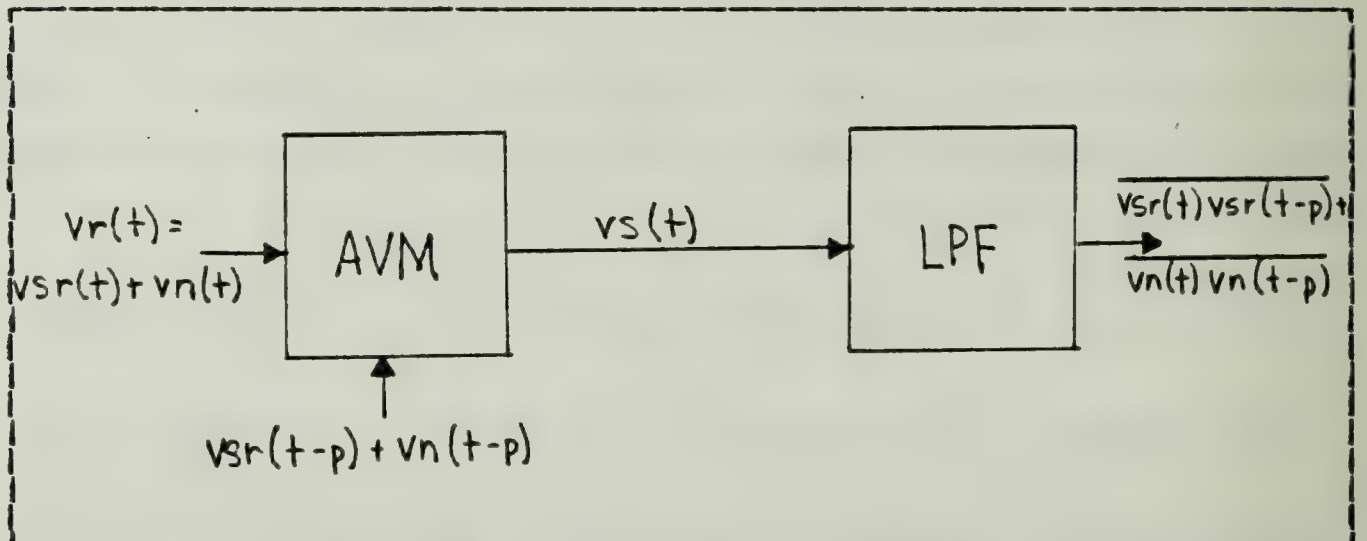


Figure 3.17 Voltage Correlator

$$\begin{aligned}
 & \overline{[v_{sr}(t) + v_n(t)] \cdot [v_{sr}(t-p) + v_n(t-p)]} = & \text{(eqn 3.32)} \\
 & \overline{v_{sr}(t)v_{sr}(t-p)} + \overline{v_n(t)v_{sr}(t-p)} + \\
 & \overline{v_{sr}(t)v_n(t-p)} + \overline{v_n(t)v_n(t-p)} =
 \end{aligned}$$

$$R_{v_{sr}} v_{sr}(p) + R_{v_n} v_{sr}(p) + R_{v_{sr}} v_n(p) + R_{v_n} v_n(p)$$

$$R_{v_n} v_{sr}(p) = 0$$

$$R_{v_{sr}} v_n(p) = 0; \text{ because average } v_n = 0$$

Therefore the result is simply  $R_{v_{sr}} = v_{sr}(p) + R_{v_n} v_n(p)$  or the autocorrelation function of the desired signal plus the autocorrelation function of the noise. Assuming the shape of the autocorrelation function of both components (both the original signal and the noise which vary with  $p$ , time) is known, what is done in practice is to vary the value of  $p$  until the ratio of  $R_{v_{sr}} v_{sr}(p) / R_{v_n} v_n(p)$  is a maximum or the autocorrelation function of the signal is a maximum while the autocorrelation of the noise is minimum and the signal is recovered.

#### D. THE MATCHED FILTER

The matched filter is a linear filter which has the characteristic response desired for optimum reception of the desired signal. In other words, we desire the response of the system to the linear filter to be in some way proportional to the autocorrelation function of the desired signal and in no way related to the autocorrelation function of the noise. This system could be illustrated as in Figure 3.18.



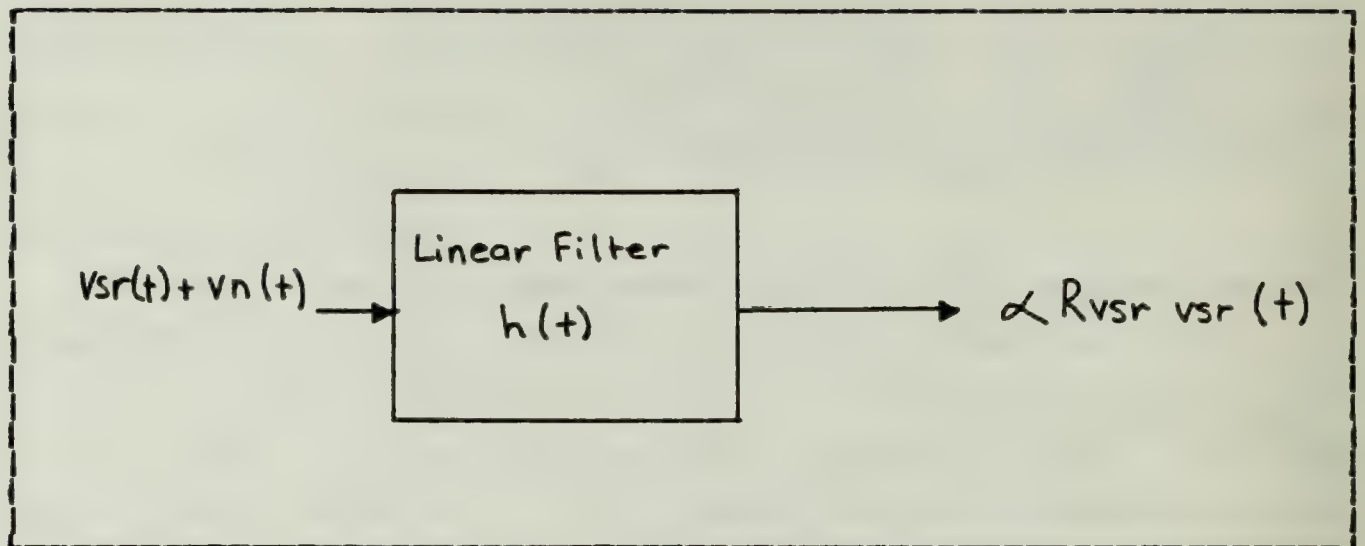


Figure 3.18 Matched Filter Block Representation

The question is what should the makeup or  $h(t)$ , (i.e., the response of the linear filter) be? The output of a linear system is a convolution so the output of the linear filter,  $h(t)$  to an input  $v_{sr}(t)$  would be  $v_{sr}(t) * h(t)$ , where  $*$  denotes convolution.

In the previous section it was stated that the autocorrelation function,  $R_{v_{sr}} v_{sr}(p)$ , a function of  $p$ , is given by

$$R_{v_{sr}} v_{sr}(p) = \int_{-\infty}^{\infty} v_{sr}(t) v_{sr}(t-p) dt \quad (\text{eqn 3.33})$$

By simply changing variables, equation 3.33 can be rewritten as a function of time,  $t$ .

$$R_{v_{sr}} v_{sr}(t) = \int_{-\infty}^{\infty} v_{sr}(p) v_{sr}(p-t) dp \quad (\text{eqn 3.34})$$

From convolution theory it is known that the response of a linear filter as above,  $v_{sr}(t) * h(t)$  can be expressed mathematically as in equation 3.35.

$$v_{sr}(t) * h(t) = \int_{-\infty}^{\infty} v_{sr}(p) h(t-p) dp \quad (\text{eqn 3.35})$$

As stated earlier, the desired output of the linear filter is the autocorrelation function  $R_{vSR} vSR(t)$ . Note the similarities between the last two equations. If in equation 3.35 a simple substitution of  $h(t) = vSR(-t)$  is performed, the desired results are obtained.

$$vSR(t) * h(t) = \int_{-\infty}^{\infty} vSR(p) vSR(p-t) dp \quad (\text{eqn 3.36})$$

$$= R_{vSR} vSR(t)$$

Therefore, for every signal that is desired to be recovered, the matched filter will do the job very nicely if the response of that filter,  $h(t)$  is equal to the inverse of the signal that is desired to be detected.

#### IV. ANALOG TO DIGITAL CONVERSION

Key to the understanding of digital communications is how the analog information is converted to digital information. As illustrated earlier, each analog voltage or signal can be represented by appropriately spaced sample values. These sample values are still analog in that they can take on any value in the range of the original analog signal. What is desired is to change this infinite set of decimal numbers to a finite set of decimal numbers. This is called analog to digital conversion. Common types of A to D conversion include Pulse Code Modulation (PCM), Differential Pulse Code Modulation (DPCM), Delta Modulation (DM), Pulse Amplitude Modulation (PAM), Pulse Duration Modulation (PDM), and Pulse Position Modulation (PPM). PCM, DPCM, and DM will be examined in this chapter because of their widespread usage and easy application to digital technology.

##### A. PULSE CODE MODULATION (PCM)

PCM is the most widely used A to D conversion technique. It involves assigning the sample value obtained in sampling to one of several quantization levels within the range of the voltage sampled and then representing those quantization levels by various binary code words.

For example, the highest significant frequency component in human speech is 3300 Hz. In order to capture accurately the signal produced in speech, one would require a sampling rate greater than or equal to  $2B$  or  $2(3300) = 6600$  samples/sec. The telephone company uses 8000 samples/sec.

$$B = 3300 \text{ Hz}$$

$$f \geq 2B = 6600 \text{ samples/sec}$$

use  $f = 8000$  samples/sec to ensure no aliasing



If a signal is sampled at the rate of 8000 samples/sec, then in the transmission of those samples, assuming the samples are transmitted immediately and not delayed, the transmission rate must be

$$T_s = 1/f$$

$$T_s = 1/8000 \text{ samples/sec}$$

$$T_s = 125 \text{ microsec/sample}$$

The number of quantization levels employed to represent the various analog sample values is arbitrary. The greater the number of levels, the more accurate the reconstruction of the original signal but the more rapid the data transmission rate must be (in all cases there will be some error present after recovery). Consider again the example of voice as illustrated in Figure 4.1.

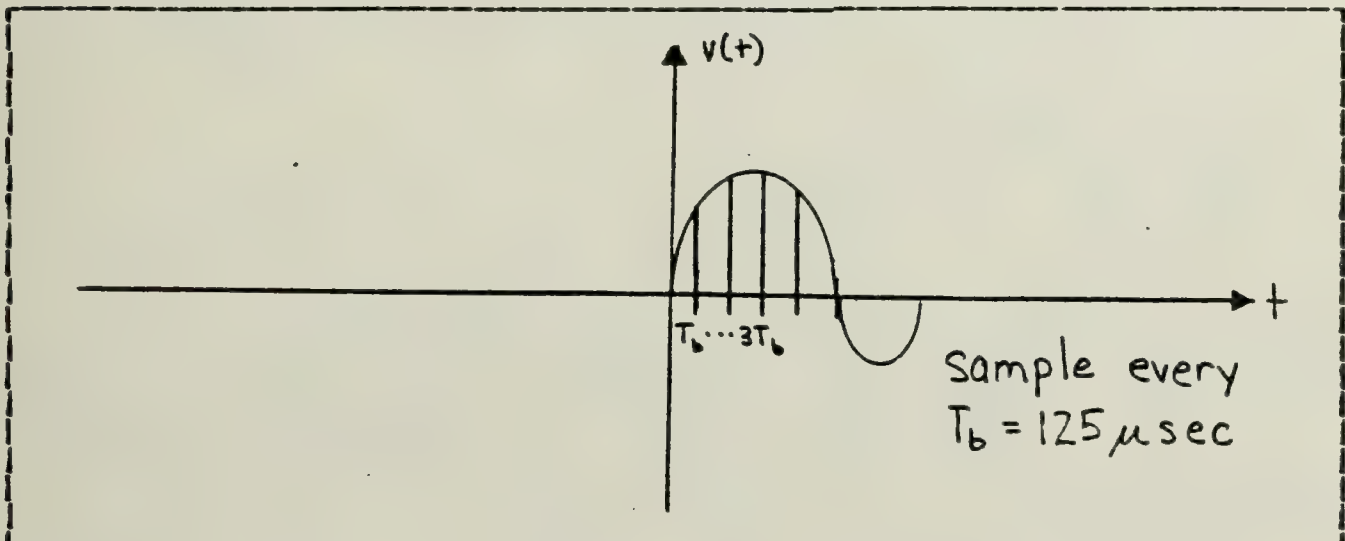


Figure 4.1 Samples of Voltage

Assume it is desired to represent each sample value (decimal number) with an 8 bit binary code word. Then the number of quantization levels is given by equation 4.1.

The spacing between quantization levels is the range of voltages to be represented divided by the number of quantization levels. The analog to digital conversion takes

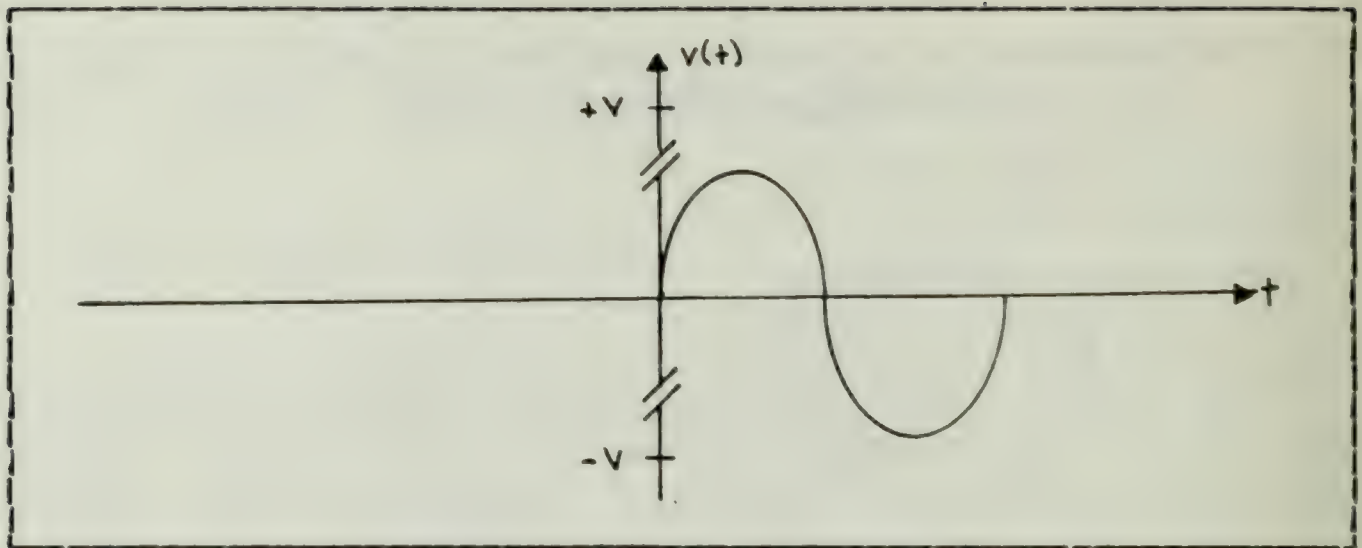


Figure 4.2 Analog Nature of Samples

$$\# \text{quantization levels} = 2^{*n} \quad (\text{eqn } 4.1)$$

for  $n = 8$  bits/binary code word

$$\begin{aligned} \# \text{quantization levels} &= 2^{*8} \\ &= 256 \end{aligned}$$

place by determining into which quantization level the respective sample values fall and assigning that sample value the binary number associated with that quantization level.

Now each sample value is assigned to an eight bit binary code word. For voice transmission it has been shown that the minimum transmission rate was  $T_s = 125$  microsec/sample. The bit rate is then simply equal to the number of bits per sample times the inverse of the transmission rate as illustrated in equations 4.2 and 4.3.

If voice was to be time multiplexed onto a single channel, the bit rate for that channel would be the number of signals per channel times the bit rate for the most time intensive signal as illustrated in equation 4.4.

$$\begin{aligned} \text{bit rate} &= (n \text{ bits/sample}) (1 \text{ sample}/T \text{ sec}) && \text{(eqn 4.2)} \\ \text{bit rate} &= (8 \text{ bits/sample}) (1 \text{ sample}/125 \text{ microsec}) \\ &= 64 \text{ kbps} \end{aligned}$$

$$\begin{aligned} \text{bit duration} &= 1/\text{bit rate} && \text{(eqn 4.3)} \\ &= 1/64 \text{ kbps} = 15.6 \text{ microsec/bit} \end{aligned}$$

$$\begin{aligned} (\# \text{signals/channel}) (\text{bit rate/signal}) &= && \text{(eqn 4.4)} \\ \text{bit rate/channel} & && \end{aligned}$$

In other words, if 24 voice signals are to be multiplexed onto a single channel, the bit rate must be

$$\begin{aligned} \text{bit rate/channel} &\geq && \text{(eqn 4.5)} \\ (24 \text{ signals/channel}) (64 \text{ kbps/signal}) \\ &= 1536 \text{ Mbps} \end{aligned}$$

As can be seen, the data transmission rate increases significantly, necessitating better and better hardware and a wider and wider bandwidth. Advantages should be readily apparent for A to D conversion techniques or modulation techniques that reduce the bit rate required.

Recovery of the analog signal is by a process called Digital to Analog Conversion. All that will be said about D to A conversion is that it is the inverse process of A to D conversion and that a slight error is always introduced during the process due to the discrete nature of the quantization process during A to D and D to A conversion.

## B. DIFFERENTIAL PULSE CODE MODULATION (DPCM)

In DPCM, what is converted to binary numbers is not the quantization level (decimal number) which each sample value



is represented by but the successive differences between quantization levels. The idea is that the range of maximum difference values will be smaller than the range of actual sample values. It is therefore possible to represent that range of delta values with fewer bits/binary word and therefore fewer quantization levels and ultimately a lower bit rate. [Ref. 11]

In actual practice, in a DPCM conversion technique, it is a statistical estimate of each successive sample value which is subtracted from the actual value that is converted to binary code words. The result is the same in that the range of amplitudes is reduced and therefore fewer bits/word are required to represent the sample thereby lowering the data transmission rate required to transmit the signal. [Ref. 11]

### C. DELTA MODULATION (DM)

Delta modulation is a form of DPCM where successive quantization levels of the output differ by only 1 bit. That is to say that any successive quantization level can be represented by varying only one bit of successive output binary code words. In other words a type of gray code is employed. This A to D technique is implemented through the use of a DM coder or linear delta modulator. This DM coder approximates a given input signal with a series of linear segments of uniform slope. A comparison is made between the value of this approximation and the input signal at each sample increment. The sign of this difference value is what is encoded and is used to increment the DM coder in the direction of the input signal. By using the differential sign value of the input signal and the incremented approximation from the DM coder the linear approximation from the linear delta modulator is said to "track" the input

signal. [Ref. 8] Slope overload of this type of modulation technique occurs when the slope of the incoming signal exceeds the ability of the DM system to follow the source at the sampling rate being utilized. [Ref. 11].

## V. DIGITAL SIGNAL MODULATION TECHNIQUES

Digital signal modulation techniques are the methods of encoding information for transmission utilizing digital technology. Factors affecting digital modulation include, but are not limited to, the physics of the method, hardware requirements, bandwidth considerations, power requirements, data transmission rates and error probabilities. It is these aspects which will be explored in the following sections.

### A. DIGITAL MODULATION FORMATS

Modulation is the technique by which the characteristics of one waveform (called the carrier) are varied or modified by the characteristics of another (called the source). The carrier waveform of interest is the sinusoid. It should be obvious that the attributes of a particular sinusoid that differentiate it from every other sinusoid are its amplitude, phase and frequency. It follows that the characteristics of the carrier waveform to be varied by the source are its amplitude, phase, frequency or any combination of the three. This gives rise to the broad general formats of Amplitude Shift Keying (ASK), Phase Shift Keying (PSK) and Frequency Shift Keying (FSK). All other digital modulation techniques are variations of or combinations of these basic formats. Other factors involved in the description of the digital modulation technique being employed include the number of bits being encoded at one time, the employment of error correction techniques and the baseband (source) waveform.



When each bit of the baseband waveform is individually encoded by any of the techniques previously mentioned (ASK, PSK, FSK), the technique being utilized is referred to as binary encoding. When more than 1 bit of the source code is modulated onto the carrier at one time it is called block encoding. Block encoding allows for one of  $m = 2^k$  waveforms where  $k =$  number of bits.

This paper will not go deeply into signal detection or demodulation techniques, however, a basic understanding of what is involved is necessary and again further defines the digital modulation format being employed. In general terms, signal detection is referred to as either coherent or noncoherent detection of the transmitted signal. Coherent detection, perhaps the easiest to understand, is when all possible waveforms of the modulated carrier waveform are available at the receiver and the waveform at the receiver is in phase with the transmitted carrier. Noncoherent detection is involved when the receiver does not have knowledge of the phase of the transmitted information and one of a number of phase estimation techniques must be employed for signal recovery.

## B. HARDWARE

Although significant advances have been made in recent years to improve the quality of hardware associated with satellite communications, most components are not what would be considered "off the shelf items". The hardware components most commonly referred to with regard to digital communications include the sampler, encoder, modulator, multiplexer and transmitter. There are variations of the above hardware requirements necessary for certain types of digital formatting, however, those exceptions will be addressed separately when the individual modulation techniques are examined.

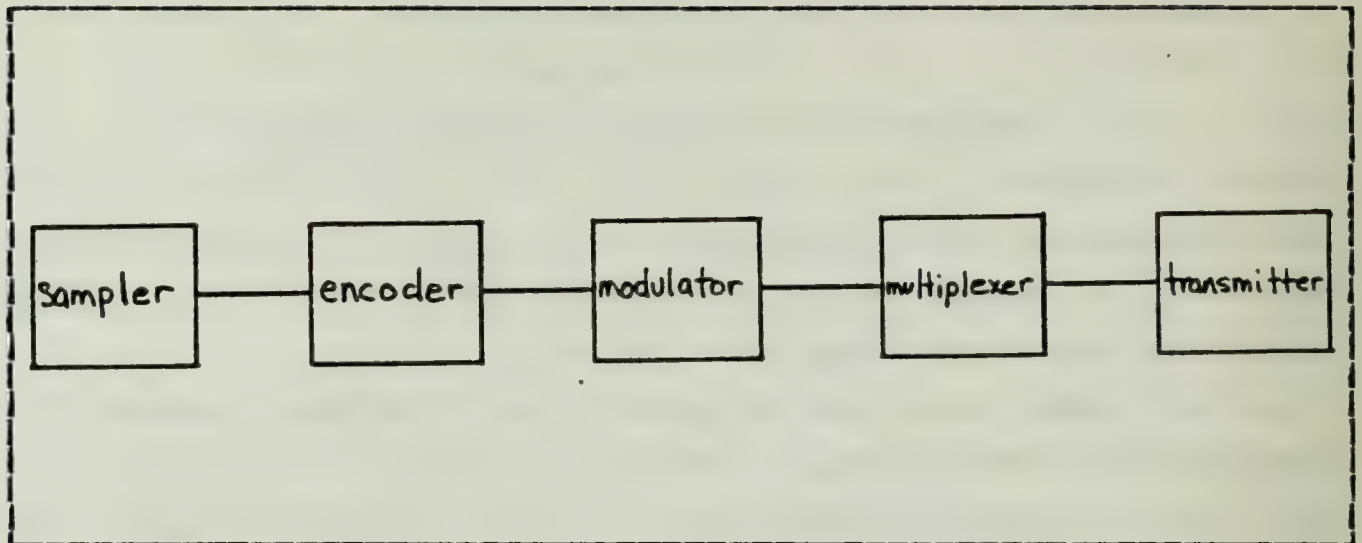


Figure 5.1 Basic Hardware Component Block Diagram

1. Sampler

The sampler is the device or component which samples or extracts those single characteristic values of a naturally occurring analog signal which are ultimately encoded into a digital word by the encoder. The frequency at which this sampler must operate was derived in the discussion of the Sampling Theorem. It was determined that the sampler must operate at a frequency greater than 2 times the highest significant Fourier frequency component of the source voltage. Modern solid state devices capable of taking thousands of samples/sec are available at modest costs.

2. Encoder

Often referred to as the A to D converter (analog to digital converter), the encoder transforms the samples of the analog signal derived from the sampler into a digital format through one of the analog to digital conversion techniques described in the chapter on A to D conversion. These digital bits can be stored for later use, coded,

delayed or used immediately either individually or in groups to modify one of the characteristic qualities of the carrier waveform. When sample values of the analog source signal are converted into digital bits of information, they are simply that, digital bits of information. The carrier can only be modified to represent the source information by interaction with another voltage. Therefore the value of the digital bit (binary), either 0 or 1, is used to generate a baseband waveform or voltage which does the actual modulation of the carrier waveform.

a. Common Baseband Waveforms

It should be obvious that since it is desired to represent binary code words with a representative baseband waveform what is required is two levels of voltage. There are two basic logic schemes for representation of the baseband waveform. They are bipolar or unipolar logic. Bipolar logic involves representing the 0's or 1's of the binary codeword as either  $+V$  or  $-V$  as illustrated in Figure 5.2.

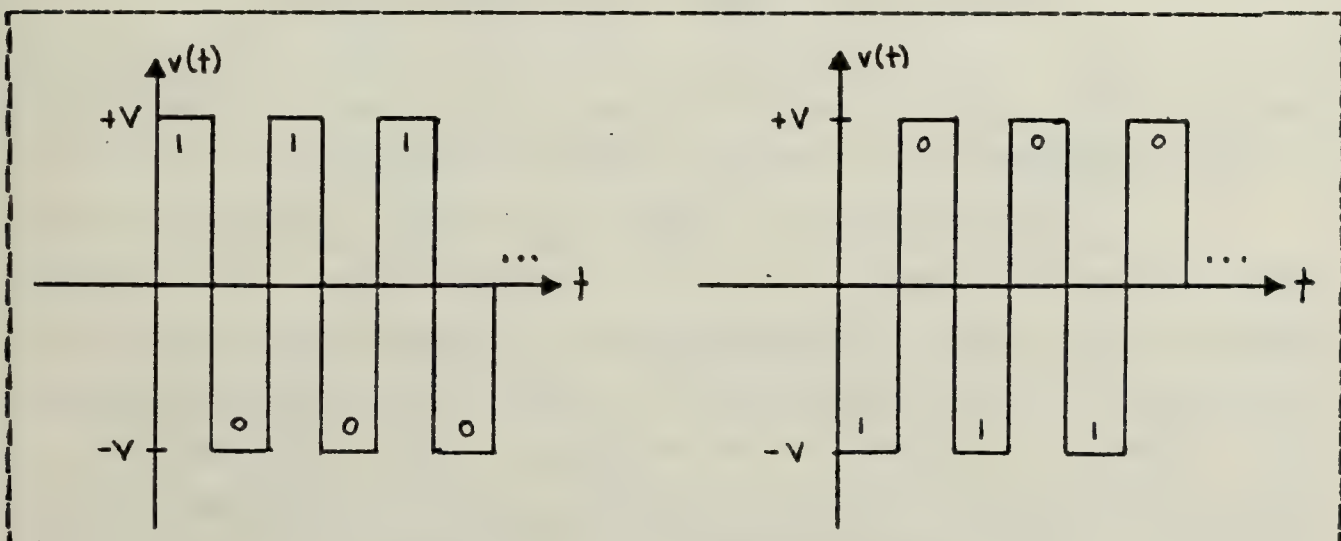


Figure 5.2 Bipolar Logic



The 0 or the 1 of the binary code word can be represented either as  $+V$ ,  $-V$  or  $-V$ ,  $+V$  respectively. This is a matter of convention but must be clearly understood in the various component designs in order to ensure compatibility between parts of the system.

Unipolar logic utilizes a voltage to represent either of the possible binary digits and the absence of voltage to represent the other. Two common conventions of unipolar logic are represented in Figures 5.3 and 5.4.

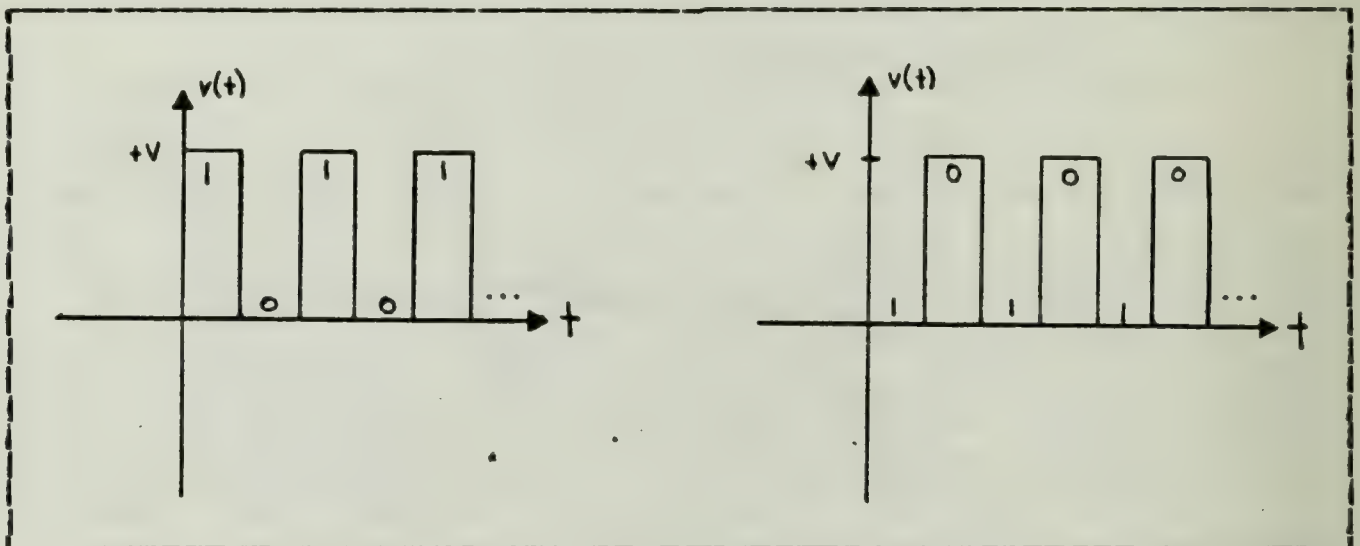


Figure 5.3 Unipolar Positive Logic

Again, which form of unipolar logic might be employed in the A to D converter is a matter of convention.

Variations of these two basic encoding formats have the advantages of ease of generation and improved decoding and clock recoverability. Two variations commonly used in satellite communications are the Non Return to Zero (NRZ) and the Manchester waveforms. The NRZ waveform is simply the voltage representation which corresponds to the stream of bits represented in the logic scheme chosen. There is no transition as long as the same bit is present. Choosing bipolar logic as an example, a NRZ waveform can be described schematically as in Figure 5.5.

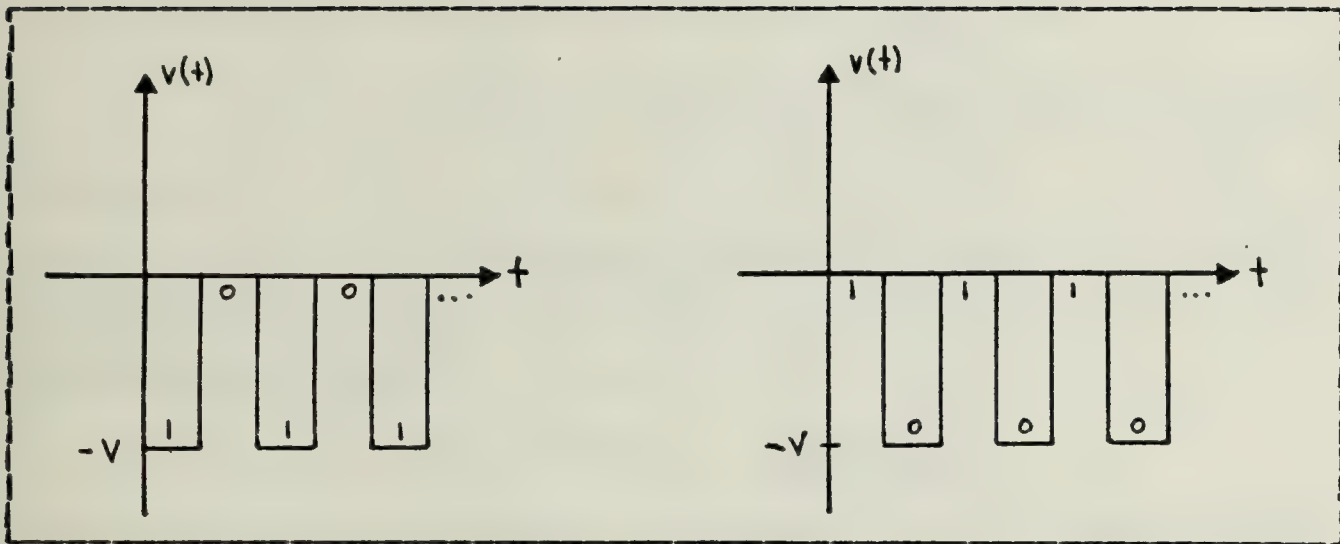


Figure 5.4 Unipolar Negative Logic

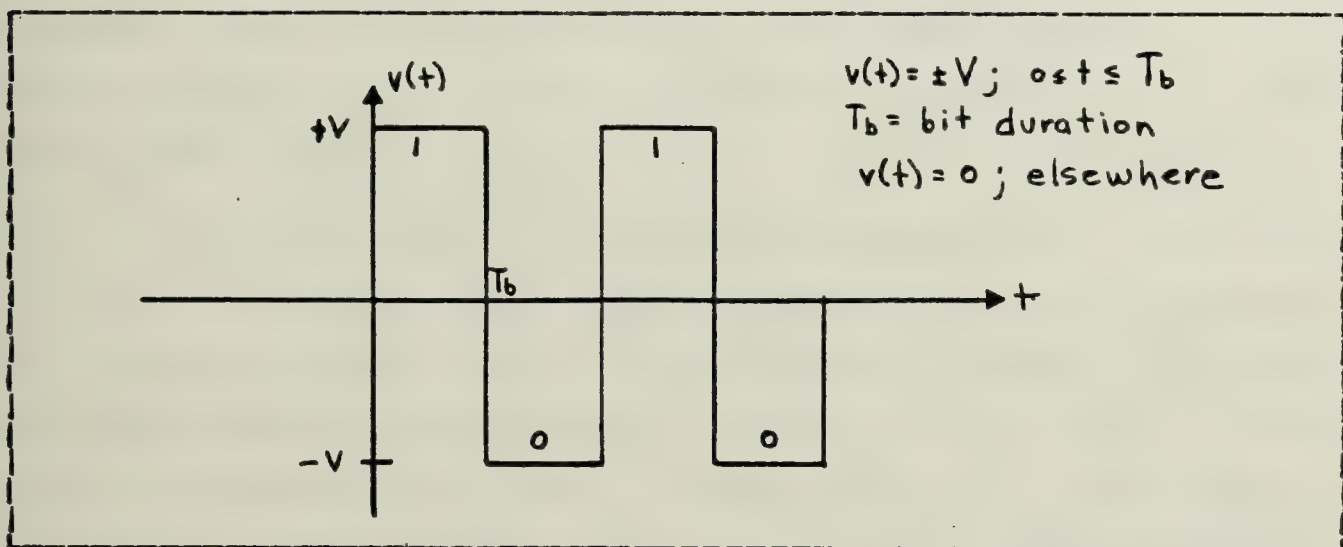


Figure 5.5 Non Return to Zero Waveform

The Manchester waveform of the same source voltage is represented with a transition at the midpoint of the bit duration from either  $+V$  to  $-V$  or  $-V$  to  $+V$ . This form of coding has the advantage of clock synchronization in all cases of digital encoding. Schematically, the Manchester code can be represented as in Figure 5.6.

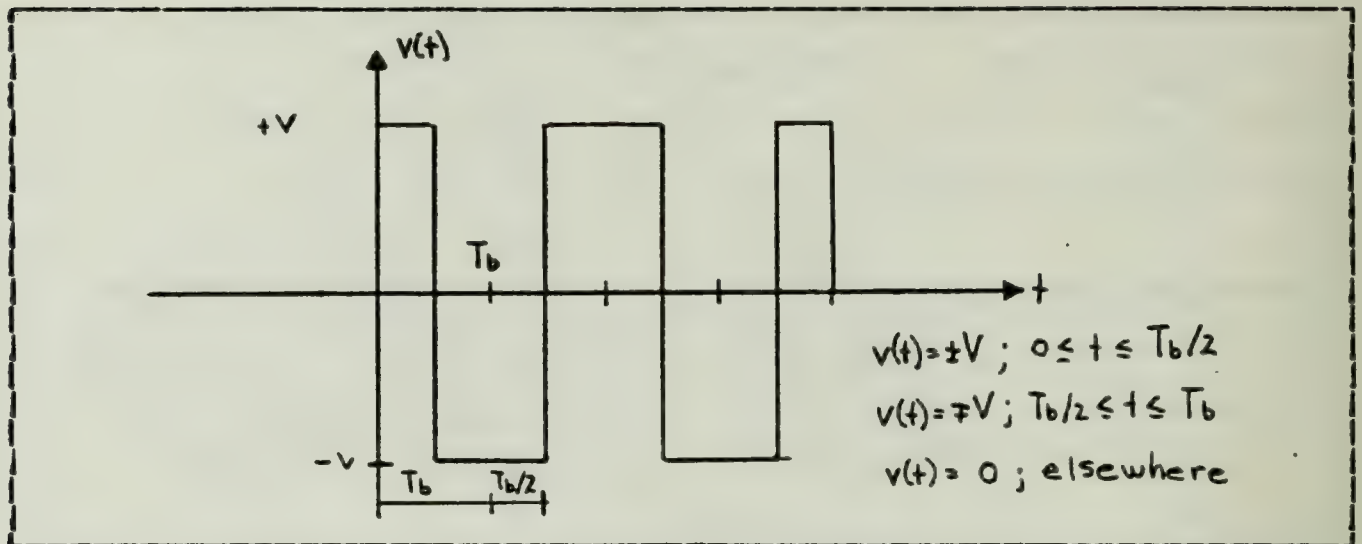


Figure 5.6 Manchester Waveform

### 3. Modulator

The modulator in a digital communications system is more commonly referred to as a modem (modulator/demodulator). This device does the actual transformation of the digital information into a waveform that can be transmitted from one point to another. The modulation as mentioned earlier can modify either the amplitude, phase or frequency of the carrier wave. In simple terms, the digital information derived from the analog source is mapped into the carrier wave. It is this mapping that determines the power and bandwidth characteristics of the modem.

### 4. Multiplexer

Multiplexing is a common method for increasing the utility of a given communications channel. Variations in multiplexing techniques give rise to the multiple access techniques employed in satellite communications.



a. Frequency-Division Multiplexing (FDM)

Frequency-Division Multiplexing is a technique whereby the capacity of a communications channel is increased by adding one signal to another. These signals occupy discrete nonoverlapping frequency bands. A specific signal is recovered by use of a band pass filter for the frequency band desired. [Ref. 7]

b. Time-Division Multiplexing (TDM)

Time-Division Multiplexing is a technique used to increase the capacity of a given communication channel by adding signals in time. That is, specific signals or portions of signals are allocated specific time slots or portions of the carrier wave. These time allocations cannot overlap and the signal is recovered through proper synchronization of the recovery hardware with the multiplexer. [Ref. 7]

c. Code-Division Multiplexing (CDM)

Code-Division Multiplexing is a technique for increasing the capacity of a given communications channel through the assignment of a characteristic code to the digital signal before it is multiplexed in the time or frequency domain. Since the code used in multiplexing is known at the demultiplexer, recovery of the digital source code could be accomplished through the use of a matched filter or correlator. [Ref. 7]

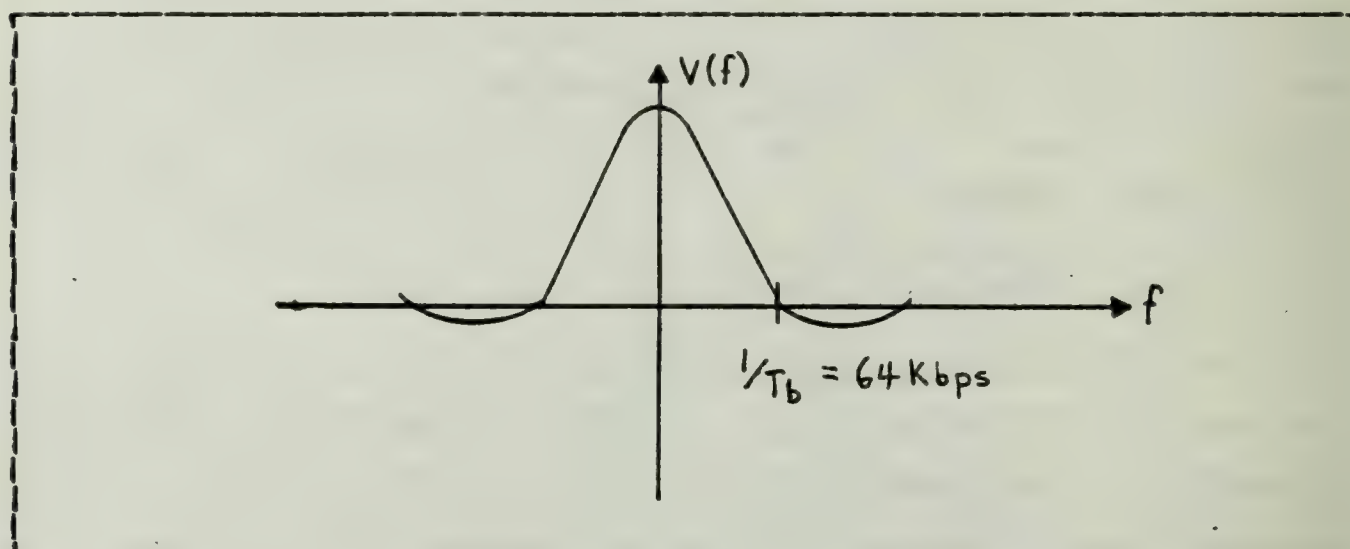
C. BANDWIDTH

For purposes of discussion and for uniformity, when examining the various digital signal modulation techniques of interest, the bandwidth will be defined as the highest significant Fourier frequency component of the signal in

question. For digital communications, it can be shown that by the above definition, the bandwidth of the baseband waveform is  $1/T_b$ , where  $T_b$  is the bit duration. For example:

1. Voice sampled at the rate of 8000 samples/sec
2. Each sample represented by an 8 bit code word
3.  $R_b = \text{bit rate} = (8000 \text{ samples/sec}) (8 \text{ bits/sample})$   
 $R_b = 64 \text{ kbps}$
4.  $T_b = \text{bit duration} = 1/R_b$   
 $T_b = 15.6 \text{ microsec}$

The Fourier transform of the baseband waveform of duration 15.6 microsec is represented in Figure 5.7.



**Figure 5.7** Fourier Transform of Baseband Waveform

It should be readily apparent that as the bit rate of the digital baseband waveform increases, so does the bandwidth. It is desirable to keep the bandwidth of the signal to a minimum due to limitations on availability of the electromagnetic spectrum. In addition, interference with signals which occupy adjacent frequency bands is reduced and propagation limitations of the medium and hardware limitations such as self-inductance and capacitance are minimized.

## D. SPECIFIC TECHNIQUES

Now with some of the basics of digital signals already introduced it is possible to investigate some of the more significant modulation formats employed in digital satellite communications. The organization of this section will be around the three general types of modulation, i.e., phase, amplitude and frequency. Within each broad category, several techniques will be described under the subareas of binary encoding, where the information is modulated onto the carrier bit by bit, and block encoding, where groups of bits do the modulation. Variations of these areas will be pointed out. The characteristics of the modulation technique will be introduced including, where practical, an analytical description, phasor representation, error probability, spectral efficiency, advantages and disadvantages.

### 1. Phase Shift Keying (PSK)

Phase shift keying is a generic form of signal modulation where the baseband waveform is used to modify or change the phase of the carrier. Phase, as stated earlier is one of the characteristics which distinguish one sinusoid from another. In general, phase shift keying offers the advantages of power and bandwidth efficiency.

#### a. Binary Phase Shift Keying (BPSK)

As the name suggests, Binary Phase Shift Keying (BPSK) results in a modulated carrier waveform consisting of two phases of the same sinusoid. The baseband waveform is used to change the phase of the carrier bit by bit from one phase to the other. In the case of bipolar logic, the bit stream consists of  $\pm 1$ . The general waveform of the carrier can be represented as a cosine wave of amplitude  $A$ , angular



frequency  $\omega$  and initial phase angle  $\delta$  as in equation 5.1.

$$v_c(t) = A \cos(\omega t + \delta) \quad (\text{eqn 5.1})$$

Since two phases are represented in  $v_c(t)$ , it is customary to let them differ by  $\pi$  radians. This can be represented by the modification of equation 5.1 to account for a  $\pi$  phase shift depending on whether the bit,  $v(t)$  to be modulated is either a  $+1$  or  $-1$  as shown in equation 5.2.

$$v_c(t) = A \cos(\omega t + \delta + \pi/2(1-v(t))) \quad (\text{eqn 5.2})$$

Equation 5.2 can be expanded to result in a simplified form showing that the modulated carrier waveform is simply a product of the baseband waveform and the carrier. See equation 5.3 where  $v(t)$  represents the baseband waveform.

$$v_c(t) = A v(t) \cos(\omega t + \delta) \quad (\text{eqn 5.3})$$

A phasor diagram can be constructed to represent the modulated carrier waveform in BPSK and is shown in Figure 5.8. As can be seen by the phasor diagram in Figure 5.8, the phase of the modulated waveform depends on the value of the baseband waveform and differs by  $\pi$  depending on whether the value of the modulated bit is  $\pm 1$ .

Remembering the section on the properties of the Fourier transform, it is possible to analyze the spectral and power efficiency of BPSK. Recall BPSK can be created by multiplication. Multiplication of two voltages is covered by the frequency translation property of the Fourier transform. Frequency translation results in double the bandwidth or spectral requirement of the translated waveform and half the power. In the case of a NRZ baseband waveform,

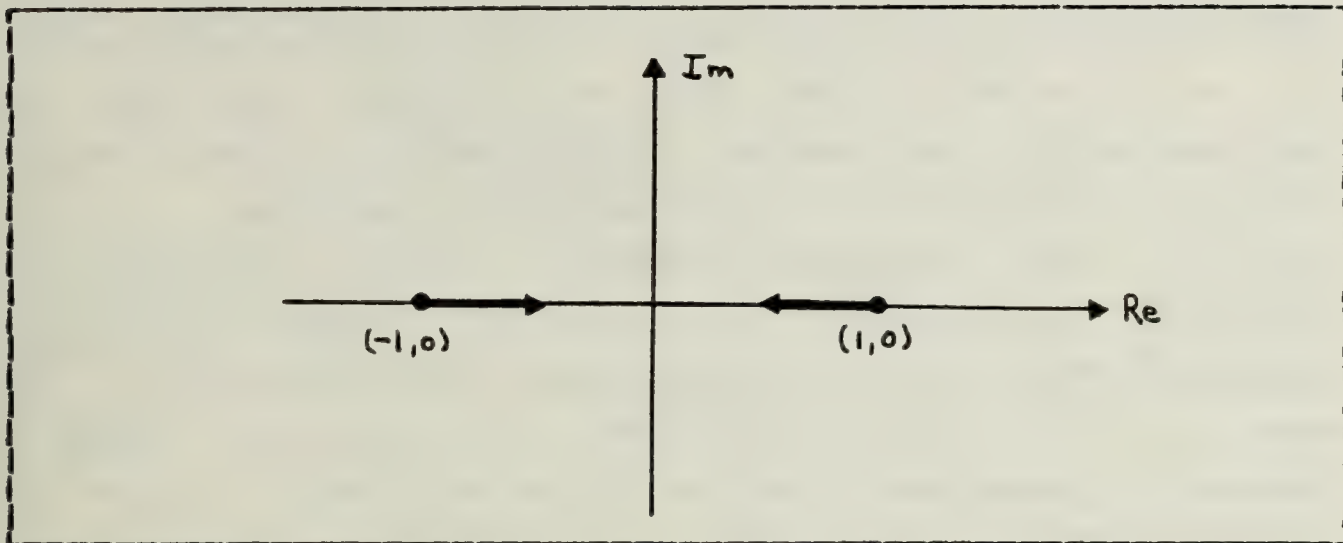


Figure 5.8 Phasor Diagram of BPSK

the bandwidth of the non-translated pulse, determined from the largest significant Fourier frequency component is  $1/T_b$ , where  $T_b$  is the bit duration. The bandwidth of the translated voltage (BPSK carrier) is  $2/T_b$ . The power in a sinusoid is  $A^2/2$  and none of this power capacity is lost in BPSK since only the phase of the original sinusoid is changed. For a Manchester baseband waveform, the non-translated signal has bandwidth  $2/T_b$  and the translated wave has bandwidth  $4/T_b$ .

The advantage of BPSK is that it is relatively efficient in bandwidth utilization and power capacity. It does have the disadvantage of fairly large sidelobe components which contribute to interference with adjacent frequencies. This usually necessitates some type of filtering before the signal can be transmitted.

b. Differential Binary Phase Shift Keying (DBPSK)

The recovery of a Binary Phase Shift Keyed signal requires the use of a phase coherent reference signal as described in the previous sections on autocorrelation and matched filters. This may not always be practical or

possible so a technique called Differential Binary Phase Shift Keying has been developed. DBPSK uses the same carrier type as BPSK and offers many of the same advantages and disadvantages. As stated, the primary difference is the lack of necessity for a phase coherent reference.

The baseband waveform,  $v(t)$ , for DBPSK is generated by comparing the phase of the next bit to be transmitted to that of the previously transmitted bit. For example, if a +1 is encoded onto the carrier as  $v(t)$ , it represents a particular phase of the carrier. If the next bit to be sent is also a +1, there is no phase change, i.e., the bit remains the same over two successive bit durations. If, however, the second bit to be sent is a -1, this represents a  $\pi$  phase change and is decoded as a bit change over successive bit durations. This can be expressed mathematically as in equation 5.4 where  $b_k$  represents the bit which will be encoded on the carrier and  $b_{k-1}$  represents the previous bit already encoded and  $a_k$  represents the present bit.

$$b_k = b_{k-1} \cdot a_k \quad (\text{eqn 5.4})$$

As can be seen when  $b_k = -1$ , there is a phase change and the bit changes from one bit to the next depending on the logic type being employed. If  $b_k = +1$ , this represents the phase remaining constant over two successive bit durations, i.e., no bit change. See Figure 5.9 [Ref. 8]. Recovery is effected by correlation of the received bit to the previously received bit.

Differential Binary Phase Shift Keying (DBPSK) differs from Differentially Encoded Binary Phase Shift Keying (DEBPSK) in the recovery step only. DBPSK takes advantage of the modulation of phase shift in the recovery process while DEBPSK still utilizes a phase coherent carrier



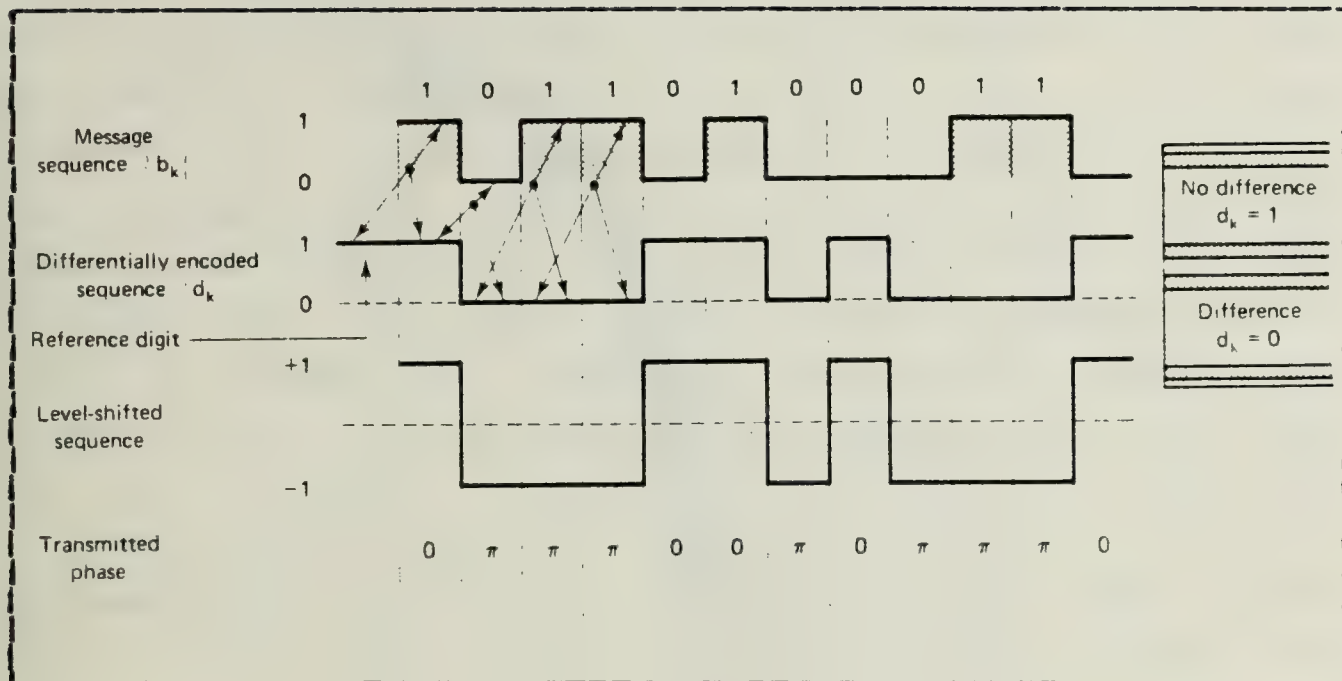


Figure 5.9 Differential Binary Phase Shift Keying

signal in demodulation. Both cases use the same scheme for differential encoding.

### c. Orthogonal Binary Phase Shift Keying (Orthogonal BPSK)

In Binary Phase Shift Keying there is always a probability, with the interjection of noise on the transmitted signal that an individual bit will be decoded incorrectly from that which was actually transmitted. In some environments, this probability is so high that special encoding techniques must be employed in order to reduce the probability that a bit or a binary code word will be decoded in error. One such technique is Orthogonal BPSK. Orthogonal BPSK is a type of group encoding scheme where groups or blocks of bits from the original signal are assigned redundant bits according to a predetermined assignment routine. This new expanded sequence of bits is then transmitted as before in BPSK.

For example, assume that it has been determined that the respective values of the original signal will be represented by a  $k$  bit binary code word. It has been demonstrated that there exists  $2^{**k}$  binary code words,  $k$  bits long in this type of arrangement. With Orthogonal BPSK there also exists  $2^{**k}$  orthogonal sequences of bits (called channel symbols, bauds or chips) [Ref. 6], which represent the  $k$  binary code words. However, the number of bits or chips in the sequence is increased to  $n = 2^{**k}$ , thereby providing the desired redundancy. Each of the  $2^{**k}$  sequences of  $n$  chips is constructed to be orthogonal to every other sequence in order to assure maximum separation. See Figure 5.10 [Ref. 6].

$k = 2$ data word	Orthogonal chip sequence		Transmitted waveforms
			BPSK carriers with phase shifts below:
1 1	1 1 1 1	$C_1(t)$	$\pi \quad \pi \quad \pi \quad \pi$
1 0	1 -1 -1 -1	$C_2(t)$	$\pi \quad \pi \quad -\pi \quad -\pi$
0 1	1 -1 -1 1	$C_3(t)$	$\pi \quad -\pi \quad -\pi \quad \pi$
0 0	1 -1 1 -1	$C_4(t)$	$\pi \quad -\pi \quad \pi \quad -\pi$
			$\leftarrow \tau \rightarrow$ $\leftarrow T_w \rightarrow$

Figure 5.10 Orthogonal Binary Phase Shift Keying

Decoding is also done by blocks through a bank or  $2^{**k}$  correlators. Since the decoding is done in blocks, the probability that the entire transmitted code word will be incorrectly decoded is reduced even though individual bits are incorrectly decoded. This is the major advantage of Orthogonal BPSK. Figure 5.11 [Ref. 6], shows the reduction in error probability with Orthogonal BPSK

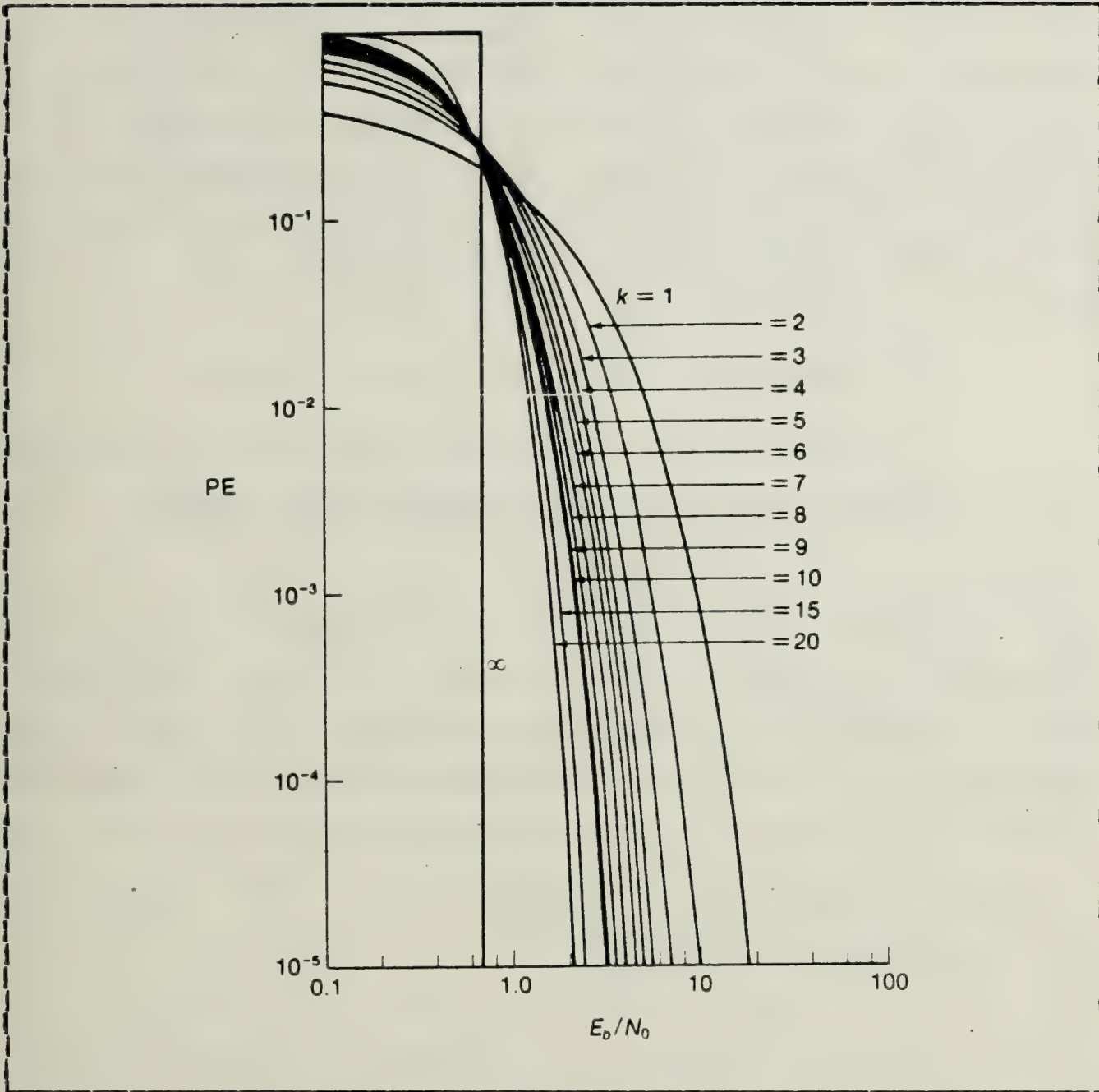


Figure 5.11 Probability of Error for Orthogonal BPSK

The major disadvantage to Orthogonal BPSK is that it results in a reduced data bit rate at a given transmission rate or the necessity for a higher transmission rate if the same bit rate is to be maintained. For example, suppose the bit rate of a PSK signal is  $R$  bits/sec. If each of the  $k$  bits in the PSK signal are represented by  $2 \cdot k$  chips as is the case in Orthogonal BPSK modulation, then the



bit rate must be  $(2^{k/k})R$  if the same data rate is to be maintained. Additionally, since the necessary bit rate increases, the bit duration decreases. It was shown in the section on bandwidth that as bit duration decreases as the result of a higher bit rate, an increase in bandwidth is the result. Therefore, the tradeoff in decreased transmission error comes at the expense of bandwidth and bit rate. [Ref. 6]

d. Quadrature Phase Shift Keying (QPSK)

Quadrature Phase Shift Keying derives its name from a four-phase modulation of the carrier waveform. These modulations are achieved by simultaneously modulating the inputs from two bit streams onto a single carrier. The two bit sequences could be from two separate sources or successive bits from a single source. To take advantage of the orthogonality of the sine and cosine functions, QPSK can be viewed as in equation 5.5 as the sum of a BPSK modulated sine and cosine.

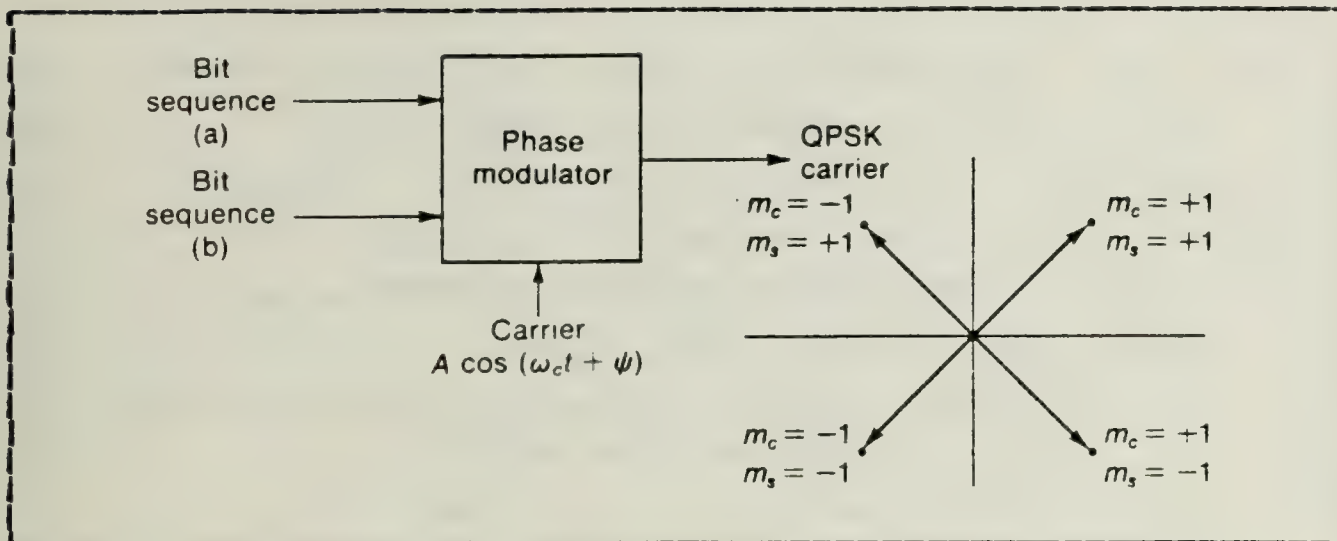
$$v_c(t) = A v_1(t) \cos(\omega t + \delta) + A v_2(t) \sin(\omega t + \delta) \quad (\text{eqn 5.5})$$

The signal can also be represented as in equation 5.6 where the phase angle,  $\theta$ , of the modulated carrier is shown in equation 5.7.

$$v_c(t) = 2^{.5} A \cos(\omega t + \theta + \delta) \quad (\text{eqn 5.6})$$

$$\theta = \arctan (v_1(t)/v_2(t)) \quad (\text{eqn 5.7})$$

The phases possible as a result of substitution of the possible bits in bipolar logic in equation 5.7 are  $\pm 45$  degrees and  $\pm 135$  degrees. See Figure 5.12.



**Figure 5.12 Modulation and Phases of QPSK**

Since two bits are being encoded at one time for transmission, the bit rate in QPSK is twice that of BPSK. The BPSK bit rate is equal to the QPSK symbol rate, (symbol rate being the time the quadrature bits exist on the carrier). It is the symbol rate in QPSK which determines the bandwidth and since symbol rate is equal to bit rate in BPSK, the bandwidth of the two modulation techniques is identical. The great advantage of QPSK lies in the bandwidth utilization. Two bits of information are transmitted thereby effectively doubling the bit rate at no additional cost in bandwidth. A phasor representation of QPSK is illustrated in Figure 5.13. Each quadrature component of the modulated signal contains half the power of the total carrier or  $(A^2/2)/2 = A^2/4$ .

**e. Offset Quadrature Phase Shift Keying (OQPSK)**

In Quadrature Phase Shift Keying, both bits of the two baseband waveforms, which are to be modulated onto the carrier, change at the beginning of the respective symbol time. This allows the phase of the carrier to change up to 180 degrees for each symbol as illustrated in the

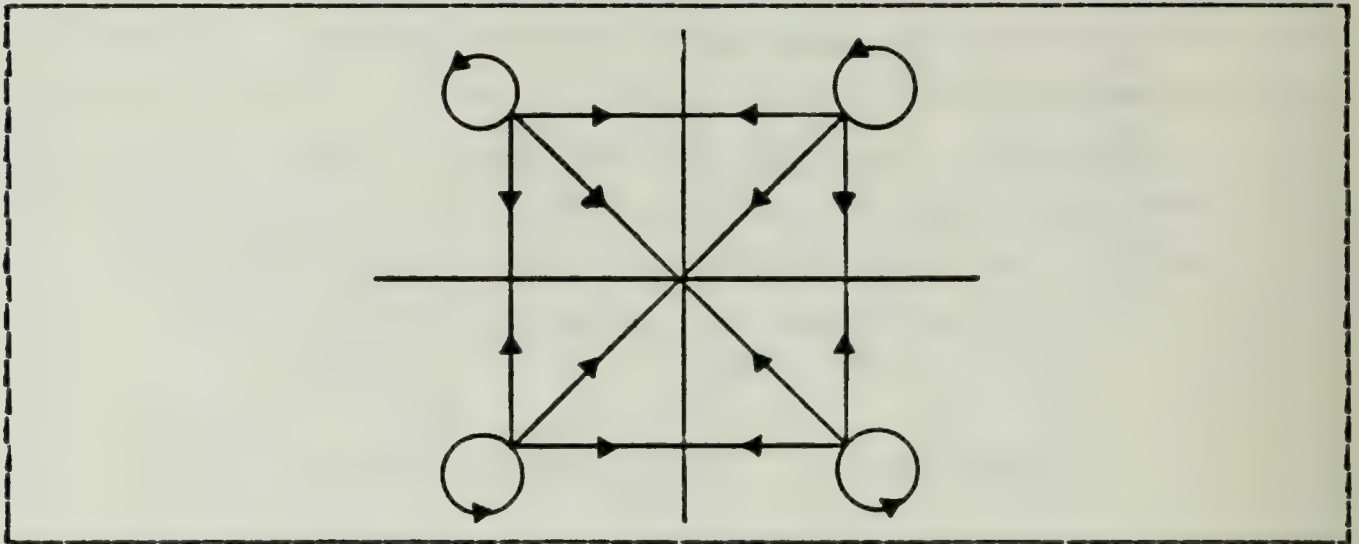


Figure 5.13 Phasor Diagram of QPSK

phasor diagram in Figure 5.13. The phase change of the modulated signal can be limited to 90 degrees through a modulation technique called Offset Quadrature Phase Shift Keying. This is accomplished by delaying the application of the second baseband channel for  $T_s/2$  seconds, where  $T_s$  is symbol duration as illustrated in Figure 5.14. Modulation is then accomplished as before in QPSK.

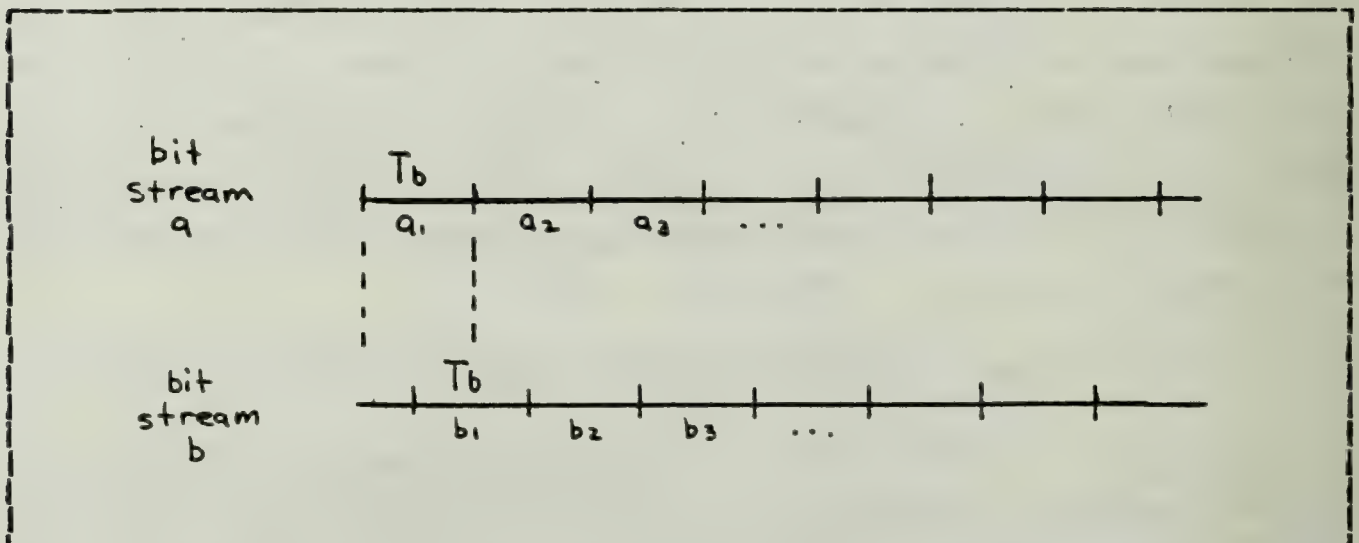


Figure 5.14 Offset Quadrature Phase Shift Keying



The result is a QPSK modulated signal in which the maximum phase shift between successive bits is 90 degrees. The bandwidth and power spectra of an OQPSK modulated signal are the same as that of a QPSK modulated signal. The advantage of OQPSK is in the limit which is placed on phase shift during encoding which simplifies the encoding hardware. Additionally OQPSK has spectral and interference advantages during decoding [Ref. 6].

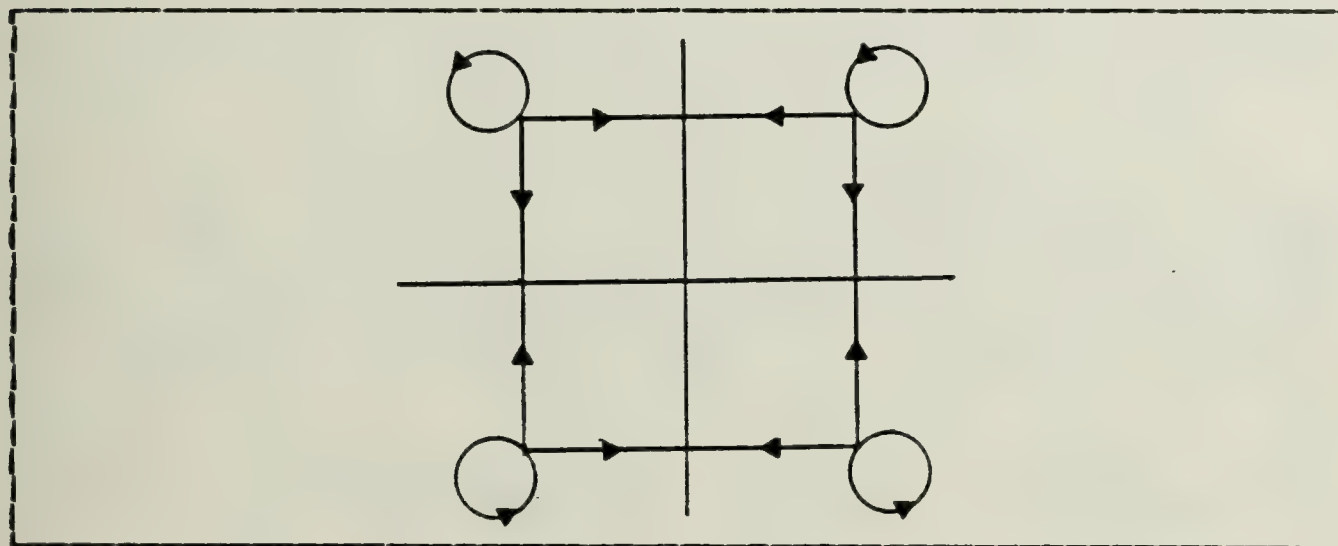


Figure 5.15 Phasor Diagram of OQPSK

f. Multiple Phase Shift Keying (MPSK)

Another form of digital modulation is known as Multiple or M-ary Phase Shift Keying (MPSK). Once again, as in Orthogonal Binary Phase Shift Keying,  $k$  data bits are transmitted. The  $k$  bits represent the word length and there are  $m = 2^k$  different binary words,  $k$  bits long, for this type of modulation. In MPSK, the binary code word is used to vary the phase of the carrier. There are therefore  $m = 2^k$  different phases in MPSK. These sequences of binary digits need not be orthogonal as in Orthogonal Binary Phase

Shift Keying. The length of the transmitted symbol is only  $k$  bits long as compared to  $2^k$  bits for Orthogonal BPSK. The phases therefore are not all orthogonal leading to greater difficulty in decoding and a higher probability that an individual code word will be incorrectly decoded, especially as  $m$  becomes large.

As in other types of Phase Shift Keying, MPSK can be represented as in equation 5.8 where  $\theta = \text{phase} = (2\pi/m) i; i=0, 1, \dots, m-1$ .

$$v_c(t) = A \cos(\omega t + \theta + \delta) \quad (\text{eqn 5.8})$$

An advantage of MPSK is that as long as the symbol transmission rate does not increase, the carrier bandwidth does not increase even as the size of the binary code word increases. The disadvantage, as stated earlier, has to do with error rates in decoding. Decoding requires a phase coherent reference of considerable stability especially as  $m$  increases and respective phases become closer together. Practical limits for this type of modulation is  $m = 8$  or a 3 bit binary code word due to the complexity of the encoding and decoding hardware.

## 2. Amplitude Shift Keying (ASK)

Amplitude Shift Keying, as the name implies is modulation of the carrier through variation of its amplitude due to variations in the baseband waveform. The information in the baseband waveform is thereby imparted to the carrier through this modulation. Since the amplitude of the carrier varies between successive symbols or binary code words, the power also varies. Recovery is by comparison of the transmitted power during each symbol to the possible  $m = 2^k$  power levels.

a. Multiple Amplitude Shift Keying (MASK)

MASK is a type of group encoding where  $k$  bits are combined into a single waveform for transmission and subsequent recovery. Since each symbol or binary code word contains  $k$  bits, there are  $m = 2^{**k}$  different symbols and consequently  $m$  different amplitudes of the carrier possible during the symbol duration time. An assignment scheme which maps the binary code words into the  $m$  different amplitudes would be to simply space them  $A_i = A/m_i$  volts apart, where  $A$  is the maximum amplitude and  $m_i$  represents the decimal equivalent of the binary code word from  $i = 1$  to  $2^{**k}$ .

The analytic form of a MASK waveform is represented in equation 5.9 where all the variables are the same as presented earlier except for  $A_i$  which is equal to one of  $i$  different amplitudes depending on the bit sequence to be transmitted.

$$v_c(t) = A_i \cos(\omega t + \delta) \quad (\text{eqn 5.9})$$

MASK is not very popular in satellite communications since it depends on a carrier of very stable amplitude. This is not generally practical under actual conditions. However, MASK does have the spectral advantage of a constant bandwidth of  $2/T_s$ ,  $T_s$  equal to symbol duration, even as the binary code word increases in length.

b. Quadrature Amplitude Shift Keying (QASK)

Quadrature Amplitude Shift Keying (QASK) is a hybrid digital signal modulation technique. It represents separate amplitude modulation of the quadrature components of a common carrier. In that sense it is both ASK and PSK. Implementation of QASK involves simultaneous application of  $M$ -ary Amplitude Shift Keying to the quadratures. See



equation 5.10, where A1 and A2 are derived according to an assignment scheme as in MASK.

$$vc(t) = A1 \cos(\omega t + \delta) + A2 \sin(\omega t + \delta) \quad (\text{eqn 5.10})$$

By this method, the effective bit rate is doubled over that of ordinary MASK. In other words, QASK results in the same data rate as modulating 2k bits onto the carrier at the same time or during the same symbol duration with MASK. The advantage to QASK lies in the fact that this increase in bit rate comes at no further increase in bandwidth. The disadvantages associated with MASK are still present with an additional increase in the complexity of the decoding equipment. Decoding of QASK must be done in two steps. First the signal is recovered in phase through a phase-coherent correlator and then each amplitude modulated signal is recovered as before in MASK by comparison of the power levels of the received signal.

### 3. Frequency Shift Keying (FSK)

Frequency Shift Keying (FSK), is the generic term used to describe a number of digital signal modulation techniques which cause variations in the carrier frequency by interaction with the baseband waveform. Decoding is generally through measurement of the frequency of the received signal.

#### a. Minimum Shift Keying or Fast Frequency Shift Keying

Power requirement for the transmission or retransmission of a satellite signal is extremely important. It is desirable to limit power required to accurately

transmit a piece of information to the minimum possible consistent with allowable error rates. For this reason, detection of FSK signals is limited to coherent methods as power required increases significantly for non-coherent methods.

MSK or FFSK are identical and represent a modulation technique known as continuous phase FSK. They get their names from the fact that "fast" indicates more bits per second can be transmitted in a given bandwidth than ordinary BPSK and "minimum" refers to the minimum modulation index for which orthogonal signalling occurs [Ref. 7].

Analytically, FFSK can be expressed as in equation 5.11 and equation 5.12.

$$v_c(t) = A \cos((\omega_c + \Delta\omega)t) \quad (\text{eqn 5.11})$$

$$v_c(t) = A \cos(\pm\Delta\omega t) \cos(\omega_c t) - A \sin(\pm\Delta\omega t) \sin(\omega_c t) \quad (\text{eqn 5.12})$$

As is noted, FFSK can be envisioned as the summation of separately modulated in-phase and quadrature components of the carrier by the baseband waveform. The technique is similar to that used in OQPSK. The frequency variation in the carrier waveform is between  $\omega_c + \Delta\omega$  and  $\omega_c - \Delta\omega$ . Maximum separation in phase of these two frequency components occurs at  $\pi$  as noted in equation 5.13, where  $T_b$  represents the bit or symbol duration.

$$(\omega_2 - \omega_1)T_b = \pi \quad (\text{eqn 5.13})$$

This can be converted to the modulation index mentioned above simply by converting to frequency as in equation 5.14.

$$h = (f_2 - f_1)T_b = .5$$

(eqn 5.14)

For FFSK, the frequency deviation or separation between frequencies of the transmitted carrier is exactly  $1/2T_b$  and the deviation from the carrier frequency is  $1/4T_b$ . FFSK can therefore be rewritten from equation 5.12 as in equation 5.15 by substitution for  $\Delta\omega = 2\pi(1/4T_b)$ .

$$v_c(t) = A \cos(\pm\pi t/2T_b) \cos(2\pi f_c t) - A \sin(\pm\pi t/2T_b) \sin(2\pi f_c t) \quad \text{(eqn 5.15)}$$

The error rate in this modulation technique is identical to that of BPSK. Frequency Shift Keying can be implemented to incorporate any given modulation index by simple calculation of the frequency separation with equation 5.14 and substitution into equation 5.15. For M-ary Frequency Shift Keying, the separation between frequencies must be at least  $1/T_s$ , where  $T_s$  is the symbol duration, in order to avoid carrier energy from one frequency being incorrectly interpreted as carrier energy from another frequency. This results in a modulation index of 1. Figure 5.16 represents the phasor diagram of an FFSK modulated signal.

#### 4. Quadrature Partial Response Signalling (QPRS)

Quadrature Partial Response Signalling (QPRS) represents a specific type of the general class of digital signal modulation techniques called Partial Response Signalling. Partial Response Signalling is a method in which a controlled amount of intersymbol interference (ISI) is allowed during the encoding process. Previously any overlap in successive signals resulted in aliasing and an increased probability of error in decoding. The idea was



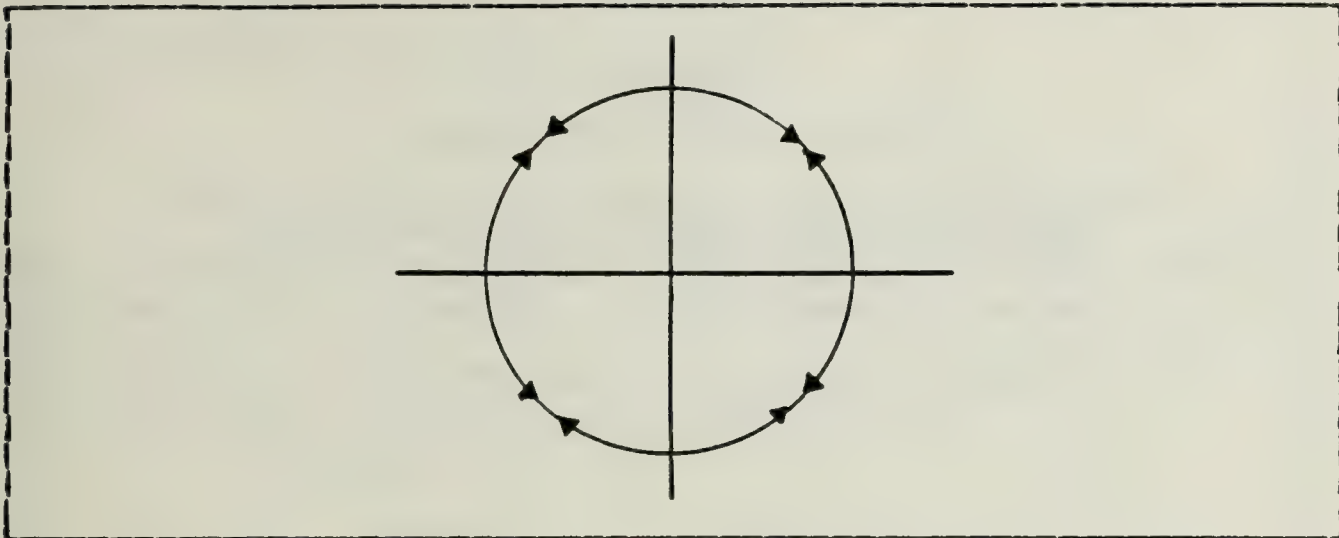


Figure 5.16 Phasor Diagram of FFSK or MSK

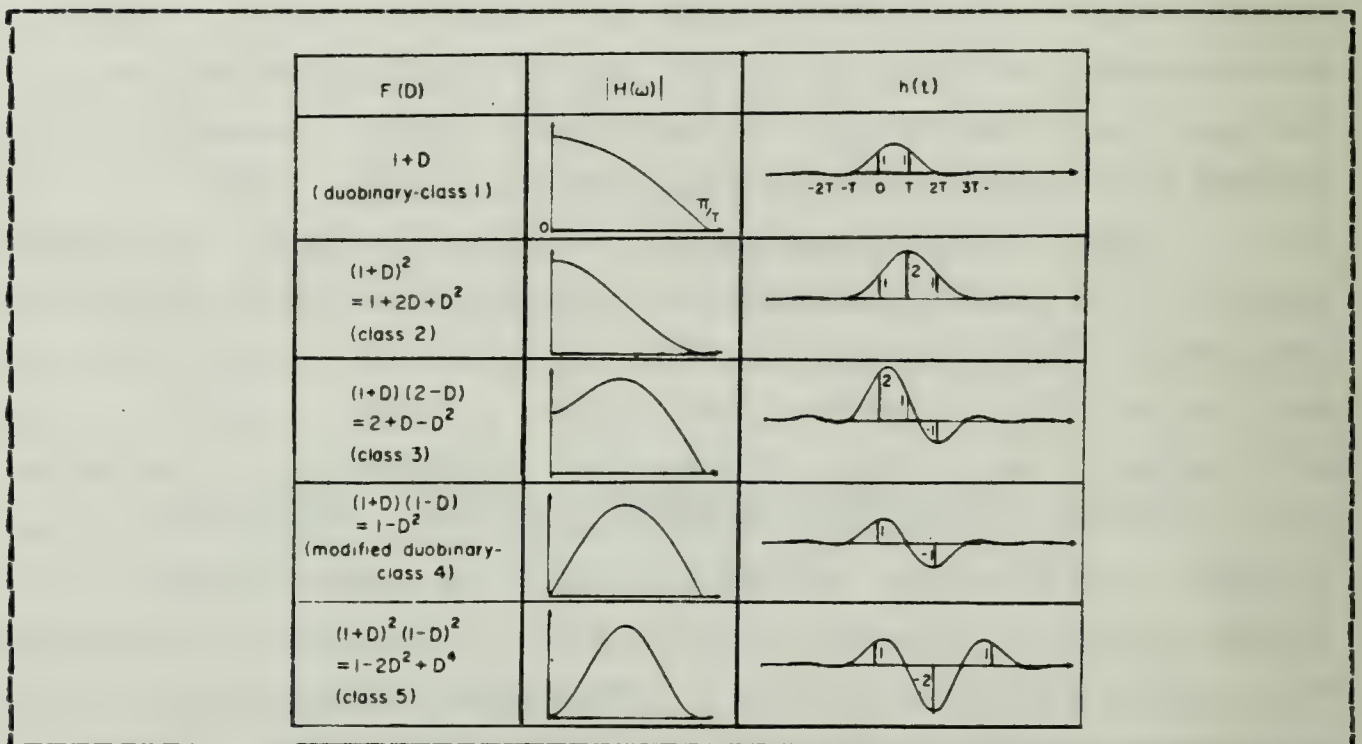
first introduced by Lender in 1963 [Ref. 12]. His concept was that in knowing and controlling the amount of interference allowed during the encoding process, compensations can then be made for the ISI at the receiver. Allowing for a limited amount of ISI makes it possible to transmit at rates equal to the Nyquist rate, something that is not possible with many other systems [Ref. 13].

QPRS is implemented, as with previous quadrature systems, through simultaneous modulation of the quadrature components of a common carrier. However, this time the modulation is accomplished with a PRS system. The modulation of the respective quadrature components represents the impulse response of one of a number of linear filters to the bits of the baseband waveform. The linear filter employed depends on the class of the Partial Response Signalling system being used. There are 5 classes of linear filters commonly used in PRS and they are illustrated analytically in Table 1 and graphically in Figure 5.17.

Output of a QPRS system can be represented as in equation 5.16, where  $a_n$  and  $b_n$  represent the  $n$ th bit to be modulated and  $h(t - nTs)$  represents the contribution of the

**TABLE 1**  
**QPRS SYSTEM POLYNOMIALS**

SYSTEM POLYNOMIAL $F(D)$	FREQUENCY RESPONSE $H(\omega)$ for $ \omega  \leq \pi/T$	IMPULSE RESPONSE $h(t)$
$1 + D$	$2T \cos \frac{\omega T}{2}$	$\frac{4T}{\pi} \frac{\cos(\pi t/T)}{T^2 - 4t^2}$
$1 + 2D + D^2$	$4T \cos^2 \frac{\omega T}{2}$	$\frac{2T^3}{\pi} \frac{\sin(\pi t/T)}{T^2 - t^2}$
$2 + D - D^2$	$T + T \cos \omega T + j3T \sin \omega T$	$\frac{T^2}{\pi} \sin(\pi t/T) \left( \frac{3t-T}{t^2-T^2} \right)$
$1 - D^2$	$j3T \sin \omega T$	$\frac{2T^2}{\pi} \frac{\sin(\pi t/T)}{t^2 - T^2}$
$1 - 2D^2 + D^4$	$-4T \sin^2 \omega T$	$\frac{8T^3}{\pi} \frac{\sin(\pi t/T)}{t^2 - 4T^2}$



**Figure 5.17 QPRS Linear Filter Impulse Responses**

impulse response of the linear filter at time  $t$  for bit  $n$ .  
[Ref. 14]

$$v_c(t) = \sum_{n=-\infty}^{\infty} a_n h(t - n T_s) \cos(\omega t + \delta) + \sum_{n=-\infty}^{\infty} b_n h(t - n T_s) \sin(\omega t + \delta) \quad (\text{eqn 5.16})$$

QPRS has the advantage of rapid transmission rates with no increase in bandwidth and excellent error handling performance. The cost of such performance improvement comes in the form of a higher required signal to noise (SNR) ratio when compared to other binary systems.



## VI. COMPUTER SIMULATION OF DIGITAL SIGNAL MODULATION TECHNIQUES

Due to the nature of digital signal modulation, it is possible to simulate systems through the use of the digital computer. This chapter is a description of the computer simulation included in the Appendix of this thesis.

### **A. GENERAL DESCRIPTION OF THE PROGRAM**

The program was constructed using top-down design and modular programming. FORTRAN was selected as the programming language due to its capabilities in the area of numerical operations and the availability of mathematical routines to be used as modules in the program. Testing was accomplished on each module as the program was developed. The program was designed to be used interactively in order to allow for instructional use. Although the program was made user friendly in that error checking is accomplished on user inputs, care must be taken to insure instructions are followed correctly.

The program consists of twenty modules, sixteen of which were written by the author, three which were taken from the Double Precision International Mathematics and Statistics Library (IMSLDP) and one which was taken from a NON-IMSL library which resides on the IBM 3033 located at the Naval Postgraduate School. The IBM 3033 was used exclusively for development and testing.

The development was accomplished using the WATFIV compiler; however, the program was written to operate with the VS FORTRAN compiler as well. Testing and operation was done using the VS FORTRAN compiler. Figure 6.1 is a block

diagram of the relationship of the modules for a representative modulation subroutine.

## B. MAIN CONTROL MODULE

MAIN operates as the control module for the entire computer simulation and accomplishes limited output. It introduces the program and allows the user to input various parameters of the digital signal modulation technique to be simulated and various other control functions. These inputs include:

1. Modulation technique
2. Class of QPRS system (when appropriate)
3. Digital logic scheme
4. Baud or symbol rate
5. Bits/binary code word (when appropriate)
6. Carrier frequency
7. Number of samples to be generated
8. Carrier max amplitude
9. Initial phase angle
10. Use of random number generator or no
11. Seed for RNG (when appropriate)
12. Number of repetitions of the simulation

Once the user has input the desired characteristics of the modulation to be simulated, MAIN calls the appropriate module to begin the actual calculation. The values necessary to calculate the time series of the signal are passed to the subroutines as parameters. MAIN calls the required subroutine the number of times the user specifies as repetitions. Each call to a subroutine produces the Discrete Fourier Transform and amplitude spectrum of the signal. It is these respective amplitude spectrums which produce the statistics in the final output after MAIN calls the statistics subroutine the final time.

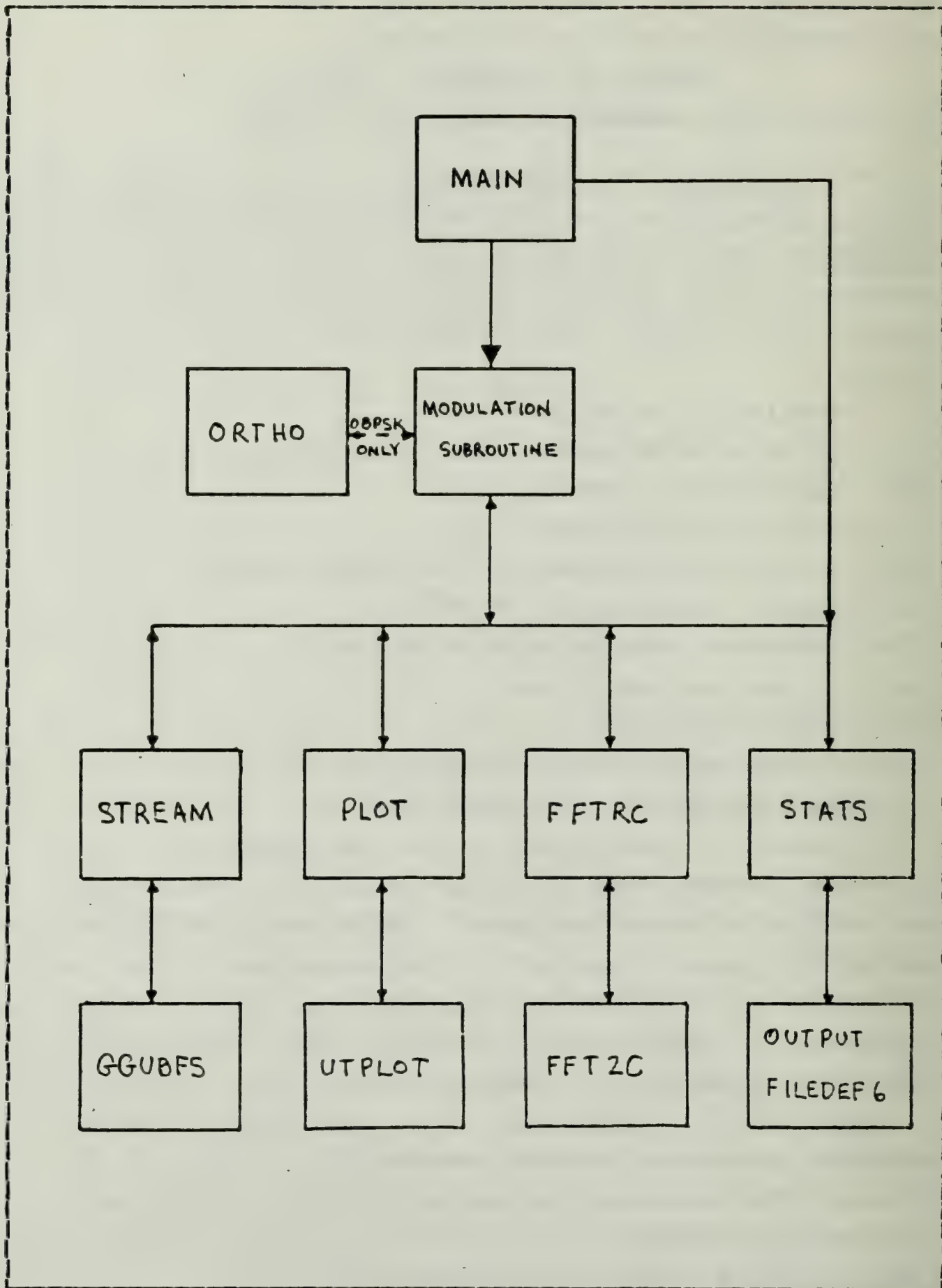


Figure 6.1 Representative Block Diagram



### C. SUBROUTINE BPSK

This description of SUBROUTINE BPSK, a module that simulates Binary Phase Shift Keying, will serve as a guide on the construction and operation of subsequent modules involved in the actual signal modulation. A complete description including its interaction with other program subroutines will be included. Since other signal modulation subroutines interact in the same manner within the program, subsequent descriptions will concentrate on differences to this basic module.

SUBROUTINE BPSK begins with variable declarations as with all FORTRAN programs. All variables are declared to be either type integer or double precision. All calculations in the program are carried out in double precision accuracy.

Variable initialization is next. Significant variable initialized at this point are TIME, the time of the first sample which is set to 0. STEP, the delta time between samples is set to the normalized Nyquist sampling rate. OMEGA, the angular frequency and DELTA, the initial phase angle are computed in radians. NBAUD, a variable which keeps track of the number of symbols which have been modulated, is set to 1 or the first symbol is being modulated. NBAUD remains at its present value until total elapsed time exceed 1 symbol duration or BAUDD, the normalized baud rate.

Subsequently, an initial call is made to SUBROUTINE STREAM to receive the value of the first bit to be modulated. The modulation is then carried out according to equation 6.1.

The values of each sample are assigned to an array for storage along with the corresponding time increment. Modulation continues with the first drawn bit until 1 bit or symbol duration is exceeded when STREAM is called to draw

$$v_c(t) = A v(t) \cos(\omega t + \delta)$$

(eqn 6.1)

A = amplitude

v(t) = bit to be modulated

w = angular frequency

t = time

delta = initial phase angle

another bit. This entire process is repeated until the number of samples specified by the user in MAIN is reached.

At this point, only on the first repetition of the simulation, SUBROUTINE PLOT is called which gives the user the opportunity to plot the time series of the simulated signal. Upon return from PLOT the subroutine calls the IMSL routine FFTRC to generate the Discrete Fourier Transform of the time series produced. This is accomplished on each repetition of the program.

Again on the first repetition only, information on the number of the principle harmonic of the FFT is displayed. Plot is called again to give the user the opportunity to plot the amplitude spectrum. The values of the FFT and the amplitude spectrum are computed on each successive repetition of the program and the amplitude spectrum is added to the statistics being accumulated by a call to SUBROUTINE STATS.

#### D. SUBROUTINE DBPSK

SUBROUTINE DBPSK simulated Differential Binary Phase Shift Keying and is essentially the same as SUBROUTINE BPSK. The difference lies in the fact that what is encoded during the modulation process is the product of the random bit drawn and the value of the bit previously modulated. A

reference bit equal to +1 is used to determine the first bit to be modulated. In this manner, modulation of +1 means no change occurs between successive bits and the phase of the carrier remains the same. Modulation of a -1 indicates a change of bits has occurred and is indicated by a change of phase of the carrier between successive bit durations. The rest of the module remains unchanged.

#### E. SUBROUTINE OBPSK

Orthogonal Binary Phase Shift Keying is accomplished in this module. Again SUBROUTINE OBPSK is constructed essentially the same as SUBROUTINE BPSK. In Orthogonal BPSK the only bit streams that are modulated are  $n$  bits long representing each binary code word, where  $n = 2^{**k}$  ( $k =$  number of bits per word is limited to 6 bits in MAIN). Each sequence of  $n$  bits is orthogonal to every other allowed sequence.

The way this is accomplished in this module is first to draw a series of  $k$  random bits from SUBROUTINE STREAM. This series of  $k$  1's and 0's is then changed to its decimal equivalent from 0 to  $2^{**k}-1$  and saved for later use.

The program continues with a call to SUBROUTINE ORTHO which generates an  $n \times n$  orthogonal matrix of +1's and -1's through the use of the Hadamard matrix and the Kronecker product [Ref. 15]. The Hadamard matrix is a  $2 \times 2$  orthogonal matrix of 1's and -1's. The Kronecker product is the matrix which is formed when a matrix is expanded to twice its original size by replacing each element of the Hadamard matrix by the product of the Hadamard matrix and the original matrix. The value of the decimal equivalent to the  $k$  random bits is used to identify the row of the matrix and the respective columns represent the value of the binary digit which is modulated onto the carrier by BPSK. This



sequence continues until either one bit duration is exceeded, when another orthogonal bit is received from ORTHO or when all the orthogonal bits in a row are modulated. Then another sequence of  $k$  random bits is converted to its decimal equivalent and a new row and column of the orthogonal matrix is identified. Execution continues in this manner until all required samples have been produced. It should be noted that the bit rate for Orthogonal BPSK is  $n$  times the baud rate. This significantly increases the signal bandwidth requirement and decreases the time increment between successive samples in the simulation. The rest of the module is the same as those previously presented.

#### F. SUBROUTINE QPSK

Quadrature Phase Shift Keying, simulated in SUBROUTINE QPSK, is accomplished through simultaneous BPSK modulation of the quadrature components of the carrier. Two random bits are drawn from successive calls of SUBROUTINE STREAM and each bit is in turn modulated onto the carrier components and the sum formed. This process is repeated at time intervals equal to the normalized Nyquist sampling rate until TIME exceeds one symbol duration, BAUDD. At this time two new random bits are drawn and the modulation continues until all required samples are produced. The rest of the modules is the same as those before. Equation 6.2 represents the analytical expression used to simulate

#### G. SUBROUTINE OQPSK

SUBROUTINE OQPSK is essentially the same as SUBROUTINE QPSK with these minor deviations. The time of the first sample is artificially initialized as  $TIME = .5(BAUDD)$  or  $1/2$  the symbol duration. In this manner, the two random

$$v_c(t) = A v_1(t) \cos(\omega t + \delta) + A v_2(t) \sin(\omega t + \delta) \quad (\text{eqn 6.2})$$

$v_1(t)$  = in-phase bit

$v_2(t)$  = quadrature bit

bits first being modulated are both known to have existed on the carrier for 1/2 the symbol duration. The first of these two bits only remains on the carrier until  $\text{TIME} = 1(\text{BAUDD})$  when a third random bit is drawn and modulation begins with this bit. The second random bit is allowed to remain until  $\text{TIME} = 1.5(\text{BAUDD})$  when a fourth random bit is drawn to replace it and so on. In this manner the two bits being modulated are offset in the time they change by 1/2 the symbol duration. This necessitates keeping track of the number of bits modulated on both the in-phase and quadrature component of the carrier. The remaining portion of the module is the same as SUBROUTINE QPSK.

#### H. SUBROUTINE MPSK

This module simulates M-ary Phase Shift Keying. This is accomplished by changing the phase of the carrier to any one of  $n = 2^k$  phases where  $k$  is the number of bits in the binary code word. The max length of the binary code word is limited to 10 in the MAIN program. This modulation is accomplished by use of equation 6.3 to simulate the MPSK system.

The program first draws  $k$  random bits and converts them to the decimal equivalent. At this point modulation begins at  $\text{TIME} = 0$  according to equation 6.3. Modulation continues until one symbol duration is exceeded when a new set of  $k$  random bits is drawn, conversion to decimal takes place and

$$vc(t) = A \cos[wt + (2\pi m)/n + \delta] \quad (\text{eqn 6.3})$$

$m$  = decimal equivalent of binary code word

$n = 2^{**k}$ ;  $k$  = bits in binary code word

modulation continues or until all desired samples have been produced. The program continues as in previous subroutines.

### I. SUBROUTINE MASK

The principles of operation of SUBROUTINE MASK is similar to SUBROUTINE MPSK which also involves a group encoding system. SUBROUTINE MASK allows for M-ary Amplitude Shift Keying. The analytic expression used in the simulation is provided in equation 6.4  $A_i$  is one of  $2^{**k}$  equally spaced amplitudes.

$$vc(t) = A_i \cos(wt + \delta) \quad (\text{eqn 6.4})$$

Once again a set of  $k$  random bits is drawn from SUBROUTINE STREAM. The decimal equivalent of the  $k$  bits is computed and used to adjust the amplitude  $A_i$  according to the assignment routine expressed in equation 6.5.

$$A_i = A/m \quad (\text{eqn 6.5})$$

$A$  = max amplitude

$m = 1$  to  $n$ ;  $n = 2^{**k}$

This process is repeated at each symbol duration until all desired samples have been produced. The rest of the module remains unchanged.



## J. SUBROUTINE QASK

SUBROUTINE QASK combines the elements of the previously described amplitude modulation technique with the now familiar quadrature modulation technique. Two separate streams of random numbers are generated and converted to their decimal form. These numbers are used to produce two separate amplitudes with the same assignment scheme used in MASK. These amplitudes modulate the quadrature components of a common carrier and the output is formed by the sum of the quadratures. Modulation continues in the above manner until a symbol duration elapses when two new amplitudes are calculated. Processing terminates when all required samples are computed.

## K. SUBROUTINE MSK

Minimum Shift Keying is frequency shift keying in which the modulation index is  $1/2$ . The modulation index is represented in equation 6.6

$$h = .5 = (\text{delta}f) T_b \quad (\text{eqn } 6.6)$$

$\text{delta}f$  = frequency separation

$T_b$  = bit duration

The frequency deviation from the carrier is  $.5(\text{delta}f) = 1/4T_b$ . Converting this frequency deviation to radians yields  $\pi/2T_b$ . This angular frequency deviation from the central carrier is either  $\pm\pi/1T_b$  depending on the value of the bit to be modulated.

The simulation in SUBROUTINE MSK is accomplished by drawing a random bit and generating the signal according to equation 6.7.

$$vc(t) = A \cos[(\omega \pm \pi/2T_b)t + \delta] \quad (\text{eqn 6.7})$$

Once again modulation continues until one bit duration is elapsed when another bit is drawn. Modulation stops when all required samples are produced. The rest of the subroutine remains unchanged.

#### L. SUBROUTINE MFSK

This subroutine modulates a signal simulating M-ary Frequency Shift Keying. The separation between adjacent frequencies is established as  $1/T_s$ , where  $T_s$  is the symbol duration. This amounts to a modulation index equal to 1. Initially the entire range of frequency deviation from the carrier is calculated as  $(n-1)T_s$ . The mean frequency deviation is then calculated. In other words the range of frequencies of the modulated signal is the frequency of the central carrier  $\pm 1/2$  the magnitude of the entire range of frequency deviation. The minimum frequency is used as the base for computing the frequency to be used to modulate the signal during each respective symbol duration.

At this point in the subroutine a series of  $k$  random bits ( $k \leq 10$ ) is drawn and converted to decimal. This decimal number is multiplied by the frequency separation and added to the minimum frequency and converted to radians. The analytical expression of the signal produced with this subroutine is illustrated in equation 6.8.

$$vc(t) = A \cos[(\omega + 2\pi mf)t + \delta] \quad (\text{eqn 6.8})$$

$mf$  = modulation frequency

This modulation continues until a symbol duration elapses when a new frequency deviation or  $mf$  is calculated

or modulation stops do to completion of all required samples.

#### H. SUBROUTINE QPRS

This module simulates 5 different classes of Quadrature Partial Response Signalling systems. The technique employed involves the summation, at each time increment, of all responses to the linear filter representing the class. This is accomplished for all of the bits being modulated during the duration of the simulation.

Initially, two arrays of random bits are generated representing those bits which would be modulated onto the quadrature components of the carrier during the length of time the signal is to be simulated. The impulse response to each of these  $n$  bits is calculated for the time of the respective sample. These impulse responses represent the modulation for the bit in question onto the respective portions of the carrier and the sum is formed according to equation 6.9.

$$v_c(t) = \sum_{n=-\infty}^{\infty} A h(t - nTs) \cos(\omega t + \delta) + \sum_{n=-\infty}^{\infty} A h(t - nTs) \sin(\omega t + \delta) \quad (\text{eqn 6.9})$$

$h(t - nTs) =$  impulse response of  $n$ th bit at time  $t$

The time of the first sample is initialized at  $6(\text{BAUDD})$  in order to ensure that when the first sample is taken, contributions from bits modulated at time 0 are also summed do to the overlapping nature of the QPR signals. The value of the last sample is also formed by the sum of the quadrature components existing at  $6(\text{BAUDD})$  after it is produced. The rest of the program operation remains the same as previous modules.



#### N. SUBROUTINE STREAM

This subroutine interacts with the IMSL random number generator SUBROUTINE GGUBFS to produce a random number between 0 and 1 and make assignment on the basis of the value of this random number to random bits according to the digital logic scheme specified in the MAIN program. It also enables the user to manually insert binary digits if that option was specified previously in MAIN.

#### O. SUBROUTINE ORTHO

CRTHO produces an  $n \times n$  orthogonal matrix from the Hadamard matrix by forming successive Kronecker products [Ref. 15]. It also selects the appropriate row and column of the orthogonal bit to be modulated and passes this bit back to SUBROUTINE OBPSK. Point of entry to SUBROUTINE ORTHO is controlled by a flag set in OBPSK.

#### P. SUBROUTINE PLOT

PLOT is an interactive module that allows the user to determine whether or not a graph of the output of program calculations is to be produced. The user is also able to selectively vary certain parameters associated with the plot. PLOT calls SUBROUTINE UTPLOT which actually produces the graph and performs the output. UTPLOT is a NON-IMSL library routine from the Naval Postgraduate School.

#### Q. SUBROUTINE STATS

The statistical calculations associated with the amplitude spectrums of the various functions generated on successive repetitions of the program are computed in SUBROUTINE STATS. Upon completion of each repetition, each module calls STATS where the sum, sum of the squares, sum of

the cubes and sum of the quartics of each element of the amplitude spectrum of the signal produced are computed and stored. When the last repetition of the program is complete, MAIN instructs STATS to compute and output the final statistics, i.e., mean, variance, skewness and kurtosis of each component of the amplitude spectrum according to the estimations contained in [Ref. 16]. Additionally the mean variance and variance of the variance is calculated and output. Point of entrance to this subroutine is controlled by flags set in the individual modulation subroutines or in MAIN.

## VII. STATISTICS OF THE FAST FOURIER TRANSFORMS

### A. DESCRIPTION OF THE PROBLEM

As stated in Chapter II, if it can be shown that the statistics of the FFT can be somehow linked to a particular digital signal modulation technique, then the modulation technique can be identified on the basis of those statistics alone. Although it is beyond the scope of this thesis to actually derive those relationships if they exist, it seems prudent to attempt to determine if there is statistical differences between the components of the FFT's as they are derived in the enclosed computer simulations.

The statistic chosen to do the hypothesis testing is the F-test since the true mean and variance of the distribution need not be known [Ref. 17]. In addition, the information necessary to calculate the F-statistic is readily accumulated in SUBROUTINE STATS. Calculation of the F-statistic necessitated the writing of a small computer program to be used for that purpose once the output from the main program was generated and reformatted.

### B. ANALYSIS-OF-VARIANCE

If there were no differences attributable to the modulation technique, then it would be reasonable to assume that for a certain set of characteristics, the components of successive FFT's would be the same. In other words, if a signal was modulated by BPSK and a statistically significant number of FFT's were generated of the signal, then the statistics of those FFT's would be assumed to be related. On the other hand, if it can be shown that the statistics are not related (i.e., not from the same distribution), then



significance can be placed in the variation among modulation techniques.

### 1. The F-Distribution

The assumptions necessary to use the F-distribution are:

1. Normal distribution
2. Random samples
3. Independent samples

These assumptions are not too difficult to intuitively accept in the model that will be proposed. Once again it would be expected that the FFT's of successive identically modulated signals to be related. For the simulations conducted as part of this test, they were all generated using bits from a random number generator for which it can be shown that each sequence of bits passes statistical tests for randomness and independence. The assumption of normality could also be argued on the basis of the central limit theorem.

The value of the F-statistic with which comparison will be made is 1.51 for 14 degrees of freedom in the numerator, an infinite number of degrees of freedom in the denominator and a 90% confidence region. How these values were obtained will be detailed under the design of the experiment.

### 2. Design of the Experiment

The experimental data used in the computation of the statistics was chosen to minimize the random contributions of variables other than the modulation technique. The computer simulations included in this thesis were used to generate the statistics and a separate program also included in the appendix was used to calculate the F-statistic. The variables in the experiment include:

1. Modulation technique
2. Logic type
3. Baud rate
4. Bits per binary code word
5. Carrier frequency
6. Carrier amplitude
7. Initial phase angle
8. Number of samples generated
9. Time between samples
10. Seed for random number generator
11. Number of repetitions or trials

Fifteen different modulation techniques were compared. In all cases bipolar logic was simulated, the baud rate was held constant at 1200 baud and the maximum carrier amplitude was established at 1 volt. Additionally the phase angle of all simulations was 0 degrees and the seed for the random number generator was 1 thereby ensuring the same random sequence of bits. The number of bits per binary code word did vary among the modulation techniques. When the modulation technique was M-ary, the bits per code word was always 3 so it is possible to infer that bits per code word is a function of the modulation technique. The time between samples was the normalized Nyquist sampling rate. This did vary between modulation techniques but was necessary to derive an accurate FFT. Carrier frequency also varied between simulations but was always twice the lowest carrier frequency recommended in the computer simulation. In most cases this was 2400 Hz, or twice the baud rate. Finally each simulation was repeated 100 times and the statistics of the FFT's based on those 100 repetitions. The number of samples produced was always 64 so there are 33 components of the FFT.

What this amounts to can be illustrated through an Analysis-of-Variance table as shown in Table 2 [Ref. 17].

**TABLE 2**  
**ANALYSIS-OF-VARIANCE TABLE**

*n* Observations in Each of *r* Groups

		Observations	Sum	Mean
Group	1	$Y_{11} \quad Y_{12} \quad Y_{13} \quad \dots \quad Y_{1n}$	$Y_{1.}$	$\bar{Y}_{1.}$
	2	$Y_{21} \quad Y_{22} \quad Y_{23} \quad \dots \quad Y_{2n}$	$Y_{2.}$	$\bar{Y}_{2.}$
	3	$Y_{31} \quad Y_{32} \quad Y_{33} \quad \dots \quad Y_{3n}$	$Y_{3.}$	$\bar{Y}_{3.}$
	<i>r</i>	$Y_{r1} \quad Y_{r2} \quad Y_{r3} \quad \dots \quad Y_{rn}$	$Y_{r.}$	$\bar{Y}_{r.}$

In the case of the simulation described above, this amounts to 100 observations (repetitions or trials) from each of 15 groups (modulation techniques) for each of the 33 components of the FFT. The degrees of freedom in the numerator is equal to (15-1) or 14 while the degrees of freedom in the denominator is equal to 15(100-1) or 1485. Table values for the F-distribution use infinity as the degree of freedom for values greater than 120.

An F-test was also performed to determine if there is a relationship among the mean variance, mean skewness or mean kurtosis of the 15 different modulation techniques. Again the assumption is made that these statistics would be normally distributed from modulation techniques which had the same or statistically similar components of their FFT's. In this case there were again 15 modulation techniques to compare (14 degrees of freedom) but only 33 observations (the variance, skewness or kurtosis of the respective elements of the FFT). This number of observations yields 15(33-1) = 480 degrees of freedom in the denominator.



### 3. Hypothesis and Hypothesis Testing

The hypothesis posed for the model is that the means of the individual components of the FFT's from the 15 modulation techniques are from the same distribution. Therefore, the respective means must be equal. Also tested is that the mean variance, mean skewness and mean kurtosis of all components of the FFT's of each modulation technique are equal. These hypothesis are illustrated in equations 7.1 through equation 7.4. Each hypothesis is tested and compared to the F-distribution selecting a rejection region or level of significance of .9. This equates to an F-statistic of 1.51 for 14 degrees of freedom in the numerator and an infinite number of degrees of freedom in the denominator.

### 4. Results

The means of the 33 components of the FFT for each of the 15 modulation techniques were compared using the F-distribution. The results are shown in Table 3. Table 4 shows the F-statistics associated with the comparison of the mean of the variance, mean of the skewness and mean of the kurtosis for the respective groups. In addition a complete summation of the results of the F-test are included in the appendix.

### C. **CONCLUSION**

The results of the F-test indicate that the null hypothesis must be rejected at the .9 significance level and the alternate hypothesis accepted that the means are not equal for FFT components 9-14. In addition the F-statistic for the comparison of mean variance, mean skewness and mean kurtosis all fall in the rejection area.

$$H_0: \bar{X}_{11} = \bar{X}_{21} = \dots = \bar{X}_{ij} \quad (\text{eqn 7.1})$$

$$H_1: \bar{X}_{11} \neq \bar{X}_{21} \neq \dots \neq \bar{X}_{ij}$$

$i$  =  $i$ th modulation technique

$j$  =  $j$ th component of the FFT

$$H_0: \text{MVAR}_1 = \text{MVAR}_2 = \dots = \text{MVAR}_{15} \quad (\text{eqn 7.2})$$

$$H_1: \text{MVAR}_1 \neq \text{MVAR}_2 \neq \dots \neq \text{MVAR}_{15}$$

$$H_0: \text{MSKEW}_1 = \text{MSKEW}_2 = \dots = \text{MSKEW}_{15} \quad (\text{eqn 7.3})$$

$$H_1: \text{MSKEW}_1 \neq \text{MSKEW}_2 \neq \dots \neq \text{MSKEW}_{15}$$

$$H_0: \text{MKUR}_1 = \text{MKUR}_2 = \dots = \text{MKUR}_{15} \quad (\text{eqn 7.4})$$

$$H_1: \text{MKUR}_1 \neq \text{MKUR}_2 \neq \dots \neq \text{MKUR}_{15}$$

This indicates that there is a difference between the statistics of the FFT's of the respective modulation technique. Since all the parameters used in the generation of these statistics were held essentially constant among the trials, it can be assumed that these differences are due to the modulation technique employed. It has not been established yet what those differences may be. This is an area where future research and study is warranted.

The results of this test would seem to point to those components of the FFT which offer the best opportunity to develop those relationships or differences. FFT components 9-14 have obvious differences; however, FFT components 20, 24 and 25 also have large F-statistics but do not fall in the rejection area. These components may also be

**TABLE 3**  
**F-STATISTICS OF FFT COMPONENTS**

FFT CCMPONENT	AMONG GROUP SUM	WITHIN GROUP SUM	TOTAL OF SQUARES	F-STAT
1	0.121	17.69	17.81	0.727
2	0.112	18.48	18.59	0.642
3	0.108	20.65	20.76	0.557
4	0.135	25.40	25.54	0.562
5	0.172	32.74	32.91	0.558
6	0.254	41.72	41.98	0.647
7	0.422	50.10	50.52	0.893
8	0.962	82.83	83.79	1.231
9	7.591	390.2	397.8	2.063
10	49.09	1681.0	1730.1	3.097
11	53.01	1732.1	1785.1	3.246
12	61.56	2071.7	2133.3	3.152
13	45.73	1832.8	1878.6	2.646
14	3.864	250.5	254.4	1.636
15	0.539	98.13	98.67	0.582
16	0.366	110.1	110.5	0.353
17	0.504	151.0	151.5	0.354
18	0.943	140.3	141.2	0.713
19	1.299	158.5	159.8	0.870
20	2.671	200.7	203.4	1.412
21	1.853	226.3	228.2	0.868
22	2.277	232.1	234.3	1.041
23	1.712	197.6	199.3	0.919
24	2.512	181.9	184.4	1.476
25	2.144	171.3	173.5	1.327
26	1.257	143.7	145.0	0.928
27	1.233	113.9	115.1	1.149



TABLE 3  
F-STATISTICS OF FFT COMPONENTS (con't)

28	0.858	115.8	116.7	0.785
29	0.547	63.72	64.27	0.911
30	0.356	55.77	56.13	0.677
31	0.183	41.41	51.59	0.469
32	0.051	13.79	13.84	0.389
33	0.031	10.92	10.96	0.304

TABLE 4  
F-STATISTICS OF THE MEAN VARIANCES,  
MEAN SKEWNESS AND MEAN KURTOSIS

	AMONG GROUP SUM	WITHIN GROUP SUM	TOTAL OF SQUARES	F-STAT
VARIANCE	3.144 E5	3.755 E6	4.070 E6	2.870
SKEWNESS	5.307 E10	7.031 E11	7.562 E11	2.588
KURTOSIS	8.280 E11	1.262 E13	1.344 E13	2.250

interesting to examine. In addition, it should be noted that the statistics of the FFT associated with each respective modulation technique also display some striking differences. These statistics may prove in some way to be a fingerprint of the modulation techniques themselves.

APPENDIX A

FORTRAN PROGRAM FOR DIGITAL SIGNAL MODULATION

```

*****
**      DIGITAL SIGNAL MODULATION TECHNIQUES
**
**      TITLE:  DIGITAL SIGNAL MODULATION TECHNIQUES
**
**      PURPOSE:  THIS PROGRAM GENERATES A SIMULATED WAVEFORM WHICH IS
**      CHARACTERISTIC OF ONE OF A VARIETY OF DIGITAL SIGNAL MODULATION
**      TECHNIQUES.  THE PROGRAM IS INTERACTIVE AND DESIGNED TO BE AS
**      USER FRIENDLY AS POSSIBLE.  THE PROGRAM ALLOWS THE USER TO
**      SPECIFY ANY ONE OF A NUMBER OF PARAMETERS OR VARIABLES WHICH
**      MAKE CHARACTERISTIC OF THE INDIVIDUAL MODULATION TECHNIQUE.
**      THE PROGRAM ALSO ALLOWS THE USER TO DEVELOP A PLOT OF THE
**      SIMULATED WAVEFORM, GENERATE THE FAST FOURIER TRANSFORM,
**      OR DO LIMITED STATISTICAL ANALYSIS.
**
**      WRITTEN BY:  CRAIG CARLSON, LCDR, USN
**      IN PARTIAL FULFILLMENT OF MASTER OF SCIENCE, SYSTEMS TECHNOLOGY
**
**      THIS PROGRAM WAS WRITTEN FOR EXECUTION ON AN IBM 3033 COMPUTER
**      UTILIZING THE VMS/CMS OPERATING SYSTEM.  WITHIN THE CMS SYSTEM
**      THERE IS A ROUTINE TO CLEAR THE SCREEN WHEN DESIRED.  THIS
**      FUNCTION IS CALLED THROUGH THE COMMAND *
**      CALL FRTCMS(CLRSCRN).  IF THE PROGRAM IS BEUSED ON THIS
**      OPERATING SYSTEM, REMOVAL OF THE COMMENT C BEFORE LINES OF
**      CODE CONTAINING THIS COMMAND WILL IMPROVE THE INTERACTIVE
**      PORTION OF THE PROGRAM.
**
**      *****
**      MAIN CONTROL MODULE **
**
**      PURPOSE ***
**
**      THIS PROGRAM IS THE MAIN BODY OF THE SIGNAL GENERATOR.  IT
**      CONTROLS INPUT AND OUTPUT AND CALLS VARIOUS SUBROUTINES DURING
**      ITS OPERATION.
**
**      VARIABLE DEFINITIONS ***
**
**      ANS=
**      TYPE1=
**      TYPE2=
**      TYPE3=
**      ANS2=

```

CC













```

C      READ (5,*) IBAUD
C      BAUD= IBAUD
C      BAUDD=1. DO/BAUD
C      ** DETERMINE THE BITS EITHER BY THE TYPE OF DIGITAL
C      MODULATION SPECIFIED OR BY USER INPUT ***
C      CALL FRTCMS ('CLRSCRN ')
C      IF (TYPE1.EQ. 1.OR.TYPE1.EQ. 2.OR.TYPE1.EQ. 9) THEN
C          IBITS=1
C          BITS=IBITS
C      ELSE IF (TYPE1.EQ. 3) THEN
79      WRITE(10,80)
80      FORMAT(' ENTER THE NUMBER OF BITS IN EACH SYMBOL. ' // OR LESS'
C          ' CAUTION! THE NUMBER OF BITS PER SYMBOL MUST BE 6 OR LESS'
C          ' // ' : ')
C      READ (5,*) IBITS
C      IF (IBITS.LT. 1.OR. IBITS.GT. 6) THEN
C          CALL FRTCMS ('CLRSCRN ')
90      WRITE(10,90)
C          FORMAT(' ERROR' //)
C          GO TO 79
C      ELSE
C          BITS=IBITS
C      END IF
C      ELSE IF (TYPE1.EQ. 4.OR.TYPE1.EQ. 5.OR.TYPE1.EQ. 11) THEN
C          IBITS=2
C          BITS=IBITS
C      ELSE IF (TYPE1.EQ. 6.OR.TYPE1.LE. 8.OR.TYPE1.EQ. 10) THEN
99      WRITE(10,100)
100     FORMAT(' ENTER THE NUMBER OF BITS IN EACH SYMBOL. ' // OR LESS'
C         ' CAUTION! THE NUMBER OF BITS PER SYMBOL MUST BE 10 OR LESS'
C         ' // ' : ')
C      READ (5,*) IBITS
C      IF (IBITS.LT. 1.OR. IBITS.GT. 11) THEN
C          CALL FRTCMS ('CLRSCRN ')
C          WRITE(10,110)
110     FORMAT(' ERROR' //)
C          GO TO 99

```



```

ELSE      BITS=IBITS
END IF
END IF
*** DETERMINE AND DISPLAY THE BIT RATE ***
IF (TYPE1.EQ.3) THEN
  BITR=(2.D0**IBITS)*BAUD
ELSE
  BITR=BITS*BAUD
END IF
CALL FRTCMS ('CLRSCRN ')
WRITE (10,120) BITR
FORMAT (' THE BIT RATE FOR THE SPECIFIED SIGNAL IS',F9.0,' BITS/SEC
*)
*** HAVE THE USER ENTER THE CARRIER FREQUENCY ***
IF (TYPE1.EQ.3) THEN
  FC=BITR
ELSE IF (TYPE1.EQ.9) THEN
  FC=.25D0*BAUD
ELSE IF (TYPE1.EQ.10) THEN
  FC=BAUD+((2.D0**IBITS)-1.D0)*(BAUD/2.D0)
ELSE
  FC=BAUD
END IF
WRITE (10,130) FC
FORMAT (' ENTER THE CARRIER FREQUENCY AT THIS TIME.',F9.0,' // HZ.,//.
* : ')
READ (5,*) IFREQ
FREQ=IFREQ
IF (FREQ.LT.FC) THEN
  CALL FRTCMS ('CLRSCRN ')
  WRITE (10,140)
  FORMAT (' THE CARRIER FREQUENCY IS LESS THAN RECOMMENDED.',/)
  GO TO 149
END IF
*** HAVE THE USER ENTER THE NUMBER OF SAMPLES TO BE GENERATED ***
CALL FRTCMS ('CLRSCRN ')

```

```

MAI 02450
MAI 02460
MAI 02470
MAI 02480
MAI 02490
MAI 02500
MAI 02510
MAI 02520
MAI 02530
MAI 02540
MAI 02550
MAI 02560
MAI 02570
MAI 02580
MAI 02590
MAI 02600
MAI 02610
MAI 02620
MAI 02630
MAI 02640
MAI 02650
MAI 02660
MAI 02670
MAI 02680
MAI 02690
MAI 02700
MAI 02710
MAI 02720
MAI 02730
MAI 02740
MAI 02750
MAI 02760
MAI 02770
MAI 02780
MAI 02790
MAI 02800
MAI 02810
MAI 02820
MAI 02830
MAI 02840
MAI 02850
MAI 02860
MAI 02870
MAI 02880
MAI 02890
MAI 02900
MAI 02910
MAI 02920
MAI 02930
MAI 02940

```

```

C 149 WRITE(10,150)
C 150 FORMAT(' ENTER THE NUMBER OF SAMPLES OF THE MODULATED SIGNAL,
* TO BE PRODUCED. THIS NUMBER MUST BE EQUAL TO 2**N
* WHERE N IS A POSITIVE INTEGER EQUAL TO OR LESS THAN 10.://,
* I.E., 2, 4, 8, ..., 1024.://,')
C READ(5,*) IN
C N=IN
C IF(IN.EQ.2.OR.IN.EQ.4.OR.IN.EQ.8.OR.IN.EQ.16.OR.IN.EQ.32.OR.
*IN.EQ.64.OR.IN.EQ.128.OR.IN.EQ.256.OR.IN.EQ.512.OR.IN.EQ.1024) THEN
GO TO 170
C ELSE
CALL FRTCMS('CLRSCRN ')
WRITE(10,160)
FORMAT('ERROR',/)
GO TO 149
C END IF
C *** DETERMINE AND DISPLAY INFO ABOUT THE RECORD LENGTH ***
C 160 IF(TYPE1.EQ.2) THEN 0*(FREQ+BITR))
ELSE STEP=1.D0/(2.D0*(FREQ+BITR))
ELSE STEP=1.D0/(2.D0*(FREQ+(1.25D0*BAUD)))
ELSE IF(TYPE1.EQ.10) THEN
STEP=1.D0/(2.D0*(FREQ+((2.D0**IBITS)-1.)*(BAUD/2.D0)+BAUD))
ELSE IF(TYPE1.EQ.11) THEN
STEP=1.D0/(2.D0*(FREQ+(PI+1.D0)*BAUD))
ELSE
STEP=1.D0/(2.D0*(FREQ+BAUD))
END IF
C IF(TYPE1.EQ.3.AND.((N-1.)*STEP).LE.1./BITR) THEN
C CALL FRTCMS('CLRSCRN ')
WRITE(10,180)
FORMAT(' ELAPSED TIME FOR SIGNAL GENERATION WILL BE LESS THAN
* INCREASING THE NUMBER OF SAMPLES TO BE PRODUCED, BY
* INCREASING CARRIER FREQUENCY, BY DECREASING THE BAUD
* RATE OR BY DECREASING THE SIZE OF THE BINARY CODE WORD.
* ENTER A 1. ENTER ANY OTHER INTEGER VALUE TO CONTINUE.://,')
C READ(5,*) ANS
C IF(ANS.EQ.1) THEN
GO TO 29
END IF

```

```

MAI 02950
MAI 02960
MAI 02970
MAI 02980
MAI 02990
MAI 03000
MAI 03010
MAI 03020
MAI 03030
MAI 03040
MAI 03050
MAI 03060
MAI 03070
MAI 03080
MAI 03090
MAI 03100
MAI 03110
MAI 03120
MAI 03130
MAI 03140
MAI 03150
MAI 03160
MAI 03170
MAI 03180
MAI 03190
MAI 03200
MAI 03210
MAI 03220
MAI 03230
MAI 03240
MAI 03250
MAI 03260
MAI 03270
MAI 03280
MAI 03290
MAI 03300
MAI 03310
MAI 03320
MAI 03330
MAI 03340
MAI 03350
MAI 03360
MAI 03370
MAI 03380
MAI 03390
MAI 03400
MAI 03410
MAI 03420
MAI 03430
MAI 03440

```



```

C      ELSE IF ((N-1.D0)*STEP).LE.BAUDD) THEN
C      CALL FRTCMS('CLRSCRN ')
C      WRITE(10,190)
190    FORMAT(1,ELAPSED TIME FOR SIGNAL GENERATION WILL BE LESS THAN
*      /,SYMBOL DURATION. IF YOU WISH TO CHANGE THIS BY
*      /,INCREASING CARRIER FREQUENCY, BY DECREASING THE BAUDD
*      /,DECREASING CARRIER FREQUENCY, BY DECREASING THE BAUDD
*      /,RATE OR BY DECREASING THE SIZE OF THE BINARY CODE WORD.
*      /,ENTER A 1. ENTER ANY OTHER INTEGER VALUE TO CONTINUE.
*      /,WITH THE SIMULATION AS SPECIFIED.//,,:')
C      READ(5,*)ANS
C      IF(ANS.EQ.1) THEN
C      GO TO 29
C      END IF
C      END IF
C      *** HAVE THE USER ENTER THE AMPLITUDE OF THE CARRIER WAVE ***
C      CALL FRTCMS('CLRSCRN ')
199    WRITE(10,200)
200    FORMAT(,ENTER THE AMPLITUDE OF THE CARRIER.//,,:')
C      READ(5,*)IAMP
C      AMP=IAMP
C      *** HAVE THE USER ENTER THE INITIAL PHASE ANGLE ***
C      CALL FRTCMS('CLRSCRN ')
209    WRITE(10,210)
210    FORMAT(,ENTER THE INITIAL PHASE ANGLE FROM 0 TO 360 DEGREES.//,,:')
C      READ(5,*)IIPHAS
C      IF(IIPHAS.LT.0.OR.IIPHAS.GT.360) THEN
C      CALL FRTCMS('CLRSCRN ')
C      WRITE(10,220)
220    FORMAT(,ERROR.//)
C      GO TO 209
C      ELSE
C      IPHAS=IIPHAS
C      END IF
C

```

```

MAI 03450
MAI 03460
MAI 03470
MAI 03480
MAI 03490
MAI 03500
MAI 03510
MAI 03520
MAI 03530
MAI 03540
MAI 03550
MAI 03560
MAI 03570
MAI 03580
MAI 03590
MAI 03600
MAI 03610
MAI 03620
MAI 03630
MAI 03640
MAI 03650
MAI 03660
MAI 03670
MAI 03680
MAI 03690
MAI 03700
MAI 03710
MAI 03720
MAI 03730
MAI 03740
MAI 03750
MAI 03760
MAI 03770
MAI 03780
MAI 03790
MAI 03800
MAI 03810
MAI 03820
MAI 03830
MAI 03840
MAI 03850
MAI 03860
MAI 03870
MAI 03880
MAI 03890
MAI 03900
MAI 03910
MAI 03920
MAI 03930
MAI 03940

```





```

C      ELSE IF((TYPE1.EQ.2) THEN
C      CALL DBPSK(TYPE2, BAUD, FREQ, IPHAS, AMP, ANS2, DSEED, IN, REP)
C
C      ELSE IF((TYPE1.EQ.3) THEN
C      CALL OBPSK(TYPE2, BAUD, BITS, FREQ, IPHAS, AMP, ANS2, DSEED, IN, REP)
C
C      ELSE IF((TYPE1.EQ.4) THEN
C      CALL QPSK(TYPE2, BAUD, BITS, FREQ, IPHAS, AMP, ANS2, DSEED, IN, REP)
C
C      ELSE IF((TYPE1.EQ.5) THEN
C      CALL OQPSK(TYPE2, BAUD, BITS, FREQ, IPHAS, AMP, ANS2, DSEED, IN,
*      REP)
C
C      ELSE IF((TYPE1.EQ.6) THEN
C      CALL MPSK(TYPE2, BAUD, BITS, FREQ, IPHAS, AMP, ANS2, DSEED, IN,
*      REP)
C
C      ELSE IF((TYPE1.EQ.7) THEN
C      CALL MASK(TYPE2, BAUD, BITS, FREQ, IPHAS, AMP, ANS2, DSEED, IN,
*      REP)
C
C      ELSE IF((TYPE1.EQ.8) THEN
C      CALL QASK(TYPE2, BAUD, BITS, FREQ, IPHAS, AMP, ANS2, DSEED, IN,
*      REP)
C
C      ELSE IF((TYPE1.EQ.9) THEN
C      CALL MSK(TYPE2, BAUD, FREQ, IPHAS, AMP, ANS2, DSEED, IN, REP)
C
C      ELSE IF((TYPE1.EQ.10) THEN
C      CALL MFSK(TYPE2, BAUD, BITS, FREQ, IPHAS, AMP, ANS2, DSEED, IN, REP)
C
C      ELSE IF((TYPE1.EQ.11) THEN
C      CALL OPFS(TYPE2, TYPE3, BAUD, BITS, FREQ, IPHAS, AMP, ANS2, DSEED,
*      IN, REP)
C
C      END IF
C
C      CONTINUE
C 280
C      ** OUTPUT A FINGERPRINT OF THE MODULATION ACCOMPLISHED ***
C
C      IF((TYPE1.EQ.1) THEN
C      WRITE(6,1) THEN
C      FORMAT(1, MODULATION TECHNIQUE = BPSK')
C 281
C      ELSE IF((TYPE1.EQ.2) THEN
C      WRITE(6,2)
C      FORMAT(1, MODULATION TECHNIQUE = DBPSK')
C 282
C      ELSE IF((TYPE1.EQ.3) THEN
C      WRITE(6,3)

```

```

MAI 04450
MAI 04460
MAI 04470
MAI 04480
MAI 04490
MAI 04500
MAI 04510
MAI 04520
MAI 04530
MAI 04540
MAI 04550
MAI 04560
MAI 04570
MAI 04580
MAI 04590
MAI 04600
MAI 04610
MAI 04620
MAI 04630
MAI 04640
MAI 04650
MAI 04660
MAI 04670
MAI 04680
MAI 04690
MAI 04700
MAI 04710
MAI 04720
MAI 04730
MAI 04740
MAI 04750
MAI 04760
MAI 04770
MAI 04780
MAI 04790
MAI 04800
MAI 04810
MAI 04820
MAI 04830
MAI 04840
MAI 04850
MAI 04860
MAI 04870
MAI 04880
MAI 04890
MAI 04900
MAI 04910
MAI 04920
MAI 04930
MAI 04940

```



MAI 04950  
 MAI 04960  
 MAI 04970  
 MAI 04980  
 MAI 04990  
 MAI 05000  
 MAI 05010  
 MAI 05020  
 MAI 05030  
 MAI 05040  
 MAI 05050  
 MAI 05060  
 MAI 05070  
 MAI 05080  
 MAI 05090  
 MAI 05100  
 MAI 05110  
 MAI 05120  
 MAI 05130  
 MAI 05140  
 MAI 05150  
 MAI 05160  
 MAI 05170  
 MAI 05180  
 MAI 05190  
 MAI 05200  
 MAI 05210  
 MAI 05220  
 MAI 05230  
 MAI 05240  
 MAI 05250  
 MAI 05260  
 MAI 05270  
 MAI 05280  
 MAI 05290  
 MAI 05300  
 MAI 05310  
 MAI 05320  
 MAI 05330  
 MAI 05340  
 MAI 05350  
 MAI 05360  
 MAI 05370  
 MAI 05380  
 MAI 05390  
 MAI 05400  
 MAI 05410  
 MAI 05420  
 MAI 05430  
 MAI 05440

```

283     ELSE FORMAT('1', MODULATION TECHNIQUE = ORTHOGONAL BPSK')
        IF (TYPE(6, EQ. 4)) THEN
284     FORMAT('1', MODULATION TECHNIQUE = QPSK')
        IF (TYPE(6, EQ. 5)) THEN
285     FORMAT('1', MODULATION TECHNIQUE = OQPSK')
        IF (TYPE(6, EQ. 6)) THEN
286     FORMAT('1', MODULATION TECHNIQUE = MPSK')
        IF (TYPE(6, EQ. 7)) THEN
287     FORMAT('1', MODULATION TECHNIQUE = MASK')
        IF (TYPE(6, EQ. 8)) THEN
288     FORMAT('1', MODULATION TECHNIQUE = QASK')
        IF (TYPE(6, EQ. 9)) THEN
289     FORMAT('1', MODULATION TECHNIQUE = MSK')
        IF (TYPE(6, EQ. 10)) THEN
290     FORMAT('1', MODULATION TECHNIQUE = MFSK')
        IF (TYPE(6, EQ. 11)) THEN
291     FORMAT('1', MODULATION TECHNIQUE = QPRS')
    C
292     IF (TYPE 1, EQ. 11, AND, TYPE3, EQ. 1) THEN
        WRITE(6, QPRS CLASS 1 FILTER')
        ELSE IF (TYPE(6, EQ. 11, AND, TYPE3, EQ. 2)) THEN
293     WRITE(6, QPRS CLASS 2 FILTER')
        ELSE IF (TYPE(6, EQ. 11, AND, TYPE3, EQ. 3)) THEN
294     WRITE(6, QPRS CLASS 3 FILTER')
        ELSE IF (TYPE(6, EQ. 11, AND, TYPE3, EQ. 4)) THEN
295     WRITE(6, QPRS CLASS 4 FILTER')
        ELSE IF (TYPE(6, EQ. 11, AND, TYPE3, EQ. 5)) THEN
296     WRITE(6, QPRS CLASS 5 FILTER')
    C
297     IF (TYPE 2, EQ. 1) THEN
        WRITE(6, BIPOLAR LOGIC')
        ELSE IF (TYPE(6, EQ. 2)) THEN
298     WRITE(6, UNIPOLAR POSITIVE LOGIC')
  
```





MAI 05950  
MAI 05960  
MAI 05970  
MAI 05980  
MAI 05990  
MAI 06000  
MAI 06010  
MAI 06020  
MAI 06030  
MAI 06040  
MAI 06050  
MAI 06060  
MAI 06070  
MAI 06080  
MAI 06090  
MAI 06100  
MAI 06110

```
END IF  
*** DETERMINE IF ANOTHER SIMULATION IS TO BE RUN ***  
WRITE(10,330)  
FORMAT(' ENTER A 1 IF YOU DESIRE TO SIMULATE ANOTHER SIGNAL. ',//,  
* ' ENTER ANY OTHER INTEGER IF YOU ARE READY TO QUIT. ',//,  
: ')  
READ (5,*)ANS  
IF (ANS.EQ.1) THEN  
  CALL FRTCMS('CLRSCRN ')  
  GO TO 29  
END IF  
STOP  
END
```

C  
C  
C  
330  
C  
C  
C  
C

BPS00110  
 BPS00120  
 BPS00130  
 BPS00140  
 BPS00150  
 BPS00160  
 BPS00170  
 BPS00180  
 BPS00190  
 BPS00200  
 BPS00210  
 BPS00220  
 BPS00230  
 BPS00240  
 BPS00250  
 BPS00260  
 BPS00270  
 BPS00280  
 BPS00290  
 BPS00300  
 BPS00310  
 BPS00320  
 BPS00330  
 BPS00340  
 BPS00350  
 BPS00360  
 BPS00370  
 BPS00380  
 BPS00390  
 BPS00400  
 BPS00410  
 BPS00420  
 BPS00430  
 BPS00440  
 BPS00450  
 BPS00460  
 BPS00470  
 BPS00480  
 BPS00490  
 BPS00500  
 BPS00510  
 BPS00520  
 BPS00530  
 BPS00540  
 BPS00550  
 BPS00560  
 BPS00570  
 BPS00580  
 BPS00590  
 BPS00600

```

SUBROUTINE BPSK (TYPE2, BAUD, FREQ, IPHAS, AMP, ANS2,
 *DSEED, IN, REP)
 *** PURPOSE ***
 THIS SUBROUTINE MODULATES THE CARRIER USING BINARY PHASE SHIFT
 KEYING AS THE MODULATION TECHNIQUE.
 *** PARAMETER DEFINITIONS ***
 TYPE2= INDICATES LOGIC TYPE TO BE EMPLOYED
 BAUD= SYMBOL RATE OR BAUD RATE
 FREQ= CARRIER FREQUENCY
 IPHAS= INITIAL PHASE ANGLE IN DEGREES
 AMP= AMPLITUDE
 ANS2= INTEGER VARIABLE TO BE PASSED TO STREAM GENERATOR
 DSEED= DOUBLE PRECISION SEED FOR RANDOM NUMBER GENERATOR
 IN= NUMBER OF POSITIONS IN ARRAY TO BE UTILIZED
 REP= AN INTEGER EQUAL TO THE NUMBER OF THE REPETITION OF
 THE CALL THE THIS SUBROUTINE
 *** VARIABLE DEFINITIONS ***
 ARRAY= A DOUBLE PRECISION ARRAY FOR PASSING THE BINARY
 DIGITS AND STORING THE VALUE OF THE TIME FUNCTION
 AND THE FFT
 AN INTEGER USED TO REPRESENT THE ROW OF ARRAY
 TIME VARIABLE USED TO REPRESENT THE ROW OF ARRAY
 VALUE OF THE BINARY DIGIT
 INTERVAL AT WHICH THE SIGNAL IS REPRODUCED
 NUMERICAL CONSTANT FREQUENCY
 CARRIER ANGULAR OFFSET IN RADIAN
 INITIAL PEASE OFFSET IN RADIAN
 A COUNT OF THE NUMBER OF BAUDS MODULATED
 BAUD DURATION
 COMPLEX ARRAY TO RECEIVE THE FFT OF THE TIME SERIES
 INTEGER WORKING ARRAY USED BY FUNCTION FFTRC
 MAXIMUM VALUE TO BE PLOTTED ON THE ABCISSAE
 MINIMUM VALUE TO BE PLOTTED ON THE ABCISSAE
 INTERVAL BETWEEN POINTS ON THE ORDINATE
 VALUE OF THE PRINCIPLE HARMONIC
 AN INTEGER WHICH CONTROL ENTRANCE AND EXIT FROM
 SUBPROGRAM STATS
 ARRAY LENGTH DIVIDED BY 2 PLUS 1
 *** VARIABLE DECLARATIONS ***
 INTEGER R, ANS2, TYPE2, IN, RMAX, REP, FLAG, IND2P1
 DOUBLE PRECISION BAUD, FREQ, IPHAS, AMP, TIME, MT,

```

CC



```

C      *STEP,PI,OMEGA,DELTA,NBAUD,BAUDD,
C      *DSEED,MAXY,MINY,INT
C      DOUBLE PRECISION ARRAY(1024,2),SUMX(513),SUMXSQ(513),SUMX3(513),
C      *SUMX4(513)
C      COMMON ARRAY,SUMX,SUMXSQ,SUMX3,SUMX4
C      COMPLEX*16 X(513)
C      INTEGER IWK(10)
C      *** VARIABLE INITIALIZATION ***
C      R=1
C      TIME=0. DO
C      STEP=1. DO/(2. DO*(FREQ+BAUD))
C      PI=3.141592653589793 DO
C      OMEGA=2. DO*PI*FREQ
C      DELTA=IPHAS*PI/180. DO
C      NBAUD=1. DO
C      BAUDD=1. DO/BAUD
C      MAXY=0. DO
C      IND2P1=IN/2+1
C      *** DO THE MODULATION AND ASSIGN THE VALUES TO ARRAY ***
C      CALL STREAM(DSEED,ANS2,TYPE2,MT)
C      DO 10 I=1,IN
C      ARRAY(R,1)=AMP*MT*DCOS((OMEGA*TIME)+DELTA)
C      ARRAY(R,2)=TIME
C      IF(DABS(ARRAY(R,1))-GT.MAXY) THEN
C      MAXY=DABS(ARRAY(R,1))
C      END IF
C      R=R+1
C      TIME=TIME+STEP
C      IF(TIME-GT.NBAUD*BAUDD) THEN
C      CALL STREAM(DSEED,ANS2,TYPE2,MT)
C      NBAUD=NBAUD+1. DO
C      END IF
C      CONTINUE
C      *** PLOT THE TIME SERIES IF DESIRED ***
C      IF(REP.EQ.1) THEN
C      MINY=-MAXY
C      CALL PLOT(MAXY,MINY,STEP,IN)
C      END IF

```

```

BPS00610
BPS00620
BPS00630
BPS00640
BPS00650
BPS00660
BPS00670
BPS00680
BPS00690
BPS00700
BPS00710
BPS00720
BPS00730
BPS00740
BPS00750
BPS00760
BPS00770
BPS00780
BPS00790
BPS00800
BPS00810
BPS00820
BPS00830
BPS00840
BPS00850
BPS00860
BPS00870
BPS00880
BPS00890
BPS00900
BPS00910
BPS00920
BPS00930
BPS00940
BPS00950
BPS00960
BPS00970
BPS00980
BPS00990
BPS01000
BPS01010
BPS01020
BPS01030
BPS01040
BPS01050
BPS01060
BPS01070
BPS01080
BPS01090
BPS01100

```

```

C C C C C
*** GENERATE THE FFT ***
CALL FFTRC (ARRAY(1,1), IN, X, IWK)
*** CALCULATE THE AMPLITUDE SPECTRUM OF THE FUNCTION ***
*** FIND THE NUMBER AND VALUE OF THE PRINCIPLE HARMONIC ***
N=0. DO
R=1
MAXY=0. DO
DO 20 I=1, IND2P1
ARRAY(R,1)=CDABS(X(R))
ARRAY(R,2)=N/STEP
IF (ARRAY(R,1) .GT. MAXY) THEN
MAXY=ARRAY(R,1)
RMAX=R-1
END IF
R=R+1
N=N+1. DO
CONTINUE
*** DISPLAY INFO IF THE FIRST TIME THROUGH SUBPROGRAM ***
IF (REP.EQ.1) THEN
WRITE(10,30) RMAX
FORMAT(1,5,' HARMONIC. THE NEXT PLOT TO BE PRODUCED WILL',//
*! IS THE AMPLITUDE SPECTRUM. ENTER A 1 IF YOU ARE READY TO',//
*! CONTINUE WITH THE PROGRAM.',//, :')
READ(5,*) L
IF (L.NE.1) THEN
CALL FRTCMS('CLRSCRN ')
WRITE(10,40)
FORMAT(1, 'ERROR',/)
GO TO 29
END IF
END IF
*** ALLOW FOR THE PLOT OF THE AMPLITUDE SPECTRUM OF THE FFT ***
IF (REP.EQ.1) THEN
MINY=0. DO
INT=1. DO/STEP
CALL PLOT(MAXY, MINY, INT, IND2P1)
END IF

```

BPS01110  
BPS01120  
BPS01130  
BPS01140  
BPS01150  
BPS01160  
BPS01170  
BPS01180  
BPS01190  
BPS01200  
BPS01210  
BPS01220  
BPS01230  
BPS01240  
BPS01250  
BPS01260  
BPS01270  
BPS01280  
BPS01290  
BPS01300  
BPS01310  
BPS01320  
BPS01330  
BPS01340  
BPS01350  
BPS01360  
BPS01370  
BPS01380  
BPS01390  
BPS01400  
BPS01410  
BPS01420  
BPS01430  
BPS01440  
BPS01450  
BPS01460  
BPS01470  
BPS01480  
BPS01490  
BPS01500  
BPS01510  
BPS01520  
BPS01530  
BPS01540  
BPS01550  
BPS01560  
BPS01570  
BPS01580  
BPS01600

C  
C  
C  
C

```
*** ADD THE VALUES OF THE FFT TO THE ACCUMULATED STATISTICS ***  
FLAG=0  
CALL STATS (IN,REP,FLAG)  
RETURN  
END
```

BPS01610  
BPS01620  
BPS01630  
BPS01640  
BPS01650  
BPS01660  
BPS01670  
BPS01680



DBP00110  
 DBP00120  
 DBP00130  
 DBP00140  
 DBP00150  
 DBP00160  
 DBP00170  
 DBP00180  
 DBP00190  
 DBP00200  
 DBP00210  
 DBP00220  
 DBP00230  
 DBP00240  
 DBP00250  
 DBP00260  
 DBP00270  
 DBP00280  
 DBP00290  
 DBP00300  
 DBP00310  
 DBP00320  
 DBP00330  
 DBP00340  
 DBP00350  
 DBP00360  
 DBP00370  
 DBP00380  
 DBP00390  
 DBP00400  
 DBP00410  
 DBP00420  
 DBP00430  
 DBP00440  
 DBP00450  
 DBP00460  
 DBP00470  
 DBP00480  
 DBP00490  
 DBP00500  
 DBP00510  
 DBP00520  
 DBP00530  
 DBP00540  
 DBP00550  
 DBP00560  
 DBP00570  
 DBP00580  
 DBP00600

```

SUBROUTINE DBPSK (TYPE2, BAUD, FREQ, IPHAS, AMP, ANS2,
  *DSEED, IN, REP)
  *** PURPOSE ***
  THIS SUBROUTINE MODULATES THE CARRIER USING DIFFERENTIAL BINARY
  SHIFT KEYING AS THE MODULATION TECHNIQUE.
  *** PARAMETER DEFINITIONS ***
  TYPE2= INDICATES LOGIC TYPE TO BE EMPLOYED
  BAUD= SYMBOL RATE OR BAUD RATE
  FREQ= CARRIER FREQUENCY
  IPHAS= INITIAL PHASE ANGLE IN DEGREES
  AMP= AMPLITUDE
  ANS2= INTEGER VARIABLE TO BE PASSED TO STREAM GENERATOR
  DSEED= DOUBLE PRECISION SEED FOR RANDOM NUMBER GENERATOR
  IN= NUMBER OF POSITIONS OF ARRAY TO BE UTILIZED
  FLAG= THE CALL THE THIS SUBROUTINE
  AN INTEGER WHICH CONTROL ENTRANCE AND EXIT FROM
  *** VARIABLE DEFINITIONS ***
  ARRAY= A DOUBLE PRECISION ARRAY FOR PASSING THE BINARY
  AND THE FFT
  AN INTEGER USED TO REPRESENT THE ROW OF ARRAY
  TIME= TIME VARIABLE
  N= VALUE OF THE BINARY DIGIT
  STEP= INTERVAL AT WHICH THE SIGNAL IS REPRODUCED
  PI= NUMERICAL CONSTANT FREQUENCY
  OMEGA= CARRIER ANGLE IN RADIANS
  DELTA= INITIAL PHASE OF THE NUMBER OF BAUDS MODULATED
  NBAUD= A COUNT OF THE NUMBER OF BAUDS MODULATED
  BAUD= BAUD DURATION
  REF= REFERENCE VOLTAGE
  BK= VALUE OF DIFFERENTIAL BIT TO BE MODULATED
  X= VALUE OF ARRAY TO RECEIVE THE FFT OF THE TIME SERIES
  IWK= COMPLEX ARRAY USED IN SUBROUTINE FFTRC
  MAXY= MAXIMUM VALUE TO BE PLOTTED ON THE ABCISSAE
  MINY= MINIMUM VALUE TO BE PLOTTED ON THE ABCISSAE
  R= PRINCIPLE HARMONIC
  REP= AN INTEGER EQUAL TO THE NUMBER OF THE REPETITION OF
  SUBPROGRAM STATS
  IND2P1= LENGTH OF ARRAY DIVIDED BY 2 + 1
  *** VARIABLE DECLARATIONS ***
  INTEGER R, ANS2, TYPE2, IN, FMAX, REP, FLAG, IND2P1
  
```

CC

```

C      DOUBLE PRECISION BAUD, FREQ, IPHAS, AMP, TIME, MT,
*STEP, PI, OMEGA, DELTA, NBAUD, BAUDD,
*DSEED, REF, BK, MAXY, MINY, INT
C      DOUBLE PRECISION ARRAY(1024, 2), SUMX(513), SUMXSQ(513), SUMX3(513),
*SUMX4(513)
C      COMMON ARRAY, SUMX, SUMXSQ, SUMX3, SUMX4
C      COMPLEX*16 X(513)
C      INTEGER IWK(10)
C      *** VARIABLE INITIALIZATION ***
R=1
TIME=0. DO
STEP=1. DO/(2. DO*(FREQ+BAUD))
PI=3. 141592653589793D0
OMEGA=2. DO*PI*FREQ
DELTA=IPHAS*PI/180. DO
NBAUD=1. DO
BAUDD=1. DO/BAUD
REF=1. DO
MAXY=0. DO
IND2P1=IN/2+1
C      *** DO THE MODULATION AND ASSIGN THE VALUES TO ARRAY ***
CALL STREAM(DSEED, ANS2, TYPE2, MT)
BK=REF*MT
DO 10 I=1, IN
  ARRAY(R, 1)=AMP*BK*DCCS((OMEGA*TIME)+DELTA)
  ARRAY(R, 2)=TIME
  IF(ARRAY(R, 1).GT. MAXY) THEN
    MAXY=ARRAY(R, 1)
  END IF
  R=R+1
  TIME=TIME+STEP
  IF(TIME.GT. NBAUD*BAUDD) THEN
    CALL STREAM(DSEED, ANS2, TYPE2, MT)
    BK=BK*MT
    NBAUD=NBAUD+1. DO
  END IF
CONTINUE
C      *** ALLOW FOR THE PLOT OF THE TIME SERIES ***
IF(REF.EQ.1) THEN

```

```

DBP 00610
DBP 00620
DBP 00630
DBP 00640
DBP 00650
DBP 00660
DBP 00670
DBP 00680
DBP 00690
DBP 00700
DBP 00710
DBP 00720
DBP 00730
DBP 00740
DBP 00750
DBP 00760
DBP 00770
DBP 00780
DBP 00790
DBP 00800
DBP 00810
DBP 00820
DBP 00830
DBP 00840
DBP 00850
DBP 00860
DBP 00870
DBP 00880
DBP 00890
DBP 00900
DBP 00910
DBP 00920
DBP 00930
DBP 00940
DBP 00950
DBP 00960
DBP 00970
DBP 00980
DBP 00990
DBP 01000
DBP 01010
DBP 01020
DBP 01030
DBP 01040
DBP 01050
DBP 01060
DBP 01070
DBP 01080
DBP 01090
DBP 01100

```







DBP01610  
DBP01620  
DBP01630  
DBP01640  
DBP01650  
DBP01660  
DBP01670  
DBP01680  
DBP01690  
DBP01700  
DBP01710  
DBP01720  
DBP01730

```
MINY=0. D0  
INT=1.D0/STEP  
CALL PLOT(MAXY, MINY, INT, IND2P1)  
END IF  
*** ADD THE VALUES OF THE FFT TO THE ACCUMULATED STATISTICS ***  
FLAG=0  
CALL STATS(IN, REP, FLAG)  
RETURN  
END
```

C  
C  
C  
C  
C

OBP 00110  
OBP 00120  
OBP 00130  
OBP 00140  
OBP 00150  
OBP 00160  
OBP 00170  
OBP 00180  
OBP 00190  
OBP 00200  
OBP 00210  
OBP 00220  
OBP 00230  
OBP 00240  
OBP 00250  
OBP 00260  
OBP 00270  
OBP 00280  
OBP 00290  
OBP 00300  
OBP 00310  
OBP 00320  
OBP 00330  
OBP 00340  
OBP 00350  
OBP 00360  
OBP 00370  
OBP 00380  
OBP 00390  
OBP 00400  
OBP 00410  
OBP 00420  
OBP 00430  
OBP 00440  
OBP 00450  
OBP 00460  
OBP 00470  
OBP 00480  
OBP 00490  
OBP 00500  
OBP 00510  
OBP 00520  
OBP 00530  
OBP 00540  
OBP 00550  
OBP 00560  
OBP 00570  
OBP 00580  
OBP 00590  
OBP 00600

```
SUBROUTINE OBPSK (TYPE2, BAUD, BITS, FREQ, IPHAS, AMP, ANS2,  
*DSEED, IN, REP)  
** PURPOSE **  
THIS SUBROUTINE MODULATES THE CARRIER USING M-ARY PHASE SHIFT  
KEYING AS THE MODULATION TECHNIQUE.  
** PARAMETER DEFINITIONS **  
TYPE2= INDICATES LOGIC TYPE TO BE EMPLOYED  
BAUD= SYMBOL RATE OR BAUD RATE  
BITS= NUMBER OF BITS IN EACH BINARY CODE WORD  
FREQ= CARRIER FREQUENCY  
IPHAS= INITIAL PHASE ANGLE IN DEGREES  
AMP= AMPLITUDE  
ANS2= INTEGER VARIABLE TO BE PASSED TO STREAM GENERATOR  
DSEED= DOUBLE PRECISION SEED FOR RANDOM NUMBER  
IN= NUMBER OF POSITIONS IN ARRAY TO BE UTILIZED  
REP= AN INTEGER EQUAL TO THE NUMBER OF THE REPETITION OF  
THE CALL THE THIS SUBROUTINE  
** VARIABLE DEFINITIONS **  
ARRAY= A DOUBLE PRECISION ARRAY FOR PASSING THE BINARY  
DIGITS AND STORING THE VALUE OF THE TIME FUNCTION  
R= AND THE FFT  
TIME= AN INTEGER USED TO REPRESENT THE ROW OF ARRAY  
MT= TIME VARIABLE BINARY DIGIT  
STEP= VALUE OF THE BINARY DIGIT  
PI= INTERVAL AT WHICH THE SIGNAL IS REPRODUCED  
OMEGA= NUMERICAL CONSTANT  
DELTA= CARRIER ANGULAR FREQUENCY  
NBIT= INITIAL PHASE OFFSET IN RADIANS  
BITR= A COUNT OF THE NUMBER OF BITS MODULATED  
BITD= BIT RATE  
X= BIT DURATION  
IWK= COMPLEX ARRAY TO RECEIVE THE FFT OF THE TIME SERIES  
MXY= INTEGER WORKING ARRAY USED BY FUNCTION FFTRC  
MINY= MAXIMUM VALUE TO BE PLOTTED ON THE ABCISSAE  
INT= MINIMUM VALUE TO BE PLOTTED ON THE ABCISSAE  
RMAX= INTERVAL BETWEEN POINTS ON THE ORDINATE  
K= VALUE OF THE PRINCIPLE HARMONIC  
KM1= INTEGER VALUE OF THE NUMBER OF BITS IN A BINARY WORD  
N= INTEGER VALUE OF K MINUS 1  
NDP= NUMBER OF POSSIBLE BINARY CODE WORDS = 2**K  
ND2= DOUBLE PRECISION OF N  
SUM1= VARIABLY USED BY 2 TO HOLD THE DECIMAL EQUIVALENT TO A  
BINARY CODE WORD OF LENGTH K
```

CC





```

10  C C C C
SUM1=SUM1+(MT*(10.D0**KM1))
KM1=KM1-1
CONTINUE
** CONVERT THE REPRESENTATIVE BINARY VARIABLE TO ITS DECIMAL
EQUIVALENT **
SUM2=0.D0
KM1=K-1
ND2=NDP/2.D0
DC 20 I=1,K
IF(SUM1/10.D0**KM1-GE.1.D0) THEN
SUM2=SUM2+ND2
SUM1=SUM1-10.D0**KM1
ND2=ND2/2.D0
KM1=KM1-1
ELSE
ND2=ND2/2.D0
KM1=KM1-1
END IF
CONTINUE
** DO THE MODULATION AND ASSIGN TO ARRAY ***
C=1
FLAG=0
CALL CRTHO(K,MT,SUM2,C,FLAG)
FLAG=2
DO 30 I=1,IN
ARRAY(R,1)=AMP*MT*DCOS((OMEGA*TIME)+DELTA)
ARRAY(R,2)=TIME
IF(ARRAY(R,1)-GT.MAXY) THEN
MAXY=ARRAY(R,1)
END IF
R=R+1
TIME=TIME+STEP
IF(TIME-GT.NBIT*BITD) THEN
NBIT=NBIT+1.D0
IF(C.GT.N) THEN
** CONVERT A STREAM OF K BITS TO A SINGLE REPRESENTATIVE BINARY
VARIABLE IN DECIMAL FORM ***
SUM1=0.D0
KM1=K-1
DO 40 J=1,K
CALL STREAM(DSEED,ANS2,TYPE2,MT)
IF(TYPE2.EQ.1.AND.MT.EQ.-1.D0) THEN
MT=0.D0
ELSE IF(TYPE2.EQ.3.AND.MT.EQ.-1.D0) THEN

```

```

OBP01110
OBP01120
OBP01130
OBP01140
OBP01150
OBP01160
OBP01170
OBP01180
OBP01190
OBP01200
OBP01210
OBP01220
OBP01230
OBP01240
OBP01250
OBP01260
OBP01270
OBP01280
OBP01290
OBP01300
OBP01310
OBP01320
OBP01330
OBP01340
OBP01350
OBP01360
OBP01370
OBP01380
OBP01390
OBP01400
OBP01410
OBP01420
OBP01430
OBP01440
OBP01450
OBP01460
OBP01470
OBP01480
OBP01490
OBP01500
OBP01510
OBP01520
OBP01530
OBP01540
OBP01550
OBP01560
OBP01570
OBP01580
OBP01590
OBP01600

```

```

40      MT=1.D0
      END IF
      SUM1=SUM1+(MT*(10.D0**KM1))
      KM1=KM1-1
      CONTINUE

      *** CONVERT THE REPRESENTATIVE BINARY VARIABLE TO ITS DECIMAL
      EQUIVALENT ***
      SUM2=0.D0
      KM1=K-1
      ND2=NDP/2.D0
      DO 50 J=1,K
        IF (SUM1/10.D0**KM1.GE.1.D0) THEN
          SUM2=SUM2+ND2
          SUM1=SUM1-10.D0**KM1
          ND2=ND2/2.D0
          KM1=KM1-1
        ELSE
          ND2=ND2/2.D0
          KM1=KM1-1
        END IF
      END IF
      CONTINUE
      C=1
      FLAG=1
      END IF
      CALL ORTHO(K,MT,SUM2,C,FLAG)
      FLAG=2
      END IF
      CONTINUE

      *** PLOT THE TIME SERIES IF DESIRED ***
      IF (REP.EQ.1) THEN
        HINY=-MAXY
        CALL PLOT(MAXY,MINY,STEP,IN)
      END IF

      *** GENERATE THE FFT ***
      CALL FFTRC(ARRAY(1,1),IN,X,IWK)

      *** CALCULATE THE AMPLITUDE SPECTRUM OF THE FUNCTION ***
      *** FIND THE NUMBER AND VALUE OF THE PRINCIPLE HARMONIC ***
      N=0.D0
      R=1
      MAXY=0.D0
      DO 60 I=1,IND2P1

```

```

OBP01610
OBP01620
OBP01630
OBP01640
OBP01650
OBP01660
OBP01670
OBP01680
OBP01690
OBP01700
OBP01710
OBP01720
OBP01730
OBP01740
OBP01750
OBP01760
OBP01770
OBP01780
OBP01790
OBP01800
OBP01810
OBP01820
OBP01830
OBP01840
OBP01850
OBP01860
OBP01870
OBP01880
OBP01890
OBP01900
OBP01910
OBP01920
OBP01930
OBP01940
OBP01950
OBP01960
OBP01970
OBP01980
OBP01990
OBP02000
OBP02010
OBP02020
OBP02030
OBP02040
OBP02050
OBP02060
OBP02070
OBP02080
OBP02090
OBP02100

```

```

60  ARRAY(R,1)=CDABS(X(R))
CC  ARRAY(R,2)=N/STEP
CC  IF(ARRAY(R,1).GT.MAXY) THEN
CC    MAXY=ARRAY(R,1)
CC    RMAX=R-1
CC  END IF
CC  K=R+1
CC  N=N+1.DO
CC  CONTINUE
CC
69  *** DISPLAY INFO IF THE FIRST TIME THROUGH THE SUBPROGRAM ***
70  IF(REP.EQ.1) THEN
C    WRITE(10,70) RMAX
C    FORMAT(10,70) RMAX
C    * IS THE FFT HAS BEEN GENERATED! THE PRINCIPLE HARMONIC'
C    * IS THE 15 HARMONIC. THE NEXT PLOT TO BE PRODUCED WILL'
C    * BE THE AMPLITUDE SPECTRUM. ENTER A 1 IF YOU ARE READY TO'
C    * CONTINUE WITH THE PROGRAM.'
C  READ(5,*)L
C  IF(L.NE.1) THEN
C    CALL FRICMS('CLRSCRN ')
C    WRITE(10,80)
C    FORMAT(10,80)
C    GO TO 69
80  END IF
C  END IF
C  *** ALLOW FOR THE PLOT OF THE AMPLITUDE SPECTRUM OF THE FFT ***
C  IF(REP.EQ.1) THEN
C    MINY=0.DO
C    INT=1.DO/STEP
C    CALL PLOT(MAXY,MINY,INT,IND2P1)
C  END IF
C  *** ADD THE VALUES OF THE FFT TO THE ACCUMULATED STATISTICS ***
C  FLAG=0
C  CALL STATS(IN,REP,FLAG)
C  RETURN
C  END

```



QPS00110  
 QPS00120  
 QPS00130  
 QPS00140  
 QPS00150  
 QPS00160  
 QPS00170  
 QPS00180  
 QPS00190  
 QPS00200  
 QPS00210  
 QPS00220  
 QPS00230  
 QPS00240  
 QPS00250  
 QPS00260  
 QPS00270  
 QPS00280  
 QPS00290  
 QPS00300  
 QPS00310  
 QPS00320  
 QPS00330  
 QPS00340  
 QPS00350  
 QPS00360  
 QPS00370  
 QPS00380  
 QPS00390  
 QPS00400  
 QPS00410  
 QPS00420  
 QPS00430  
 QPS00440  
 QPS00450  
 QPS00460  
 QPS00470  
 QPS00480  
 QPS00490  
 QPS00500  
 QPS00510  
 QPS00520  
 QPS00530  
 QPS00540  
 QPS00550  
 QPS00560  
 QPS00570  
 QPS00580  
 QPS00590  
 QPS00600

```

SUBROUTINE QPSK(TYPE2,BAUD,BITS,FREQ,IPHAS,AMP,
*ANS2,DSEED,IN,REP)
** PURPOSE **
THIS SUBROUTINE MODULATES THE CARRIER USING QUADRATURE PHASE SHIFT
KEYING AS THE MODULATION TECHNIQUE.
** PARAMETER DEFINITIONS **
TYPE2= INDICATES LOGIC TYPE TO BE EMPLOYED
BAUD= SYMBOL RATE OR BAUD RATE
BITS= NUMBER OF BITS IN EACH BINARY CODE WORD
FREQ= CARRIER FREQUENCY
IPHAS= INITIAL PHASE ANGLE IN DEGREES
AMP= AMPLITUDE
ANS2= INTEGER VARIABLE TO BE PASSED TO STREAM GENERATOR
DSEED= DOUBLE PRECISION SEED FOR RANDOM NUMBER GENERATOR
IN= NUMBER OF POSITIONS IN ARRAY TO BE UTILIZED
REP= AN INTEGER EQUAL TO THE NUMBER OF THE REPETITION OF
THE CALL THE THIS SUBROUTINE
** VARIABLE DEFINITIONS **
ARRAY= A DOUBLE PRECISION ARRAY FOR PASSING THE BINARY
DIGITS AND STORING THE VALUE OF THE TIME FUNCTION
AND THE FFT
AN INTEGER USED TO REPRESENT THE ROW OF ARRAY
TIME VARIABLE
VALUE OF BINARY DIGIT PASSED FROM SUBPROGRAM STREAM
VALUE OF IN-PHASE BINARY DIGIT
VALUE OF QUADRATURE BINARY DIGIT
INTERVAL AT WHICH THE SIGNAL IS REPRODUCED
NUMERICAL CONSTANT FREQUENCY
CARRIER ANGLE OFFSET IN RADIANS
INITIAL PHASE OF THE NUMBER OF BAUDS MODULATED
A COUNT DURATION TO RECEIVE THE FFT OF THE TIME SERIES
COMPLEX ARRAY WORKING BY FUNCTION FFTRC
INTEGER WORKING TO BE PLOTTED ON THE ABCISSAE
MAXIMUM VALUE TO BE PLOTTED ON THE ABCISSAE
MINIMUM VALUE TO BE PLOTTED ON THE ABCISSAE
INTERVAL BETWEEN POINTS ON THE ORDINATE
VALUE OF THE PRINCIPLE HARMONIC
AN INTEGER WHICH CONTROL ENTRANCE AND EXIT FROM
SUBPROGRAM STAYS
LENGTH OF ARRAY DIVIDED BY 2 + 1
** VARIABLE DECLARATIONS **
IND2P1=
  
```

CC

```

C      INTEGER R,ANS2,TYPE2,IN,RMAX,REP,FLAG,IND2P1
C      DOUBLE PRECISION BAUD,BITS,FREQ,IPHAS,AMP,TIME,MT1,MT2,
C      *STEP,PI,OMEGA,DELTA,NEAUD,BAUDD,
C      *DSEED,MAXY,MINY,INT,MT
C      DOUBLE PRECISION ARRAY(1024,2),SUMX(513),SUMXSQ(513),SUMX3(513),
C      *SUMX4(513)
C      COMMON ARRAY,SUMX,SUMXSQ,SUMX3,SUMX4
C      COMPLEX*16 X(513)
C      INTEGER IWK(10)
C      *** VARIABLE INITIALIZATION ***
C      R=1
C      TIME=0.DO
C      STEP=1.DO/(2.DO*(FREQ+BAUD))
C      PI=3.141592653589793DO
C      OMEGA=2.DO*PI*FREQ
C      DELTA=IPHAS*PI/180.DO
C      NBAUD=1.DO
C      BAUDD=1.DO/BAUD
C      MAXY=0.DO
C      IND2P1=IN/2+1
C      *** DO THE MODULATION AND ASSIGN THE VALUES TO ARRAY***
C      CALL STKRAM(DSEED,ANS2,TYPE2,MT)
C      MT1=MT
C      CALL STREAM(DSEED,ANS2,TYPE2,MT)
C      MT2=MT
C      DO 10 I=1,IN
C      ABRAY(R,1)=AMP*MT1*DCOS((OMEGA*TIME)+DELTA)+
C      * AMP*MT2*DSIN((OMEGA*TIME)+DELTA)
C      ABRAY(R,2)=TIME
C      IF(ABRAY(R,1).GT.MAXY) THEN
C      MAXY=ABRAY(R,1)
C      END IF
C      K=R+1
C      TIME=TIME+STEP
C      IF(TIME.GT.NBAUD*BAUDD) THEN
C      CALL STREAM(DSEED,ANS2,TYPE2,MT)
C      MT1=MT
C      CALL STREAM(DSEED,ANS2,TYPE2,MT)
C      MT2=MT
C      NBAUD=NBAUD+1.DO
C      QPS00610
C      QPS00620
C      QPS00630
C      QPS00640
C      QPS00650
C      QPS00660
C      QPS00670
C      QPS00680
C      QPS00690
C      QPS00700
C      QPS00710
C      QPS00720
C      QPS00730
C      QPS00740
C      QPS00750
C      QPS00760
C      QPS00770
C      QPS00780
C      QPS00790
C      QPS00800
C      QPS00810
C      QPS00820
C      QPS00830
C      QPS00840
C      QPS00850
C      QPS00860
C      QPS00870
C      QPS00880
C      QPS00890
C      QPS00900
C      QPS00910
C      QPS00920
C      QPS00930
C      QPS00940
C      QPS00950
C      QPS00960
C      QPS00970
C      QPS00980
C      QPS00990
C      QPS01000
C      QPS01010
C      QPS01020
C      QPS01030
C      QPS01040
C      QPS01050
C      QPS01060
C      QPS01070
C      QPS01080
C      QPS01090
C      QPS01100

```



```

10      END IF
C      CONTINUE
C      *** PLOT THE TIME SERIES IF DESIRED ***
C      IF (REP.EQ.1) THEN
C          MINY=-MAXY
C          CALL PLOT(MAXY,MINY,STEP,IN)
C      END IF
C      *** GENERATE THE FFT ***
C      CALL FFTRC (ARRAY(1,1),IN,X,IWK)
C      *** CALCULATE THE AMPLITUDE SPECTRUM OF THE FUNCTION ***
C      *** FIND THE NUMBER AND VALUE OF THE PRINCIPLE HARMONIC ***
C      N=0.DO
C      K=1
C      MAXY=0.DO
C      DO 20 I=1,IN*2P1
C          ARRAY(R,1)=CDABS(X(R))
C          ARRAY(R,2)=N/STEP
C          IF (ARRAY(R,1).GT.MAXY) THEN
C              MAXY=ARRAY(R,1)
C              RMAX=R-1
C          END IF
C          R=R+1
C          N=N+1.DO
C      CONTINUE
C      *** DISPLAY INFO IF THE FIRST TIME THROUGH THE SUBPROGRAM ***
C      IF (REP.EQ.1) THEN
C          WRITE(10,30) RMAX
C          FORMAT(10,30) RMAX
C          * IS THE FFT HARMONIC. THE PRINCIPLE HARMONIC
C          * BE THE AMPLITUDE SPECTRUM. THE NEXT PLOT TO BE PRODUCED WILL
C          * CONTINUE WITH THE PROGRAM. ENTER A 1 IF YOU ARE READY TO
C          READ(5,*)L
C          IF (L.NE.1) THEN
C              CALL FRTCMS('CLRSCRN ')
C              WRITE(10,40)
C              FORMAT(10,40)
C              GO TO 29
C          END IF
40      END IF

```





SUBROUTINE OQPSK(TYPE2, BAUD, BITS, FREQ, IPHAS, AMP,  
\*ANS2, DSEED, IN, REP)

\*\*\* PURPOSE \*\*\*

THIS SUBROUTINE MODULATES THE CARRIER USING OFFSET QUADRATURE  
PHASE SHIFT KEYING AS THE MODULATION TECHNIQUE.

\*\*\* PARAMETER DEFINITIONS \*\*\*

TYPE2= INDICATES LOGIC TYPE TO BE EMPLOYED  
BAUD= SYMBOL RATE OR BAUD RATE  
BITS= NUMBER OF BITS IN EACH BINARY CODE WORD  
FREQ= CARRIER FREQUENCY  
IPHAS= INITIAL PHASE ANGLE IN DEGREES  
AMP= AMPLITUDE  
ANS2= INTEGER VARIABLE TO BE PASSED TO STREAM GENERATOR  
DSEED= DOUBLE PRECISION SEED FOR RANDOM NUMBER UTILIZED  
IN= NUMBER OF POSITIONS IN ARRAY TO BE UTILIZED  
REP= AN INTEGER EQUAL TO THE NUMBER OF THE REPETITION OF  
THE CALL THE THIS SUBROUTINE

\*\*\* VARIABLE DEFINITIONS \*\*\*

ARRAY= A DOUBLE PRECISION ARRAY FOR PASSING THE BINARY  
AND THE FFT  
AN INTEGER USED TO REPRESENT THE ROW OF ARRAY  
TIME= TIME VARIABLE  
MT= VALUE OF BINARY DIGIT PASSED FROM SUBPROGRAM STREAM  
MT1= VALUE OF IN-PHASE BINARY DIGIT  
MT2= VALUE OF QUADRATURE BINARY DIGIT  
STEP= INTERVAL AT WHICH THE SIGNAL IS REPRODUCED  
PI= NUMERICAL CONSTANT  
OMEGA= CARRIER ANGULAR FREQUENCY  
DELTA= INITIAL PHASE OFFSET IN RADIANS  
NBAUD1= A COUNT OF THE NUMBER OF BAUDS MODULATED IN-PHASE  
NBAUD2= COMPLETE WORKING TO RECEIVE THE FFT OF THE TIME SERIES  
X= INTEGER VALUE TO BE PLOTTED ON THE ABCISSAE  
IWK= MAXIMUM VALUE TO BE PLOTTED ON THE ORDINATE  
MAXY= MINIMUM VALUE BETWEEN POINTS ON THE HARMONIC  
MINTY= INTERVAL OF THE PRINT CONTROL ENTRANCE AND EXIT FROM  
RMX= AN INTEGER WHICH  
FLAG= SUBPROGRAM STAYS  
IND2P1= LENGTH OF ARRAY DIVIDED BY 2 + 1

\*\*\* VARIABLE DECLARATIONS \*\*\*

OQP00110  
OQP00120  
OQP00130  
OQP00140  
OQP00150  
OQP00160  
OQP00170  
OQP00180  
OQP00190  
OQP00200  
OQP00210  
OQP00220  
OQP00230  
OQP00240  
OQP00250  
OQP00260  
OQP00270  
OQP00280  
OQP00290  
OQP00300  
OQP00310  
OQP00320  
OQP00330  
OQP00340  
OQP00350  
OQP00360  
OQP00370  
OQP00380  
OQP00390  
OQP00400  
OQP00410  
OQP00420  
OQP00430  
OQP00440  
OQP00450  
OQP00460  
OQP00470  
OQP00480  
OQP00490  
OQP00500  
OQP00510  
OQP00520  
OQP00530  
OQP00540  
OQP00550  
OQP00560  
OQP00570  
OQP00580  
OQP00590  
OQP00600

CC

```

C      INTEGER R, ANS2, TYPE2, IN, RMAX, REP, FLAG, IND2P1
C      DOUBLE PRECISION BAUD, BITS, FREQ, IPHAS, AMP, TIME, MT1, MT2,
C      *STEP, PI, OMEGA, DELTA, BAUDD, NBAUD1, NBAUD2,
C      *DSEED, MAXY, MINY, INT, MT
C      DOUBLE PRECISION ARRAY(1024, 2), SUMX(513), SUMXSQ(513), SUMX3(513),
C      *SUMX4(513)
C      COMMON ARRAY, SUMX, SUMXSQ, SUMX3, SUMX4
C      COMPLEX*16 X(513)
C      INTEGER IWK(10)
C      *** VARIABLE INITIALIZATION ***
C      R=1
C      STEP=1. D0/( 2. D0*(FREQ+BAUD))
C      PI=3. 141592653589793D0
C      OMEGA=2. D0*PI*FREQ
C      DELTA=IPHAS*PI/180. D0
C      NBAUD1=1. D0
C      NBAUD2=1. D0/BAUD
C      BAUDD=1. D0/BAUD
C      TIME=.5D0*BAUDD
C      MAXY=0. D0
C      IND2P1=IN/2+1
C      *** DO THE MODULATION AND ASSIGN THE VALUES TO ARRAY ***
C      CALL STREAM (DSEED, ANS2, TYPE2, MT)
C      MT1=MT
C      CALL STREAM (DSEED, ANS2, TYPE2, MT)
C      MT2=MT
C      DO 10 I=1, IN
C      ARRAY(R, 1)=AMP*MT1*DCOS((OMEGA*TIME)+DELTA)+
C      *      AMP*MT2*DSIN((OMEGA*TIME)+DELTA)
C      ARRAY(R, 2)=TIME
C      IF(ARRAY(R, 1).GT. MAXY) THEN
C      END IF
C      R=R+1
C      TIME=TIME+STEP
C      IF(TIME.GT. NBAUD1*BAUDD) THEN
C      CALL STREAM(DSEED, ANS2, TYPE2, MT)
C      MT1=MT
C      NBAUD1=NBAUD1+1. D0
C      END IF
C      OQP00610
C      OQP00620
C      OQP00630
C      OQP00640
C      OQP00650
C      OQP00660
C      OQP00670
C      OQP00680
C      OQP00690
C      OQP00700
C      OQP00710
C      OQP00720
C      OQP00730
C      OQP00740
C      OQP00750
C      OQP00760
C      OQP00770
C      OQP00780
C      OQP00790
C      OQP00800
C      OQP00810
C      OQP00820
C      OQP00830
C      OQP00840
C      OQP00850
C      OQP00860
C      OQP00870
C      OQP00880
C      OQP00890
C      OQP00900
C      OQP00910
C      OQP00920
C      OQP00930
C      OQP00940
C      OQP00950
C      OQP00960
C      OQP00970
C      OQP00980
C      OQP00990
C      OQP01000
C      OQP01010
C      OQP01020
C      OQP01030
C      OQP01040
C      OQP01050
C      OQP01060
C      OQP01070
C      OQP01080
C      OQP01090
C      OQP01100

```





```

C      CALL FRTCMS('CLRSCRN ')
C      WRITE(10,40)
C      FORMAT(' ERROR',/)
C      GO TO 29
C      END IF
C      END IF
C      *** ALLOW FOR THE PLOT OF THE AMPLITUDE SPECTRUM OF THE FFT ***
C      IF (REP.EQ.1) THEN
C          MINY=0. D0
C          INT=1. D0/STEP
C          CALL PLOT(MAXY,MINY,INT,IND2P1)
C      END IF
C      *** ADD THE VALUES OF THE FFT TO THE ACCUMULATED STATISTICS ***
C      FLAG=0
C      CALL STATS(IN,REP,FLAG)
C      RETURN
C      END

```

```

00P01610
00P01620
00P01630
00P01640
00P01650
00P01660
00P01670
00P01680
00P01690
00P01700
00P01710
00P01720
00P01730
00P01740
00P01750
00P01760
00P01770
00P01780
00P01790
00P01800
00P01810
00P01820
00P01830
00P01840

```

MPS00110  
MPS00120  
MPS00130  
MPS00140  
MPS00150  
MPS00160  
MPS00170  
MPS00180  
MPS00190  
MPS00200  
MPS00210  
MPS00220  
MPS00230  
MPS00240  
MPS00250  
MPS00260  
MPS00270  
MPS00280  
MPS00290  
MPS00300  
MPS00310  
MPS00320  
MPS00330  
MPS00340  
MPS00350  
MPS00360  
MPS00370  
MPS00380  
MPS00390  
MPS00400  
MPS00410  
MPS00420  
MPS00430  
MPS00440  
MPS00450  
MPS00460  
MPS00470  
MPS00480  
MPS00490  
MPS00500  
MPS00510  
MPS00520  
MPS00530  
MPS00540  
MPS00550  
MPS00560  
MPS00570  
MPS00580  
MPS00590  
MPS00600

SUBROUTINE MPSK (TYPE2, BAUD, BITS, FREQ, IPHAS, AMP, ANS2,  
\*DSEED, IN, REP)

\*\*\* PURPOSE \*\*\*

THIS SUBROUTINE MODULATES THE CARRIER USING M-ARY PHASE SHIFT  
KEYING AS THE MODULATION TECHNIQUE.

\*\*\* PARAMETER DEFINITIONS \*\*\*

TYPE2= INDICATES LOGIC TYPE TO BE EMPLOYED  
BAUD= SYMBOL RATE OR BAUD RATE  
BITS= NUMBER OF BITS IN EACH BINARY CODE WORD  
FREQ= CARRIER FREQUENCY  
IPHAS= INITIAL PHASE ANGLE IN DEGREES  
AMP= AMPLITUDE  
ANS2= INTEGER VARIABLE TO BE PASSED TO STREAM GENERATOR  
DSEED= DOUBLE PRECISION SEED FOR RANDOM NUMBER GENERATOR  
IN= NUMBER OF POSITIONS IN ARRAY TO BE UTILIZED  
REP= AN INTEGER EQUAL TO THE NUMBER OF THE REPETITION OF  
THE CALL THE THIS SUBROUTINE

\*\*\* VARIABLE DEFINITIONS \*\*\*

ARRAY= A DOUBLE PRECISION ARRAY FOR PASSING THE BINARY  
AND THE FFT USED TO REPRESENT THE ROW OF ARRAY  
TIME= TIME VARIABLE  
STEP= VALUE OF THE BINARY DIGIT  
PI= INTERVAL AT WHICH THE SIGNAL IS REPRODUCED  
OMEGA= NUMERICAL CONSTANT FREQUENCY  
DELTA= CARRIER ANGLE OFFSET IN RADIANS  
NBAUD= INITIAL PHASE OF THE NUMBER OF BAUDS MODULATED  
X= BAUD DURATION  
IWK= COMPLEX ARRAY TO RECEIVE THE FFT OF THE TIME SERIES  
MAXY= INTEGER WORKING ARRAY USED BY FUNCTION FFTRC  
MINY= MAXIMUM VALUE TO BE PLOTTED ON THE ABCISSAE  
INT= MINIMUM VALUE TO BE PLOTTED ON THE ABCISSAE  
KMAX= INTERVAL BETWEEN POINTS ON THE ORDINATE  
KM1= VALUE OF THE PRINCIPLE HARMONIC  
N= INTEGER VALUE OF THE NUMBER OF BITS IN A BINARY WORD  
ND2= NUMBER OF POSSIBLE BINARY CODE WORDS = 2\*\*K  
SUM1= VARIABLE USED TO HOLD THE DECIMAL EQUIVALENT TO A  
SUM2= BINARY CODE WORD OF LENGTH K  
J= VARIABLE USED TO HOLD THE DECIMAL CONVERSION OF SUM1  
INTEGER USED TO HOLD THE DECIMAL CONVERSION OF SUM1

CC





```

10      KM1=KM1-1
      CONTINUE
      *** CONVERT THE REPRESENTATIVE BINARY VARIABLE TO ITS DECIMAL
      EQUIVALENT ***
      SUM2=0, DO
      KM1=K-1
      ND2=N/2, DO
      DO 20 I=1, K
      IF (SUM1/10, DO**KM1, GE, 1, DO) THEN
      SUM2=SUM2+ND2
      SUM1=SUM1-10, DO**KM1
      ND2=ND2/2, DO
      KM1=KM1-1
      ELSE
      ND2=ND2/2, DO
      KM1=KM1-1
      END IF
      CONTINUE
      *** DO THE MODULATION AND ASSIGN TO ARRAY ***
      J=R
      DO 30 I=J, IN
      ARRAY(R, 1) = AMP*DCOS((OMEGA*TIME)+
      ((SUM2*2, DO*PI)/N)+DELTA)
      ARRAY(R, 2) = TIME
      IF (ARRAY(R, 1) .GT. MAXY) THEN
      MAXY=ARRAY(R, 1)
      END IF
      R=R+1
      TIME=TIME+STEP
      IF (TIME.GT. NBAUD*NBAUD) THEN
      NBAUD=NBAUD+1, DO
      GO TO 9
      END IF
      CONTINUE
      END IF
      *** PLOT THE TIME SERIES IF DESIRED ***
      IF (REP, EQ, 1) THEN
      MINY=-MAXY
      CALL PLOT(MAXY, MINY, STEP, IN)
      END IF
      *** GENERATE THE FFT ***

```

```

MPS 011110
MPS 011120
MPS 011130
MPS 011140
MPS 011150
MPS 011160
MPS 011170
MPS 011180
MPS 011190
MPS 011200
MPS 011210
MPS 011220
MPS 011230
MPS 011240
MPS 011250
MPS 011260
MPS 011270
MPS 011280
MPS 011290
MPS 011300
MPS 011310
MPS 011320
MPS 011330
MPS 011340
MPS 011350
MPS 011360
MPS 011370
MPS 011380
MPS 011390
MPS 011400
MPS 011410
MPS 011420
MPS 011430
MPS 011440
MPS 011450
MPS 011460
MPS 011470
MPS 011480
MPS 011490
MPS 011500
MPS 011510
MPS 011520
MPS 011530
MPS 011540
MPS 011550
MPS 011560
MPS 011570
MPS 011580
MPS 011600

```

```

C      CALL FFTRC(ARRAY(1,1),IN,X,IWK)
C      *** CALCULATE THE AMPLITUDE SPECTRUM OF THE FUNCTION ***
C      *** FIND THE NUMBER AND VALUE OF THE PRINCIPLE HARMONIC ***
N=0.0
R=1
MAXY=0.0
DO 4 I=1,IND2P1
  ARRAY(R,1)=CDABS(X(R))
  ARRAY(R,2)=N/STEP
  IF(ARRAY(R,1).GT.MAXY) THEN
    MAXY=ARRAY(R,1)
    RMAX=R-1
  END IF
  R=R+1
  N=N+1
DO
CONTINUE
*** DISPLAY INFO IF THE FIRST TIME THROUGH THE SUBPROGRAM ***
IF(REP.EQ.1) THEN
WRITE(10,50) RMAX
FORMAT(' THE FFT HAS BEEN GENERATED! THE PRINCIPLE HARMONIC',
// ' IS THE ',I5,' HARMONIC. THE NEXT PLOT TO BE PRODUCED WILL',
// ' BE THE AMPLITUDE SPECTRUM. ENTER A 1 IF YOU ARE READY TO',
// ' CONTINUE WITH THE PROGRAM.'//)
READ(5,*)L
IF(L.NE.1) THEN
  CALL FRTCMS('CLRSCRN ')
  WRITE(10,60)
  FORMAT(' ERROR'//)
  GO TO 49
END IF
END IF
*** ALLOW FOR THE PLOT OF THE AMPLITUDE SPECTRUM OF THE FFT ***
IF(REP.EQ.1) THEN
  MINY=0.0
  INT=1.0/STEP
  CALL PLOT(MAXY,MINY,INT,IND2P1)
END IF
*** ADD THE VALUES OF THE FFT TO THE ACCUMULATED STATISTICS ***

```

```

MPS01610
MPS01620
MPS01630
MPS01640
MPS01650
MPS01660
MPS01670
MPS01680
MPS01690
MPS01700
MPS01710
MPS01720
MPS01730
MPS01740
MPS01750
MPS01760
MPS01770
MPS01780
MPS01790
MPS01800
MPS01810
MPS01820
MPS01830
MPS01840
MPS01850
MPS01860
MPS01870
MPS01880
MPS01890
MPS01900
MPS01910
MPS01920
MPS01930
MPS01940
MPS01950
MPS01960
MPS01970
MPS01980
MPS01990
MPS02000
MPS02010
MPS02020
MPS02030
MPS02040
MPS02050
MPS02060
MPS02070
MPS02080
MPS02090
MPS02100

```



C

FLAG=0

CALL STATS(IN,REP,FLAG)

C

RETURN

END

MPS02110  
MPS02120  
MPS02130  
MPS02140  
MPS02150  
MPS02160







MAS011110  
 MAS011120  
 MAS011130  
 MAS011140  
 MAS011150  
 MAS011160  
 MAS011170  
 MAS011180  
 MAS011190  
 MAS011200  
 MAS01210  
 MAS01220  
 MAS01230  
 MAS01240  
 MAS01250  
 MAS01260  
 MAS01270  
 MAS01280  
 MAS01290  
 MAS01300  
 MAS01310  
 MAS01320  
 MAS01330  
 MAS01340  
 MAS01350  
 MAS01360  
 MAS01370  
 MAS01380  
 MAS01390  
 MAS01400  
 MAS01410  
 MAS01420  
 MAS01430  
 MAS01440  
 MAS01450  
 MAS01460  
 MAS01470  
 MAS01480  
 MAS01490  
 MAS01500  
 MAS01510  
 MAS01520  
 MAS01530  
 MAS01540  
 MAS01550  
 MAS01560  
 MAS01570  
 MAS01580  
 MAS01590  
 MAS01600

```

END IF
SUM1=SUM1+(MT*(10.D0**KM1))
KM1=KM1-1
CONTINUE

*** CONVERT THE REPRESENTATIVE BINARY VARIABLE TO ITS DECIMAL
EQUIVALENT ***
SUM2=0.D0
KM1=K-1
ND2=N/2.D0
DO 20 I=1,K
  IF(SUM1/10.D0**KM1.GE.1.D0) THEN
    SUM2=SUM2+ND2
    SUM1=SUM1-10.D0**KM1
    ND2=ND2/2.D0
    KM1=KM1-1
  ELSE
    ND2=ND2/2.D0
    KM1=KM1-1
  END IF
CONTINUE

*** DO THE MODULATION AND ASSIGN TO ARRAY ***
SUM2=SUM2+1.D0
AMP1=AMP/SUM2
J=R
DO 30 I=J,IN
  ARRAY(R,2)=AMP1*DCOS((OMEGA*TIME)+DELTA)
  ARRAY(R,1)=TIME
  IF(ARRAY(R,1).GT.MAXY) THEN
    MAXY=ARRAY(R,1)
  END IF
  R=R+1
  TIME=TIME+STEP
  IF(TIME.GT.NBAUD*NBAUD) THEN
    NBAUD=NBAUD+1.D0
    GO TO 9
  END IF
CONTINUE
END IF

*** PLOT THE TIME SERIES IF DESIRED ***
IF(REP.EQ.1) THEN
  MINY=-MAXY
  CALL PLOT(MAXY,MINY,STEP,IN)
END IF

```

10  
 C  
 C  
 C  
 C

20  
 C  
 C  
 C

30  
 C  
 C  
 C  
 C



MAS02110  
MAS02120  
MAS02130  
MAS02140  
MAS02150  
MAS02160  
MAS02170  
MAS02180  
MAS02190

END IF  
\*\* ADD THE VALUES OF THE FFT TO THE ACCUMULATED STATISTICS \*\*  
FLAG=0  
CALL STATS(IN,REP,FLAG)  
RETURN  
END

C  
C  
C  
C



QAS 00110  
 QAS 00120  
 QAS 00130  
 QAS 00140  
 QAS 00150  
 QAS 00160  
 QAS 00170  
 QAS 00180  
 QAS 00190  
 QAS 00200  
 QAS 00210  
 QAS 00220  
 QAS 00230  
 QAS 00240  
 QAS 00250  
 QAS 00260  
 QAS 00270  
 QAS 00280  
 QAS 00290  
 QAS 00300  
 QAS 00310  
 QAS 00320  
 QAS 00330  
 QAS 00340  
 QAS 00350  
 QAS 00360  
 QAS 00370  
 QAS 00380  
 QAS 00390  
 QAS 00400  
 QAS 00410  
 QAS 00420  
 QAS 00430  
 QAS 00440  
 QAS 00450  
 QAS 00460  
 QAS 00470  
 QAS 00480  
 QAS 00490  
 QAS 00500  
 QAS 00510  
 QAS 00520  
 QAS 00530  
 QAS 00540  
 QAS 00550  
 QAS 00560  
 QAS 00570  
 QAS 00580  
 QAS 00590  
 QAS 00600

```

SUBROUTINE QASK (TYPE2, BAUD, BITS, FREQ, IPHAS, AMP, ANS2,
*DSEED, IN, REP)
*** PURPOSE ***
THIS SUBROUTINE MODULATES THE CARRIER USING QUADRATURE AMPLITUDE
SHIFT KEYING AS THE MODULATION TECHNIQUE.
*** PARAMETER DEFINITIONS ***
TYPE2= INDICATES LOGIC TYPE TO BE EMPLOYED
BAUD= SYMBOL RATE OR BAUD RATE
BITS= NUMBER OF BITS IN EACH BINARY CODE WORD
FREQ= CARRIER FREQUENCY
IPHAS= INITIAL PHASE ANGLE IN DEGREES
AMP= AMPLITUDE VARIABLE TO BE PASSED TO STREAM
ANS2= INTEGER PRECISION SEED FOR RANDOM NUMBER GENERATOR
DSEED= DOUBLE PRECISION SEED FOR RANDOM NUMBER GENERATOR
IN= NUMBER OF POSITIONS IN ARRAY TO BE UTILIZED
FLAG= THE CALL THE THIS SUBROUTINE
      AN INTEGER WHICH CONTROL ENTRANCE AND EXIT FROM
*** VARIABLE DEFINITIONS ***
ARRAY= A DOUBLE PRECISION ARRAY FOR PASSING THE BINARY
      DIGITS AND STORING THE VALUE OF THE TIME FUNCTION
      AND THE FFT
      AN INTEGER USED TO REPRESENT THE ROW OF ARRAY
      TIME VARIABLE
      VALUE OF THE BINARY DIGIT
      INTERVAL AT WHICH THE SIGNAL IS REPRODUCED
      NUMERICAL CONSTANT FREQUENCY
      CARRIER ANGLE OFFSET IN RADIANS
      INITIAL PHASE OF THE NUMBER OF BAUDS MODULATED
      BAUD DURATION
      COMPLEX ARRAY TO RECEIVE THE FFT OF THE TIME SERIES
      INTEGER WORKING ARRAY USED BY FUNCTION FFTRC
      MAXIMUM VALUE TO BE PLOTTED ON THE ABCISSAE
      MINIMUM VALUE TO BE PLOTTED ON THE ABCISSAE
      INTERVAL BETWEEN POINTS ON THE ORDINATE
      VALUE OF THE PRINTING NUMBER OF BITS IN A BINARY WORD
      INTEGER VALUE OF K MINUS 1
      NUMBER OF POSSIBLE BINARY CODE WORDS = 2**K
      N DIVIDED BY 2
      VARIABLE USED TO HOLD THE DECIMAL EQUIVALENT TO A
      BINARY CODE WORD OF LENGTH K
      VARIABLE USED TO HOLD THE DECIMAL CONVERSION OF SUM1
      INTEGER USED TO INDEX A LOOP
  
```

CC





QAS011110  
 QAS011120  
 QAS011130  
 QAS011140  
 QAS011150  
 QAS011160  
 QAS011170  
 QAS011180  
 QAS011190  
 QAS011200  
 QAS011210  
 QAS011220  
 QAS011230  
 QAS011240  
 QAS011250  
 QAS011260  
 QAS011270  
 QAS011280  
 QAS011290  
 QAS011300  
 QAS011310  
 QAS011320  
 QAS011330  
 QAS011340  
 QAS011350  
 QAS011360  
 QAS011370  
 QAS011380  
 QAS011390  
 QAS011400  
 QAS011410  
 QAS011420  
 QAS011430  
 QAS011440  
 QAS011450  
 QAS011460  
 QAS011470  
 QAS011480  
 QAS011490  
 QAS011500  
 QAS011510  
 QAS011520  
 QAS011530  
 QAS011540  
 QAS011550  
 QAS011560  
 QAS011570  
 QAS011580  
 QAS011590  
 QAS011600

```

SUM1=0.D0
KM1=K-1
DO 10 I=1,K
  TRNAM(DSEED,ANS2,TYPE2,MT)
  CALL TYPE2.EQ.1.AND.MT.EQ.-1.D0) THEN
  ELSE IF (TYPE2.EQ.3.AND.MT.EQ.-1.D0) THEN
    MT=1.D0
  END IF
  SUM1=SUM1+(MT*(10.D0**KM1))
  KM1=KM1-1
CONTINUE

*** CONVERT THE REPRESENTATIVE BINARY VARIABLE TO ITS DECIMAL
EQUIVALENT ***

SUM2=0.D0
KM1=K-1
ND2=N/2.D0
DO 20 I=1,K
  IF (SUM1/10.D0**KM1.GE.1.D0) THEN
    SUM2=SUM2+ND2
    SUM1=SUM1-10.D0**KM1
    ND2=ND2/2.D0
    KM1=KM1-1
  ELSE
    ND2=ND2/2.D0
    KM1=KM1-1
  END IF
CONTINUE

IF (M.EQ.1) THEN SUM2+1.D0
ELSE IF (M.EQ.2) THEN
  SUM2B=SUM2+1.D0
END IF

CONTINUE

*** DO THE MODULATION AND ASSIGN TO ARRAY ***

AMPA=AMP/SUM2A
AMPB=AMP/SUM2B
DO 30 I=J,IN
  ARRAY(R,1)=AMPA*DCOS((OMEGA*TIME)+DELTA)+
             AMPB*DSIN((OMEGA*TIME)+DELTA)
  ARRAY(R,2)=TIME
  IF (ARRAY(R,1).GT.MAXY) THEN

```

C

10  
C  
C  
C  
C

20  
C

16  
C  
C  
C

\*







MSK00110  
 MSK00120  
 MSK00130  
 MSK00140  
 MSK00150  
 MSK00160  
 MSK00170  
 MSK00180  
 MSK00190  
 MSK00200  
 MSK00210  
 MSK00220  
 MSK00230  
 MSK00240  
 MSK00250  
 MSK00260  
 MSK00270  
 MSK00280  
 MSK00290  
 MSK00300  
 MSK00310  
 MSK00320  
 MSK00330  
 MSK00340  
 MSK00350  
 MSK00360  
 MSK00370  
 MSK00380  
 MSK00390  
 MSK00400  
 MSK00410  
 MSK00420  
 MSK00430  
 MSK00440  
 MSK00450  
 MSK00460  
 MSK00470  
 MSK00480  
 MSK00490  
 MSK00500  
 MSK00510  
 MSK00520  
 MSK00530  
 MSK00540  
 MSK00550  
 MSK00560  
 MSK00570  
 MSK00580  
 MSK00600

```

SUBROUTINE MSK(TYPE2, BAUD, FREQ, IPHAS, AMP, ANS2,
*DSEED, IN, REP)
*** PURPOSE ***
THIS SUBROUTINE MODULATES THE CARRIER USING MINIMUM SHIFT KEYING
OR FAST FREQUENCY SHIFT KEYING AS THE MODULATION TECHNIQUE.
*** PARAMETER DEFINITIONS ***
TYPE2= INDICATES LOGIC TYPE TO BE EMPLOYED
BAUD= SYMBOL RATE OR BAUD RATE
FREQ= CARRIER FREQUENCY
IPHAS= INITIAL PHASE ANGLE IN DEGREES
AMP= AMPLITUDE
ANS2= INTEGER VARIABLE TO BE PASSED TO STREAM GENERATOR
DSEED= DOUBLE PRECISION SEED FOR RANDOM NUMBER GENERATOR
IN= NUMBER OF POSITIONS IN ARRAY TO BE UTILIZED
REP= AN INTEGER EQUAL TO THE NUMBER OF THE REPETITION OF
THE CALL THE THIS SUBROUTINE
*** VARIABLE DEFINITIONS ***
ARRAY= A DOUBLE PRECISION ARRAY FOR PASSING THE BINARY
DIGITS AND STORING THE VALUE OF THE TIME FUNCTION
AND THE FFT
R= AN INTEGER USED TO REPRESENT THE ROW OF ARRAY
TIME= TIME VARIABLE
MSTEP= VALUE OF THE BINARY DIGIT
PI= INTERVAL AT WHICH THE SIGNAL IS REPRODUCED
OMEGA= CARRIER ANGULAR FREQUENCY
DELTA= INITIAL PHASE OFFSET IN RADIAN
NBAUD= A COUNT OF THE NUMBER OF BAUDS MODULATED
BAUDD= BAUD DURATION
X= COMPLEX ARRAY TO RECEIVE THE FFT OF THE TIME SERIES
IWK= INTEGER WORKING ARRAY USED BY FUNCTION FFTRC
MAXY= MAXIMUM VALUE TO BE PLOTTED ON THE ABCISSAE
MINT= MINIMUM VALUE TO BE PLOTTED ON THE ABCISSAE
INT= INTERVAL BETWEEN POINTS ON THE ORDINATE
RMAX= VALUE OF THE PRINCIPLE HARMONIC
FLAG= AN INTEGER WHICH CONTROL ENTRANCE AND EXIT FROM
SUBPROGRAM STATS
IND2P1= LENGTH OF ARRAY DIVIDED BY 2 + 1
*** VARIABLE DECLARATIONS ***
INTEGER R, ANS2, TYPE2, IN, RMAX, REP, FLAG, IND2P1
DOUBLE PRECISION BAUD, FREQ, IPHAS, AMP, TIME, MT,

```

CC



```

C      *STEP,PI,OMEGA,DELTA,NBAUD,BAUDD,
C      *DSEED,MAXY,MINY,INT
C      DOUBLE PRECISION ARRAY(1024,2),SUMX(513),SUMXSQ(513),SUMX3(513),
C      *SUMX4(513)
C      COMMON ARRAY,SUMX,SUMXSQ,SUMX3,SUMX4
C      COMPLEX*16 X(513)
C      INTEGER IWK(10)
C      *** VARIABLE INITIALIZATION ***
C      R=1
C      TIME=0.D0
C      STEP=1.D0/(2.D0*(FREQ+(1.25D0*BAUD)))
C      PI=3.141592653589793D0
C      OMEGA=2.D0*PI*FREQ
C      DELTA=IPHAS*PI/180.D0
C      NBAUD=1.D0
C      BAUDD=1.D0/BAUD
C      MAXY=0.D0
C      IINDZP1=IN/2+1
C      CALL STREAM(DSEED,ANS2,TYPE2,MT)
C
C      DO 10 I=1,IN
C      ARRAY(R,1)=AMP*DCOS(((OMEGA+((MT*PI)/(2.D0*BAUDD)))*TIME)
C      +DELTA)
C      ARRAY(R,2)=TIME
C      IF(DABS(ARRAY(R,1)).GT.MAXY) THEN
C      MAXY=DABS(ARRAY(R,1))
C      END IF
C      R=R+1
C      TIME=TIME+STEP
C      IF(TIME.GT.NBAUD*BAUDD) THEN
C      CALL STREAM(DSEED,ANS2,TYPE2,MT)
C      NBAUD=NBAUD+1.D0
C      END IF
C      CONTINUE
C      *** PLOT THE TIME SERIES IF DESIRED ***
C      IF(REP.EQ.1) THEN
C      MINY=-MAXY
C      CALL PLOT(MAXY,MINY,STEP,IN)
C      END IF
C      *** GENERATE THE FFT ***
MSK00610
MSK00620
MSK00630
MSK00640
MSK00650
MSK00660
MSK00670
MSK00680
MSK00690
MSK00700
MSK00710
MSK00720
MSK00730
MSK00740
MSK00750
MSK00760
MSK00770
MSK00780
MSK00790
MSK00800
MSK00810
MSK00820
MSK00830
MSK00840
MSK00850
MSK00860
MSK00870
MSK00880
MSK00890
MSK00900
MSK00910
MSK00920
MSK00930
MSK00940
MSK00950
MSK00960
MSK00970
MSK00980
MSK00990
MSK01000
MSK01010
MSK01020
MSK01030
MSK01040
MSK01050
MSK01060
MSK01070
MSK01080
MSK01090
MSK01100

```

MSK011110  
 MSK011120  
 MSK011130  
 MSK011140  
 MSK011150  
 MSK011160  
 MSK011170  
 MSK011180  
 MSK011190  
 MSK011200  
 MSK011210  
 MSK011220  
 MSK011230  
 MSK011240  
 MSK011250  
 MSK011260  
 MSK011270  
 MSK011280  
 MSK011290  
 MSK011300  
 MSK011310  
 MSK011320  
 MSK011330  
 MSK011340  
 MSK011350  
 MSK011360  
 MSK011370  
 MSK011380  
 MSK011390  
 MSK011400  
 MSK011410  
 MSK011420  
 MSK011430  
 MSK011440  
 MSK011450  
 MSK011460  
 MSK011470  
 MSK011480  
 MSK011490  
 MSK011500  
 MSK011510  
 MSK011520  
 MSK011530  
 MSK011540  
 MSK011550  
 MSK011560  
 MSK011570  
 MSK011580  
 MSK011600

```

C      CALL FFTRC(ARRAY(1,1),IN,X,IWK)
C
C      *** CALCULATE THE AMPLITUDE SPECTRUM OF THE FUNCTION ***
C      *** FIND THE NUMBER AND VALUE OF THE PRINCIPLE HARMONIC ***
C
C      N=0. DO
C      R=1
C      MAXY=0. DO
C      DO 20 I=1,IND2P1
C      ARRAY(R,1)=CDABS(X(R))
C      ARRAY(R,2)=N/STEP
C      IF(ARRAY(R,1).GT.MAXY) THEN
C      MAXY=ARRAY(R,1)
C      RMAX=R-1
C      END IF
C      R=R+1
C      N=N+1. DO
C      CONTINUE
C
C      *** DISPLAY INFO IF THE FIRST TIME THROUGH SUBPROGRAM ***
C      IF(REP.EQ.1) THEN
C      WRITE(10,30)RMAX
C      FORMAT('THE FFT HAS BEEN GENERATED! THE PRINCIPLE HARMONIC',//
C      *' IS THE',I5,' HARMONIC. THE NEXT PLOT TO BE PRODUCED WILL',//
C      *' BE THE AMPLITUDE SPECTRUM. ENTER A 1 IF YOU ARE READY TO',//
C      *' CONTINUE WITH THE PROGRAM.',//,' :')
C      READ(5,*)L
C      IF(L.NE.1) THEN
C      CALL FRTCMS('CLRSCRN ')
C      WRITE(10,40)
C      FORMAT('ERROR',/)
C      GO TO 29
C      END IF
C      END IF
C
C      *** ALLOW FOR THE PLOT OF THE AMPLITUDE SPECTRUM OF THE FFT ***
C      IF(REP.EQ.1) THEN
C      MINY=0. DO
C      INT=1. DO/STEP
C      CALL PLOT(MAXY,MINY,INT,IND2P1)
C      END IF
  
```

MSK01610  
MSK01620  
MSK01630  
MSK01640  
MSK01650  
MSK01660  
MSK01670

\*\*\* ADD THE VALUES OF THE FFT TO THE ACCUMULATED STATISTICS \*\*\*

FLAG=0  
CALL STATS(IN,REP,FLAG)  
RETURN  
END

C  
C  
C



SUBROUTINE MFSK (TYPE2, BAUD, BITS, FREQ, IPHAS, AMP, ANS2,  
\*DSEED, IN, REP)

\*\*\* PURPOSE \*\*\*

THIS SUBROUTINE MODULATES THE CARRIER USING M-ARY FREQUENCY SHIFT  
KEYING AS THE MODULATION TECHNIQUE.

\*\*\* PARAMETER DEFINITIONS \*\*\*

TYPE2= INDICATES LOGIC TYPE TO BE EMPLOYED  
BAUD= SYMBOL RATE OR BAUD RATE  
BITS= NUMBER OF BITS IN EACH BINARY CODE WORD  
FREQ= CARRIER FREQUENCY  
IPHAS= INITIAL PHASE ANGLE IN DEGREES  
AMP= AMPLITUDE  
ANS2= INTEGER VARIABLE TO BE PASSED TO STREAM GENERATOR  
DSEED= DOUBLE PRECISION SEED FOR RANDOM NUMBER UTILIZED  
IN= NUMBER OF POSITIONS IN ARRAY TO BE UTILIZED  
REP= AN INTEGER EQUAL TO THE NUMBER OF THE REPETITION OF  
THE CALL THIS SUBROUTINE

\*\*\* VARIABLE DEFINITIONS \*\*\*

ARRAY= A DOUBLE PRECISION ARRAY FOR PASSING THE BINARY  
DIGITS AND STORING THE VALUE OF THE TIME FUNCTION  
R= AN INTEGER USED TO REPRESENT THE ROW OF ARRAY  
TIME= TIME VARIABLE  
MT= VALUE OF THE BINARY DIGIT  
STEP= INTERVAL AT WHICH THE SIGNAL IS REPRODUCED  
PI= NUMERICAL CONSTANT  
OMEGA= CARRIER ANGLE OFFSET IN RADIANS  
DELTA= INITIAL PHASE OFFSET  
NBAUD= A COUNT OF THE NUMBER OF BAUDS MODULATED  
BAUDD= BAUD DURATION  
X= COMPLEX ARRAY TO RECEIVE THE FFT OF THE TIME SERIES  
IWK= INTEGER WORKING ARRAY USED BY FFTC  
MAXY= MAXIMUM VALUE TO BE PLOTTED ON THE ABCISSAE  
MINY= MINIMUM VALUE TO BE PLOTTED ON THE ABCISSAE  
INT= INTERVAL BETWEEN POINTS ON THE ORDINATE  
RMAX= VALUE OF THE PRINCIPLE HARMONIC  
K= INTEGER VALUE OF THE NUMBER OF BITS IN A BINARY WORD  
KM1= NUMBER OF POSSIBLE BINARY CODE WORDS = 2\*\*K  
N= N DIVIDED BY 2  
ND2= VARIABLY USED TO HOLD THE DECIMAL EQUIVALENT TO A  
SUM1= BINARY CODE WORD OF LENGTH K  
SUM2= VARIABLY USED TO HOLD THE DECIMAL CONVERSION OF SUM1  
J= INTEGER USED TO HOLD THE DECIMAL CONVERSION OF SUM1

MFS000110  
MFS000120  
MFS000130  
MFS000140  
MFS000150  
MFS000160  
MFS000170  
MFS000180  
MFS000190  
MFS000200  
MFS000210  
MFS000220  
MFS000230  
MFS000240  
MFS000250  
MFS000260  
MFS000270  
MFS000280  
MFS000290  
MFS000300  
MFS000310  
MFS000320  
MFS000330  
MFS000340  
MFS000350  
MFS000360  
MFS000370  
MFS000380  
MFS000390  
MFS000400  
MFS000410  
MFS000420  
MFS000430  
MFS000440  
MFS000450  
MFS000460  
MFS000470  
MFS000480  
MFS000490  
MFS000500  
MFS000510  
MFS000520  
MFS000530  
MFS000540  
MFS000550  
MFS000560  
MFS000570  
MFS000580  
MFS000590  
MFS000600

CC





```

MFS011110
MFS011120
MFS011130
MFS011140
MFS011150
MFS011160
MFS011170
MFS011180
MFS011190
MFS011200
MFS011210
MFS011220
MFS011230
MFS011240
MFS011250
MFS011260
MFS011270
MFS011280
MFS011290
MFS011300
MFS011310
MFS011320
MFS011330
MFS011340
MFS011350
MFS011360
MFS011370
MFS011380
MFS011390
MFS011400
MFS011410
MFS011420
MFS011430
MFS011440
MFS011450
MFS011460
MFS011470
MFS011480
MFS011490
MFS011500
MFS011510
MFS011520
MFS011530
MFS011540
MFS011550
MFS011560
MFS011570
MFS011580
MFS011590
MFS011600

DO 10 I=1,K
CALL STREAM(DSEED,ANS2,TYPE2,MT)
IF (TYPE2.EQ.1.AND.MT.EQ.-1.D0) THEN
ELSE IF (TYPE2.EQ.3.AND.MT.EQ.-1.D0) THEN
MT=1.D0
END IF
SUM1=SUM1+(MT*(10.D0**KM1))
KM1=KM1-1
CONTINUE

*** CONVERT THE REPRESENTATIVE BINARY VARIABLE TO ITS DECIMAL
EQUIVALENT ***
SUM2=0.D0
KM1=K-1
ND2=N/2.D0
DO 20 I=1,K
IF (SUM1/10.D0**KM1.GE.1.D0) THEN
SUM2=SUM2+ND2
SUM1=SUM1-10.D0**KM1
NL2=ND2/2.D0
KM1=KM1-1
ELSE
ND2=ND2/2.D0
KM1=KM1-1
END IF
CONTINUE

*** DO THE MODULATION AND ASSIGN TO ARRAY ***
J=R
MREQ=-FRNGD2+(SUM2*DELF)
DO 30 I=J,IN
ARRAY(R,1)=AMP*DCOS(((OMEGA+(2.D0*PI*MREQ))*TIME)+
DELTA)
ARRAY(R,2)=TIME
IF (ARRAY(R,1).GT.MAXY) THEN
MAXY=ARRAY(R,1)
END IF
R=R+1
TIME=TIME+STEP
IF (TIME.GT.NBAUD*BAUDD) THEN
NBAUD=NBAUD+1.D0
GO TO 9
END IF
CONTINUE
END IF

*** PLOT THE TIME SERIES IF DESIRED ***

```

10  
C  
C  
C  
C

20  
C  
C  
C

30  
C  
C



```

C      IF (REP.EQ.1) THEN
C      MINY=-MAXY
C      CALL PLOT(MAXY,MINY,STEP,IN)
C      END IF
C      *** GENERATE THE FFT ***
C      CALL FFTRC (ARRAY(1,1),IN,X,IWK)
C      *** CALCULATE THE AMPLITUDE SPECTRUM OF THE FUNCTION ***
C      *** FIND THE NUMBER AND VALUE OF THE PRINCIPLE HARMONIC ***
N=0.DO
R=1
MAXY=0.DO
DO 40 I=1,IND2P1
  ARRAY(R,1)=CDABS(X(R))
  ARRAY(R,2)=N/STEP
  IF (ARRAY(R,1).GT.MAXY) THEN
    MAXY=ARRAY(R,1)
    RMAX=R-1
  END IF
  R=R+1
  N=N+1.DO
CONTINUE
40
C      *** DISPLAY INFO IF THE FIRST TIME THROUGH THE SUBPROGRAM ***
C      IF (REP.EQ.1) THEN
C      WRITE(10,50) RMAX
C      FORMAT(10,50) THE FFT HAS BEEN GENERATED! THE PRINCIPLE HARMONIC
C      * IS THE ,I5, HARMONIC. THE NEXT PLOT TO BE PRODUCED WILL
C      ** BE THE AMPLITUDE SPECTRUM. ENTER A 1 IF YOU ARE READY TO
C      ** CONTINUE WITH THE PROGRAM. ,//, ://
C      READ(5,*)L
C      IF(L.NE.1) THEN
C      CALL FRTCMS('CLRSCRN ')
C      WRITE(10,60)
C      FORMAT('ERROR',/)
C      GO TO 49
60
C      END IF
C      END IF
C      *** ALLOW FOR THE PLOT OF THE AMPLITUDE SPECTRUM OF THE FFT ***
MFS01610
MFS01620
MFS01630
MFS01640
MFS01650
MFS01660
MFS01670
MFS01680
MFS01690
MFS01700
MFS01710
MFS01720
MFS01730
MFS01740
MFS01750
MFS01760
MFS01770
MFS01780
MFS01790
MFS01800
MFS01810
MFS01820
MFS01830
MFS01840
MFS01850
MFS01860
MFS01870
MFS01880
MFS01890
MFS01900
MFS01910
MFS01920
MFS01930
MFS01940
MFS01950
MFS01960
MFS01970
MFS01980
MFS01990
MFS02000
MFS02010
MFS02020
MFS02030
MFS02040
MFS02050
MFS02060
MFS02070
MFS02080
MFS02090
MFS02100

```

```

C      IF (REP.EQ.1) THEN
C          MINY=0. D0
C          INT=1. D0/STEP
C          CALL PLOT(MAXY, MINY, INT, IND2P1)
C          END IF
C          FLAG=0
C          CALL STATS(IN, REP, FLAG)
C          RETURN
C          END

```

```

MFS02110
MFS02120
MFS02130
MFS02140
MFS02150
MFS02160
MFS02170
MFS02180
MFS02190
MFS02200
MFS02210
MFS02220
MFS02230

```

QPR001110  
 QPR001120  
 QPR001130  
 QPR001140  
 QPR001150  
 QPR001160  
 QPR001170  
 QPR001180  
 QPR001190  
 QPR001200  
 QPR001210  
 QPR001220  
 QPR001230  
 QPR001240  
 QPR001250  
 QPR001260  
 QPR001270  
 QPR001280  
 QPR001290  
 QPR001300  
 QPR001310  
 QPR001320  
 QPR001330  
 QPR001340  
 QPR001350  
 QPR001360  
 QPR001370  
 QPR001380  
 QPR001390  
 QPR001400  
 QPR001410  
 QPR001420  
 QPR001430  
 QPR001440  
 QPR001450  
 QPR001460  
 QPR001470  
 QPR001480  
 QPR001490  
 QPR001500  
 QPR001510  
 QPR001520  
 QPR001530  
 QPR001540  
 QPR001550  
 QPR001560  
 QPR001570  
 QPR001580  
 QPR001590  
 QPR001600

SUBROUTINE QPRS(TYPE2,TYPE3,BAUD,BITS,FREQ,IPHAS,AMP,  
 \*ANS2,DSEED,IN,REP)

\*\*\* PURPOSE \*\*\*

THIS SUBROUTINE MODULATES THE CARRIER USING QUADRATURE PHASE SHIFT  
 KEYING AS THE MODULATION TECHNIQUE.

\*\*\* PARAMETER DEFINITIONS \*\*\*

TYPE2= INDICATES LOGIC TYPE TO BE EMPLOYED  
 TYPE3= INDICATES THE CLASS OF QPRS  
 BAUD= SYMBOL RATE OR BAUD RATE  
 BITS= NUMBER OF BITS IN EACH BINARY CODE WORD  
 FREQ= CARRIER FREQUENCY  
 IPHAS= INITIAL PHASE ANGLE IN DEGREES  
 AMP= AMPLITUDE  
 ANS2= INTEGER VARIABLE TO BE PASSED TO STREAM GENERATOR  
 DSEED= DOUBLE PRECISION SEED FOR RANDOM NUMBER GENERATOR  
 IN= NUMBER OF POSITIONS IN ARRAY TO BE UTILIZED  
 REP= AN INTEGER EQUAL TO THE NUMBER OF THE REPEITION OF  
 THE CALL THE THIS SUBROUTINE

\*\*\* VARIABLE DEFINITIONS \*\*\*

ARRAY= A DOUBLE PRECISION ARRAY FOR PASSING THE BINARY  
 DIGITS AND STORING THE VALUE OF THE TIME FUNCTION  
 AND THE FFT  
 ASUBN= A DOUBLE PRECISION ARRAY CONTAINING THE VALUES OF  
 THE IN PHASE BINARY DIGIT TO BE MODULATED  
 BSUBN= A DOUBLE PRECISION ARRAY CONTAINING THE VALUES OF THE  
 QUADRATURE BINARY DIGIT TO BE MODULATED  
 HOFF= THE IMPULSE RESPONSE FOR THE SPECIFIED CLASS FILTER  
 T= VARIABLE WHICH DETERMINES THE MAGNITUDE OF THE IMPULSE  
 RESPONSE AT A GIVEN TIME FOR A SPECIFIED BINARY DIGIT  
 NDP= VARIABLE USED IN THE COMPUTATION OF T  
 R= AN INTEGER USED TO REPRESENT THE ROW OF ARRAY  
 TIME= TIME VARIABLE  
 MT= VALUE OF BINARY DIGIT PASSED FROM SUBPROGRAM STREAM  
 STEP= INTERVAL AT WHICH THE SIGNAL IS REPRODUCED  
 PI= NUMERICAL CONSTANT  
 OMEGA= CARRIER ANGULAR FREQUENCY  
 DELTA= INITIAL PHASE OFFSET IN RADIANS  
 BAUDD= BAUD DURATION AND QPRS BIT DURATION OF THE TIME SERIES  
 X= COMPLEX WORKING ARRAY RECEIVED BY FUNCTION FFTC  
 IWK= INTEGER VALUE TO BE PLOTTED ON THE ABCISSAE  
 MAXY= MAXIMUM VALUE TO BE PLOTS ON THE ABCISSAE  
 MINY= MINIMUM VALUE TO BE PLOTS ON THE ABCISSAE  
 INT= INTERVAL BETWEEN POINTS ON THE ORDINATE  
 RMAX= VALUE OF THE PRINCIPLE HARMONIC

CC





C

```

DO 40 J=1, INDEX
NDP=J-1
T=TIME-(NDP/BAUD)
IF(TYPE3.EQ.1) THEN
IF(T.EQ.-BAUD/2.D0.OR.T.EQ.BAUD/2.D0) THEN
HOFT=1.D0
ELSE
HOFT= {4.D0/(BAUD**2*PI)} * (DCOS(PI*T*BAUD) /
{1.D0/BAUD**2} - {4.D0*T**2})
END IF
IF (TYPE3.EQ.2) THEN
IF(T.EQ.0.D0) THEN
HOFT=2.D0
ELSE IF(T.EQ.-BAUD.OR.T.EQ.BAUD) THEN
HOFT=1.D0
ELSE
HOFT= {2.D0/(BAUD**3*PI*T)} * (DSIN(PI*T*BAUD) /
{1.D0/BAUD**2} - T**2)
END IF
IF (TYPE3.EQ.3) THEN
IF(T.EQ.0.D0) THEN
HOFT=1.D0
ELSE IF(T.EQ.-BAUD) THEN
HOFT=2.D0
ELSE IF(T.EQ.BAUD) THEN
HOFT=-1.D0
ELSE
HOFT= {1.D0/(BAUD**2*PI*T)} * (DSIN(PI*T*BAUD)) *
{3.D0*T-BAUD} / (T**2 - {1.D0/BAUD**2})
END IF
IF (TYPE3.EQ.4) THEN
IF(T.EQ.-BAUD) THEN
HOFT=1.D0
ELSE IF(T.EQ.BAUD) THEN
HOFT=-1.D0
ELSE
HOFT= {2.D0/(BAUD**2*PI)} * (DSIN(PI*T*BAUD) /
{T**2 - {1.D0/BAUD**2}})
END IF
IF (TYPE3.EQ.5) THEN
IF(T.EQ.0.D0) THEN
HOFT=-2.D0
ELSE IF(T.EQ.-2.D0*BAUD.OR.T.EQ.2.D0*BAUD) THEN
HOFT=1.D0
ELSE
HOFT= {8.D0/(BAUD**3*PI*T)} * (DSIN(PI*T*BAUD) /
{T**2 - {4.D0/BAUD**2}})
END IF

```

C

```

QPR01110
QPR01120
QPR01130
QPR01140
QPR01150
QPR01160
QPR01170
QPR01180
QPR01190
QPR01200
QPR01210
QPR01220
QPR01230
QPR01240
QPR01250
QPR01260
QPR01270
QPR01280
QPR01290
QPR01300
QPR01310
QPR01320
QPR01330
QPR01340
QPR01350
QPR01360
QPR01370
QPR01380
QPR01390
QPR01400
QPR01410
QPR01420
QPR01430
QPR01440
QPR01450
QPR01460
QPR01470
QPR01480
QPR01490
QPR01500
QPR01510
QPR01520
QPR01530
QPR01540
QPR01550
QPR01560
QPR01570
QPR01580
QPR01590
QPR01600

```





```

59 WRITE(10,60)RMAX
60 FORMAT(' THE FFT HAS BEEN GENERATED! THE PRINCIPLE HARMONIC'
C      * IS THE 15. HARMONIC. THE NEXT PLOT TO BE PRODUCED WILL'
C      * BE THE AMPLITUDE SPECTRUM. ENTER A 1 IF YOU ARE READY TO'
C      * CONTINUE WITH THE PROGRAM.'//':)
C      READ(5,*)L
C      IF(L.NE.1)THEN
C          CALL FKTCMS('CLRSCRN ')
70      WRITE(10,70)
C          FORMAT(' ERROR'//)
C          GO TO 59
C      END IF
C      END IF
C      *** ALLOW FOR THE PLOT OF THE AMPLITUDE SPECTRUM OF THE FFT ***
C      IF(REP.EQ.1)THEN
C          MINY=0. DO
C          INT=1. DO/STEP
C          CALL PLOT(MAXY,MINY,INT,IND2P1)
C          END IF
C      *** ADD THE VALUES OF THE FFT TO THE ACCUMULATED STATISTICS ***
C      FLAG=0
C      CALL STATS(IN,REP,FLAG)
C      RETURN
C      END

```

```

QPR 021110
QPR 021120
QPR 021130
QPR 021140
QPR 021150
QPR 021160
QPR 021170
QPR 021180
QPR 021190
QPR 022000
QPR 022100
QPR 022200
QPR 022230
QPR 022240
QPR 022250
QPR 022260
QPR 022270
QPR 022280
QPR 022290
QPR 022300
QPR 022310
QPR 022320
QPR 022330
QPR 022340
QPR 022350
QPR 022360
QPR 022370
QPR 022380
QPR 022390
QPR 022400
QPR 022410
QPR 022420
QPR 022430

```

STR00110  
 STR00120  
 STR00130  
 STR00140  
 STR00150  
 STR00160  
 STR00170  
 STR00180  
 STR00190  
 STR00200  
 STR00210  
 STR00220  
 STR00230  
 STR00240  
 STR00250  
 STR00260  
 STR00270  
 STR00280  
 STR00290  
 STR00300  
 STR00310  
 STR00320  
 STR00330  
 STR00340  
 STR00350  
 STR00360  
 STR00370  
 STR00380  
 STR00390  
 STR00400  
 STR00410  
 STR00420  
 STR00430  
 STR00440  
 STR00450  
 STR00460  
 STR00470  
 STR00480  
 STR00490  
 STR00500  
 STR00510  
 STR00520  
 STR00530  
 STR00540  
 STR00550  
 STR00560  
 STR00570  
 STR00580  
 STR00590  
 STR00600

```

SUBROUTINE STREAM(DSEED,ANS2,TYPE2,MT)
  *** PURPOSE ***
  THIS SUBPROGRAM ALLOWS THE INPUT OF SUCCESSIVE BITS DURING
  MODULATION
  *** PARAMETER DEFINITIONS ***
  DSEED=      DOUBLE PRECISION SEED FOR RANDOM NUMBER GENERATOR
  ANS2=      WHETHER OR NOT THE BIT STREAM IS TO BE GENERATED
              BY THE RANDOM NUMBER GENERATOR
  TYPE2=      THE TYPE OF BINARY LOGIC TO BE EMPLOYED
  MT=        VALUE OF THE BINARY DIGIT TO BE PASSED
  *** VARIABLE DEFINITIONS ***
  GGUBFS=     RANDOM NUMBER GENERATOR
  IMT=        INTEGER VALUE OF BINARY DIGIT TO BE MODULATED
  *** VARIABLE DECLARATIONS ***
  INTEGER ANS2,TYPE2,IMT
  DOUBLE PRECISION DSEED,MT
  IF (ANS2.EQ.1) THEN
    IF (TYPE2.EQ.1) THEN
      IF (GGUBFS(DSEED).LE..5D0) THEN
        ELSE
          MT=-1.D0
        END IF
      ELSE IF (TYPE2.EQ.2) THEN
        IF (GGUBFS(DSEED).LE..5D0) THEN
          ELSE
            MT=1.D0
          END IF
        ELSE IF (TYPE2.EQ.3) THEN
          IF (GGUBFS(DSEED).LE..5D0) THEN
            ELSE
              MT=-1.D0
            END IF
          ELSE IF (GGUBFS(DSEED).LE..5D0) THEN
            ELSE
              MT=0.D0
            END IF
          ELSE IF (GGUBFS(DSEED).LE..5D0) THEN
            ELSE
              MT=-1.D0
            END IF
          ELSE IF (GGUBFS(DSEED).LE..5D0) THEN
            ELSE
              MT=0.D0
            END IF
          ELSE IF (ANS2.EQ.2) THEN
            WRITE(10,10)
            FORMAT('ENTER THE NEXT BIT IN THE DESIRED BIT STREAM.',/)
            READ(5,*)IMT
          END IF
        END IF
      END IF
    END IF
  END IF

```

CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC

MT= I MT  
END IF  
RETURN  
END

STR 006 10  
STR 006 20  
STR 006 30  
STR 006 40



```

SUBROUTINE ORTHO (K, MT, SUM2, C, FLAG)
*** PURPOSE ***
THIS SUBROUTINE GENERATES AN ORTHOGONAL SET OF N VECTORS OF LENGTH
N, WHERE N IS 2**K, WHERE K IS THE LENGTH OF THE BINARY CODE WORD.
AND DETERMINES THE VALUE, MT, OF A BINARY DIGIT TO BE RETURNED
*** PARAMETER DEFINITIONS ***
K=          NUMBER OF BITS IN THE BINARY CODE WORD
MT=         THE VALUE OF THE BINARY DIGIT TO BE MODULATED
SUM2=      THE DECIMAL EQUIVALENT OF A SET OF RANDOMLY DRAWN
C=         BINARY DIGITS
FLAG=     THE COLUMN OF THE VECTOR WHERE THE BINARY DIGIT TO
          BE MODULATED IS LOCATED
          INTEGER WHICH CONTROLS POINT OF ENTRY TO SUBROUTINE
          CRTHO
*** VARIABLE DEFINITIONS ***
ROW=       ROW OF ARRAY H2N WHICH MATCHES A BINARY CODE WORD
KM1=      INTEGER VALUE OF K MINUS 1
N=        NUMBER OF POSSIBLE BINARY CODE WORDS = 2**K
H2N=     MATRIX OF N ORTHOGONAL VECTORS OF LENGTH N
RH2N=    ROW OF MATRIX H2N
CH2N=   COLUMN OF MATRIX H2N
RH2NP1= RANK OF MATRIX H2N EQUAL TO HALF THE RANK PLUS 1
CH2NP1= COLUMN OF MATRIX H2N EQUAL TO HALF THE RANK PLUS 1
X=       DOUBLE PRECISION EQUIVALENT ROW REPRESENTATION OF H2N
*** VARIABLE DECLARATIONS ***
INTEGER K, RH2N, CH2N, RH2NP1, CH2NP1, H2NR, N, KM1, H2N(65,65),
*FLAG, ROW, C
DOUBLE PRECISION X, MT, SUM2
*** VARIABLE INITIALIZATION ***
KM1=K-1
N=2**K
RH2N=1
CH2N=1
H2NR=2
RH2NP1=H2NR+1
CH2NP1=H2NR+1
H2N(1,1)=1
H2N(1,2)=1

```

CC

ORT 00610  
 ORT 00620  
 ORT 00630  
 ORT 00640  
 ORT 00650  
 ORT 00660  
 ORT 00670  
 ORT 00680  
 ORT 00690  
 ORT 00700  
 ORT 00710  
 ORT 00720  
 ORT 00730  
 ORT 00740  
 ORT 00750  
 ORT 00760  
 ORT 00770  
 ORT 00780  
 ORT 00790  
 ORT 00800  
 ORT 00810  
 ORT 00820  
 ORT 00830  
 ORT 00840  
 ORT 00850  
 ORT 00860  
 ORT 00870  
 ORT 00880  
 ORT 00890  
 ORT 00900  
 ORT 00910  
 ORT 00920  
 ORT 00930  
 ORT 00940  
 ORT 00950  
 ORT 00960  
 ORT 00970  
 ORT 00980  
 ORT 00990  
 ORT 01000  
 ORT 01010  
 ORT 01020  
 ORT 01030  
 ORT 01040  
 ORT 01050  
 ORT 01060  
 ORT 01070  
 ORT 01080  
 ORT 01090  
 ORT 01100

```

H2N {2,1} = 1
H2N {2,2} = -1
*** GENERATE A MATRIX OF N ORTHOGONAL VECTORS OF LENGTH N ***
IF (FLAG.EQ.0) THEN
DO 10 I=1,KM1
  DO 20 L=1,H2NR
    DO 30 M=1,H2NR
      H2N (RH2N,CH2N) = H2N (RH2N,CH2N)
      H2N (RH2N,CH2NP1) = H2N (RH2N,CH2N)
      H2N (RH2NP1,CH2N) = H2N (RH2N,CH2N)
      H2N (RH2NP1,CH2NP1) = -H2N (RH2N,CH2N)
      CH2N = CH2N + 1
      CH2NP1 = CH2NP1 + 1
    CONTINUE
  RH2N = RH2N + 1
  RH2NP1 = RH2NP1 + 1
  CH2N = 1
  CH2NP1 = H2NR + 1
CONTINUE
H2NR = 2 * H2NK
RH2N = 1
CH2N = 1
RH2NP1 = H2NR + 1
CH2NP1 = H2NR + 1
CONTINUE
*** ASSIGN EACH ROW OF H2N A REPRESENTATIVE DECIMAL NUMBER ***
DO 40 I=1,N
  H2N (I,65) = I - 1
CONTINUE
*** DETERMINE WHICH ROW OR H2N MATCHES THE VALUE OF THE
REPRESENTATIVE DECIMAL EQUIVALENT TO THE BINARY CODE WORD ***
DO 50 I=1,N
  X = H2N (I,65)
  IF (SUM2.EQ.X) THEN
    END IF
  CONTINUE
END IF
*** DETERMINE THE NEW ROW OF H2N ***
IF (FLAG.EQ.1) THEN
  DO 60 I=1,N

```

C C C  
 30  
 20  
 10  
 C C C  
 40  
 C C C C C  
 50  
 C C C C

```

60      X=H2N(I,65)
        IF (SUM2.EQ.X) THEN
            ROW=I
            END IF
            CONTINUE
        END IF
        *** DETERMINE THE VALUE OF MT TO BE RETURNED ***
        MT=H2N(ROW,C)
        C=C+1
        RETURN
        END

```

```

ORT01110
ORT01120
ORT01130
ORT01140
ORT01150
ORT01160
ORT01170
ORT01180
ORT01190
ORT01200
ORT01210
ORT01220
ORT01230
ORT01240

```





```

***
C      ' WILL HAVE THE OPPORTUNITY TO PLOT MORE THAN ONE SECTION'
C      ' OF THE TOTAL RECORD OF 15 POINTS. ENTER THE NUMBER'
C      ' OF POINTS TO BE PLOTTED.'
C      READ (5,*)NI
C      IF(NI.LE.IN) THEN
C      ELSE
C          CALL FRTCMS('CLRSCRN ')
C          WRITE(10,30)
C          FORMAT('ERROR',/)
C          GO TO 19
C          END IF
C          CALL FRTCMS('CLRSCRN ')
C          WRITE(10,40) IN WHICH NUMBER OF THE 15 X,Y PAIRS DO YOU WISH
C          FORMAT(' AT THE POINT' // ' YOU SPECIFY TO START THE PLOT MUST BE'
C          ' CAUTION! THE POINT YOU SPECIFY TO START THE PLOT MUST BE'
C          ' SMALL ENOUGH TO ALLOW THE ENTIRE RANGE OF POINTS YOU'
C          ' DESIRED TO HAVE PLOTTED AS SPECIFIED BY YOUR PREVIOUS INPUT'
C          )
C          READ (5,*)R
C          IF(R.GT.(IN+1)-NI) THEN
C          CALL FRTCMS(' CLRSCRN ')
C          WRITE(10,50)
C          FORMAT('ERROR',/)
C          GO TO 39
C          END IF
C          ND=NI
C          RANGE(1)=(ARRAY(R,2)+(ND*INT))
C          RANGE(2)=ARRAY(R,2)
C          RANGE(3)=1.5D0*MAXY
C          RANGE(4)=1.5D0*MINY
C          WRITE(10,60)
C          FORMAT(' ')
C          CALL UT PLOT (ARRAY(R,2), ARRAY(R,1), NI, RANGE, 2, 0)
C          WRITE(10,70)
C          FORMAT(' IF YOU WOULD LIKE ANOTHER PLOT OF THE SAME GRAPH OR'
C          ' A PLOT OVER A DIFFERENT RANGE OR FROM ANOTHER STARTING'
C          ' POINT, ENTER A 1. ENTER ANY OTHER INTEGER TO CONTINUE WITH'
C          ' THE PROGRAM.' // ' ')
C          READ (5,*)ANS

```

```

PLO00610
PLO00620
PLO00630
PLO00640
PLO00650
PLO00660
PLO00670
PLO00680
PLO00690
PLO00700
PLO00710
PLO00720
PLO00730
PLO00740
PLO00750
PLO00760
PLO00770
PLO00780
PLO00790
PLO00800
PLO00810
PLO00820
PLO00830
PLO00840
PLO00850
PLO00860
PLO00870
PLO00880
PLO00890
PLO00900
PLO00910
PLO00920
PLO00930
PLO00940
PLO00950
PLO00960
PLO00970
PLO00980
PLO00990
PLO01000
PLO01010
PLO01020
PLO01030
PLO01040
PLO01050
PLO01060
PLO01070
PLO01080
PLO01090
PLO01100

```

\*\*\*

\*\*\*\*\*

\*\*\*

```
C      IF (ANS-EC-1) THEN
      GO TC 19
      END IF
      END IF
      RETURN
      END
```

```
PLO 011110
PLO 011120
PLO 011130
PLO 011140
PLO 011150
PLO 011160
PLO 011170
PLO 011180
```



SUBROUTINE STATS(IN,REP,FLAG)

\*\*\* PURPOSE \*\*\*

THIS SUBPROGRAM COMPILES THE ACCUMULATED STATISTICS OF THE ELEMENTS OF THE AMPLITUDE SPECTRUM OF THE FFTS.

\*\*\* PARAMETER DEFINITIONS \*\*\*

IN= INTEGER OF THE NUMBER OF POSITIONS USED IN ARRAY  
REP= INTEGER OF THE NUMBER OF THE REPETITION OF THE GENERATION OF THE AMPLITUDE SPECTRUM OF THE FFT  
FLAG= AN INTEGER THAT CONTROLS THE POINT OF ENTRANCE OR EXIT FROM THE SUBROUTINE

\*\*\* VARIABLE DEFINITIONS \*\*\*

IND2P1= POSITIONS USED IN ARRAYS SUMX AND SUMXSQ  
ARRAY= AN ARRAY CONTAINING THE VALUES OF THE AMPLITUDE SPECTRUM OF THE FFT  
SUMX= AN ARRAY CONTAINING THE SUM OF THE VALUES OF THE AMPLITUDE SPECTRUM OF THE FFT  
SUMXSQ= AN ARRAY CONTAINING THE SUM OF THE SQUARES OF THE VALUES OF THE AMPLITUDE SPECTRUM OF THE FFT  
SUMX3= AN ARRAY OF THE AMPLITUDE SPECTRUM OF THE SQUARES OF THE VALUES OF THE AMPLITUDE SPECTRUM OF THE FFT  
SUMX4= AN ARRAY OF THE AMPLITUDE SPECTRUM OF THE QUARTICS OF THE VALUES OF THE AMPLITUDE SPECTRUM OF THE FFT  
XBAR= AN ARRAY OF THE AMPLITUDE SPECTRUM OF THE AVERAGE VALUE OF THE AMPLITUDE SPECTRUM  
VAR= AN ARRAY CONTAINING THE VARIANCE OF THE VALUES OF THE AMPLITUDE SPECTRUM  
SKEW= THE ELEMENTS OF THE AMPLITUDE SPECTRUM OF THE DISTRIBUTION OF THE ELEMENTS OF THE AMPLITUDE SPECTRUM  
KUR= AN ARRAY CONTAINING THE KURTOSIS OF THE DISTRIBUTION OF THE ELEMENTS OF THE AMPLITUDE SPECTRUM  
SUMVAR= SUM OF THE VARIANCES OF AN ELEMENT OF THE AMPLITUDE SPECTRUM  
AVAR= AVERAGE VARIANCE OF AN ELEMENT OF THE AMPLITUDE SPECTRUM  
VARVAR= SPECTRUM OF THE VARIANCES OF AN ELEMENT OF THE AMPLITUDE SPECTRUM  
SUMVSQ= VARIANCE OF THE SQUARE OF THE AMPLITUDE SPECTRUM  
SUMSKW= SUM OF THE SQUARES OF THE VARIANCES OF AN ELEMENT OF THE AMPLITUDE SPECTRUM  
ASKEW= SKEWNESS OF THE ELEMENTS OF THE AMPLITUDE SPECTRUM  
VARSKW= SKEWNESS OF THE ELEMENTS OF THE AMPLITUDE SPECTRUM

STA 00110  
STA 00120  
STA 00130  
STA 00140  
STA 00150  
STA 00160  
STA 00170  
STA 00180  
STA 00190  
STA 00200  
STA 00210  
STA 00220  
STA 00230  
STA 00240  
STA 00250  
STA 00260  
STA 00270  
STA 00280  
STA 00290  
STA 00300  
STA 00310  
STA 00320  
STA 00330  
STA 00340  
STA 00350  
STA 00360  
STA 00370  
STA 00380  
STA 00390  
STA 00400  
STA 00410  
STA 00420  
STA 00430  
STA 00440  
STA 00450  
STA 00460  
STA 00470  
STA 00480  
STA 00490  
STA 00500  
STA 00510  
STA 00520  
STA 00530  
STA 00540  
STA 00550  
STA 00560  
STA 00570  
STA 00580  
STA 00590  
STA 00600

CC





```

20 CONTINUE
CC
CC
30 *** COMPUTE THE FINAL STATISTICS IF FLAG EQUALS 1 ***
CC
CC
CC IF (FLAG.EQ. 1) THEN
CC
CC *** COMPUTE STATISTICS ASSOCIATED WITH EACH ELEMENT OF FFT ***
REP=REP-1
REPP=REP
DO 4 I=1,IND2P1
  XBAR(I)=SUMX(I)/REP
  IF (REP.EQ. 1) THEN
    VAR(I)=(REPP*SUMXSQ(I)-SUMX(I)**2)/REPP**2
  ELSE
    VAR(I)=((REPP*SUMXSQ(I)-SUMX(I)**2)/(REPP*(REPP-1.D0)))
  END IF
  SKEW(I)={SUMX3(I)/REPP} - {3.D0*(SUMXSQ(I)/REPP)*
*          {SUMX(I)/REPP} + {2.D0*(SUMX(I)/REPP)**2}
*          {SUMX4(I)/REPP} - {4.D0*(SUMX3(I)/REPP)**2}
*          {SUMX(I)/REPP} + {6.D0*(SUMX(I)/REPP)**4}
*          {SUMXSQ(I)/REPP} - (3.D0*(SUMX(I)/REPP)**4)
**
**
40 CONTINUE
CC
CC
CC *** OUTPUT SOME DATA ***
50 WRITE(6,50) 2X, N, 11X, MEAN, 17X, VARIANCE, 17X, SKEWNESS,
C   , 16X, KURTOSIS
70 DO 60 I=1,IND2P1
60 WRITE(6,70) I, XBAR(I), VAR(I), SKEW(I), KUR(I)
C   , FORMAT(1X, I5, 2X, E23.16, 2X, E23.16, 2X, E23.16)
CONTINUE
61 WRITE(6,61) 2X, N, 10X, SUM X, 18X, SUM X**2, 16X,
C   , SUM X**3, 16X, SUM X**4
62 DO 62 I=1,IND2P1
63 WRITE(6,63) I, SUMX(I), SUMXSQ(I), SUMX3(I), SUMX4(I)
C   , FORMAT(1X, I5, 2X, E23.16, 2X, E23.16, 2X, E23.16)
CONTINUE
CC
CC
CC *** COMPUTE STATISTICS ASSOCIATED WITH ALL ELEMENTS OF FFT ***
SUMVAR=0.D0
SUMVSQ=0.D0
VARVAR=0.D0
SUMSKW=0.D0
STA 011110
STA 011120
STA 011130
STA 011140
STA 011150
STA 011160
STA 011170
STA 011180
STA 011190
STA 011200
STA 011210
STA 011220
STA 011230
STA 011240
STA 011250
STA 011260
STA 011270
STA 011280
STA 011290
STA 011300
STA 011310
STA 011320
STA 011330
STA 011340
STA 011350
STA 011360
STA 011370
STA 011380
STA 011390
STA 011400
STA 011410
STA 011420
STA 011430
STA 011440
STA 011450
STA 011460
STA 011470
STA 011480
STA 011490
STA 011500
STA 011510
STA 011520
STA 011530
STA 011540
STA 011550
STA 011560
STA 011570
STA 011580
STA 011590
STA 011600

```



```

SUMSSQ=0.D0
VARSKW=0.D0
SUMKUR=0.D0
SUMKSKW=0.D0
VARKUR=0.D0
NDP=IND2P1
C
DO 80 I=1,IND2P1
  SUMVAR=SUMVAR+VAR(I)**2
  SUMVSO=SUMVSO+(VAR(I)**2)
  SUMSKW=SUMSKW+SKW(I)
  SUMSSQ=SUMSSQ+(SKW(I)**2)
  SUMKUR=SUMKUR+KUR(I)
  SUMKSKW=SUMKSKW+(KUR(I)**2)
CONTINUE
C
A VAR=SUMVAR/NDP
A SKW=SUMSKW/NDP
A KUR=SUMKUR/NDP
C
DO 90 I=1,IND2P1
  VARVAR=VARVAR+(VAR(I)-AVAR)**2
  VARSKW=VARSKW+(SKW(I)-ASKW)**2
  VARKUR=VARKUR+(KUR(I)-AKUR)**2
CONTINUE
C
VARVAR=VARVAR/(NDP-1.D0)
VARSKW=VARSKW/(NDP-1.D0)
VARKUR=VARKUR/(NDP-1.D0)
C
WRITE(6,91) 2X,'SUM OF VARIANCES',7X,'SUM OF VARIANCES**2')
FORMAT(1',2X,'SUM OF VARIANCES',7X,'SUM OF VARIANCES**2')
C
WRITE(6,92) SUMVAR,SUMVSO
FORMAT(1X,E23.16,2X,E23.16)
C
WRITE(6,93)
FORMAT(6',2X,'SUM OF SKEWNESS',9X,'SUM OF SKEWNESS**2')
C
WRITE(6,94) SUMSKW,SUMSSQ
FORMAT(1X,E23.16,2X,E23.16)
C
WRITE(6,95)
FORMAT(6',2X,'SUM OF KURTOSIS',9X,'SUM OF KURTOSIS**2')
C
WRITE(6,96) SUMKUR,SUMKSKW
FORMAT(1X,E23.16,2X,E23.16)
C
WRITE(6,100)
FORMAT(6',3X,'MEAN VARIANCE',7X,'VARIANCE OF THE VARIANCES',
STA 01610
STA 01620
STA 01630
STA 01640
STA 01650
STA 01660
STA 01670
STA 01680
STA 01690
STA 01700
STA 01710
STA 01720
STA 01730
STA 01740
STA 01750
STA 01760
STA 01770
STA 01780
STA 01790
STA 01800
STA 01810
STA 01820
STA 01830
STA 01840
STA 01850
STA 01860
STA 01870
STA 01880
STA 01890
STA 01900
STA 01910
STA 01920
STA 01930
STA 01940
STA 01950
STA 01960
STA 01970
STA 01980
STA 01990
STA 02000
STA 02010
STA 02020
STA 02030
STA 02040
STA 02050
STA 02060
STA 02070
STA 02080
STA 02090
STA 02100

```

```

C      *
C      )
C 101  WRITE(6,101) AVAR, VARVAR
C      FORMAT(1X,E23.16,2X,E23.16)
C 102  WRITE(6,102)
C      FORMAT(6,0,3X, ' MEAN SKEWNESS',7X, ' VARIANCE OF THE SKEWNESS')
C 103  WRITE(6,103) ASKEW, VARSKW
C      FORMAT(1X,E23.16,2X,E23.16)
C 104  WRITE(6,104)
C      FORMAT(6,0,3X, ' MEAN KURTOSIS',7X, ' VARIANCE OF THE KURTOSIS')
C 105  WRITE(6,105) AKUR, VARKUR
C      FORMAT(1X,E23.16,2X,E23.16)
C 106  WRITE(6,106)
C      FORMAT(6,1)
C      END IF
C      RETURN
C      END
STA02110
STA02120
STA02130
STA02140
STA02150
STA02160
STA02170
STA02180
STA02190
STA02200
STA02210
STA02220
STA02230
STA02240
STA02250
STA02260
STA02270
STA02280
STA02290
STA02300
STA02310
STA02320
STA02330
STA02340

```

APPENDIX B  
IMSL/NON-IMSL ROUTINES UTILIZED

IMSL ROUTINES

IMSL ROUTINE NAME	-	GGUBFS	GGU00110
COMPUTER	-	IEM/SINGLE	GGU00120
LATEST REVISION	-	JUNE 1, 1980	GGU00130
PURPOSE	-	BASIC UNIFORM (0,1) RANDOM NUMBER GENERATOR - FUNCTION FORM OF GGUBS	GGU00140
USAGE	-	FUNCTION GGUBFS (DSEED)	GGU00150
ARGUMENTS	GGUBFS	RESULTANT DEVIATE.	GGU00160
	DSEED	INPUT/OUTPUT DOUBLE PRECISION VARIABLE ASSIGNED AN INTEGER VALUE IN THE EXCLUSIVE RANGE (1.D0, 2147483647.D0) - DSEED IS REPLACED BY A NEW VALUE TO BE USED IN A SUBSEQUENT CALL.	GGU00170
PRECISION/HARDWARE	-	SINGLE/ALL	GGU00180
REQD. IMSL ROUTINES	-	NONE REQUIRED	GGU00190
NOTATION	-	INFORMATION ON SPECIAL NOTATION AND CONVENTIONS IS AVAILABLE IN THE MANUAL INTRODUCTION OR THROUGH IMSL ROUTINE UHELP	GGU00200
COPYRIGHT	-	1978 BY IMSL, INC. ALL RIGHTS RESERVED.	GGU00210
WARRANTY	-	IMSL WARRANTS ONLY THAT IMSL TESTING HAS BEEN APPLIED TO THIS CODE. NO OTHER WARRANTY, EXPRESSED OR IMPLIED, IS APPLICABLE.	GGU00220
REAL FUNCTION GGUBFS (DSEED)			GGU00230
DOUBLE PRECISION DSEED			GGU00240
		SPECIFICATIONS FOR ARGUMENTS	GGU00250
		SPECIFICATIONS FOR LOCAL VARIABLES	GGU00260



```

C
C
C
DOUBLE PRECISION  D2P31M, D2P31
DATA
DATA
DSEED = DMOD(16807, D0*DSEED, D2P31M)
GGUBFS = DSEED / D2P31
RETURN
END
D2P31M = (2**31) - 1
D2P31 = (2**31) (OR AN ADJUSTED VALUE)
D2P31M/2147483647.D0/
D2P31 / 2147483648.D0/
FIRST EXECUTABLE STATEMENT
GGU00510
GGU00520
GGU00530
GGU00540
GGU00550
GGU00560
GGU00570
GGU00580
GGU00590
GGU00600

```

FFT00110  
 FFT00120  
 FFT00130  
 FFT00140  
 FFT00150  
 FFT00160  
 FFT00170  
 FFT00180  
 FFT00190  
 FFT00200  
 FFT00210  
 FFT00220  
 FFT00230  
 FFT00240  
 FFT00250  
 FFT00260  
 FFT00270  
 FFT00280  
 FFT00290  
 FFT00300  
 FFT00310  
 FFT00320  
 FFT00330  
 FFT00340  
 FFT00350  
 FFT00430  
 FFT00440  
 FFT00450  
 FFT00460  
 FFT00470  
 FFT00480  
 FFT00490  
 FFT00500  
 FFT00510  
 FFT00520  
 FFT00530  
 FFT00540  
 FFT00550  
 FFT00560  
 FFT00570  
 FFT00580  
 FFT00590  
 FFT00600  
 FFT00610  
 FFT00620  
 FFT00630  
 FFT00640  
 FFT00650  
 FFT00660  
 FFT00670

-----

IMSL ROUTINE NAME - FFTRC

COMPUTER - IBM/DOUBLE

LATEST REVISION - JANUARY 1, 1978

PURPOSE - COMPUTE THE FAST FOURIER TRANSFORM OF A REAL VALUED SEQUENCE

USAGE - CALL FFTRC (A, N, X, IWK, WK)

ARGUMENTS A INPUT REAL VECTOR OF LENGTH N WHICH CONTAINS THE DATA TO BE TRANSFORMED.  
 N INPUT NUMBER OF DATA POINTS TO BE TRANSFORMED.  
 X N MUST BE A POSITIVE EVEN INTEGER.  
 OUTPUT COMPLEX VECTOR OF LENGTH N/2+1 CONTAINING THE FIRST N/2+1 COEFFICIENTS OF THE FOURIER TRANSFORM. THE REMAINING COEFFICIENTS MAY BE DETERMINED BY  $X(N+2-i) = \text{CONJG}(X(i))$ , FOR  $i=2, \dots, N/2$ .

IWK - INTEGER WORK VECTOR. IF N IS A POWER OF 2, THEN IWK SHOULD BE OF LENGTH M WHERE  $N=2**M$ .  
 SINGLE AND DOUBLE/H32  
 SINGLE/H36, H48, H60

PRECISION/HARDWARE - SINGL, H32, H36, H48, H60

REQD. IMSL ROUTINES - FFTCC, FFT2C

NOTATION - INFORMATION ON SPECIAL NOTATION AND CONVENTIONS IS AVAILABLE IN THE MANUAL INTRODUCTION OR THROUGH IMSL ROUTINE UHELP

REMARKS 1. FFTRC COMPUTES THE FOURIER TRANSFORM, X, ACCORDING TO THE FOLLOWING FORMULA:  

$$X(K+1) = \text{SUM FROM } J = 0 \text{ TO } N-1 \text{ OF } A(J+1) * \text{CEXP}((0.0, (2.0 * \text{PI} * J * K) / N))$$
 FOR  $K=0, 1, \dots, N/2$  AND  $\text{PI}=3.1415\dots$

THE USER CAN COMPUTE THE REMAINING X VALUES BY PERFORMING THE FOLLOWING STEPS;  

$$\text{ND2} = N/2$$

$$\text{DO } 10 \text{ I} = 2, \text{ND2}$$

$$\text{X}(N+2-i) = \text{CONJG}(X(i))$$
 10 CONTINUE

2. FFTRC CAN BE USED TO COMPUTE





```

J = 1
DO 6 I=1,ND2
  X(I) = DCMPLEX(A(J),A(J+1))
  J = J+2
6 CONTINUE
C
GAM = DCMPLEX(ZERO,ZERO)
DO 10 I=1,ND2
  GAM = GAM + X(I)
10 CONTINUE
TP = G(1)-G(2)
GAM = DCMPLEX(TP,ZERO)
C
MTWO = 2
M = 1
DO 15 I=1,IMAX
  IF (ND2 .LE. MTWO) GO TO 20
  MTWO = MTWO+MTWO
  M = M+1
15 CONTINUE
20 IF (ND2 .EQ. MTWO) GO TO 25
C
25 CALL FFT2C(X,M,IWK)
30 ALPHA = X(1) + B(2)
  X(1) = B(1) + B(2)
  ND4 = (ND2+1)/2
  IF (ND4 .LT. 2) GO TO 40
  NP2 = ND2 + 2
  THETA = RPI/ND2
  TP = THETA
  XIMAG = DCMPLEX(ZERO,ONE)
C
DO 35 K = 2,ND4
  NMK = NP2 - K
  S1 = DCONJG(X(NMK))
  ALPHA = X(K) + S1
  BETA = XIMAG*(S1-X(K))
  S1 = DCMPLEX(DCOS(THETA),DSIN(THETA))
  X(K) = (ALPHA+BETA*S1)*HALF
  X(NMK) = DCONJG(ALPHA-BETA*S1)*HALF
  THETA = THETA + TP
35 CONTINUE
40 X(ND2P1) = GAM
  ND2=N/2
  DO 90 I=2,ND2

```

COMPUTE THE CENTER COEFFICIENT

DETERMINE THE SMALLEST M SUCH THAT  
N IS LESS THAN OR EQUAL TO 2\*\*M

N IS NOT A POWER OF TWO, CALL FFTCC  
N IS A POWER OF TWO, CALL FFT2C

DECOMPOSE THE COMPLEX VECTOR X  
INTO THE COMPONENTS OF THE TRANSFORM  
OF THE INPUT DATA.

FFT011180  
FFT011190  
FFT011200  
FFT011210  
FFT011220  
FFT011230  
FFT011240  
FFT011250  
FFT011260  
FFT011270  
FFT011280  
FFT011290  
FFT011300  
FFT011310  
FFT011320  
FFT011330  
FFT011340  
FFT011350  
FFT011360  
FFT011370  
FFT011380  
FFT011390  
FFT011400  
FFT011410  
FFT011420  
FFT011430  
FFT011440  
FFT011450  
FFT011460  
FFT011470  
FFT011480  
FFT011490  
FFT011500  
FFT011510  
FFT011520  
FFT011530  
FFT011540  
FFT011550  
FFT011560  
FFT011570  
FFT011580  
FFT011590  
FFT011600  
FFT011610  
FFT011620  
FFT011630  
FFT011640  
FFT011650  
FFT011660  
FFT011670

```
C 90      X (N+2-I) = DCONJG (X (I))  
C 9005    CONTINUE  
          RETURN  
          END
```

```
FFT01680  
FFT01690  
FFT01700  
FFT01710
```





```

FFTT00610
FFTT00620
FFTT00630
FFTT00640
FFTT00650
FFTT00660
FFTT00670
FFTT00680
FFTT00690
FFTT00700
FFTT00710
FFTT00720
FFTT00730
FFTT00740
FFTT00750
FFTT00760
FFTT00770
FFTT00780
FFTT00790
FFTT00800
FFTT00810
FFTT00820
FFTT00830
FFTT00840
FFTT00850
FFTT00860
FFTT00870
FFTT00880
FFTT00890
FFTT00900
FFTT00910
FFTT00920
FFTT00930
FFTT00940
FFTT00950
FFTT00960
FFTT00970
FFTT00980
FFTT00990
FFTT01000
FFTT01010
FFTT01020
FFTT01030
FFTT01040
FFTT01050
FFTT01060
FFTT01070
FFTT01080
FFTT01090
FFTT01100

```

```

DO 10 I=1,N
  A(I) = CONJG (A(I))
10 CONTINUE
CALL FFT2C (A, M, IWK)
DO 20 I=1,N
  A(I) = CONJG (A(I)) / N
20 CONTINUE

```

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```

SUBROUTINE FFT2C (A, M, IWK)
  INTEGER M, IWK(1)
  COMPLEX*16 A(1)
  INTEGER I, ISP, J, JJ, N8, N2, LC3, S1, S2, S3, CK, SK, SQ, A0, A1, A2, A3,
  N4, N1, C1, C2, B3, TWOPI, TEMP
  DOUBLE PRECISION RAD, B1, B2, B3, TWOPI, TEMP
  COMPLEX*16 ZERO, ZA0, ZA1, ZA2, ZA3, AK2
  EQUIVALENCE (ZA0, Z0(1)), (ZA1, Z1(1)), (ZA2, Z2(1)), (ZA3, Z3(1)),
  (B1, Z1(2)), (A0, Z0(1)), (A1, Z1(1)),
  (B3, Z3(2)), (A2, Z2(1)), (B2, Z2(2)), (A3, Z3(1)),
  (SQ, 7071067811865475D0),
  (SK, 3826834323650898D0),
  (CK, 9238795325112868D0),
  (TWOPI, 6.283185307179586D0),
  (ZERO, 0.0D0), (ONE, 1.0D0),
  (SQ, SQRT(2)), SK = SIN (PI/8), CK = COS (PI/8),
  TWOPI = 2*PI
  FIRST EXECUTABLE STATEMENT

```

```

MP = M+1
N = 2**M
IWK(1) = 1
MM = (M/2)*2
KN = N+1
DO 5 I=2,MP
  IWK(I) = IWK(I-1) + IWK(I-1)
5 CONTINUE
RAD = TWOPI/N
MK = M - 4

```

INITIALIZE WORK VECTOR

```

FFTT011110
FFTT011120
FFTT011130
FFTT011140
FFTT011150
FFTT011160
FFTT011170
FFTT011180
FFTT011190
FFTT011200
FFTT011210
FFTT011220
FFTT011230
FFTT011240
FFTT011250
FFTT011260
FFTT011270
FFTT011280
FFTT011290
FFTT011300
FFTT011310
FFTT011320
FFTT011330
FFTT011340
FFTT011350
FFTT011360
FFTT011370
FFTT011380
FFTT011390
FFTT011400
FFTT011410
FFTT011420
FFTT011430
FFTT011440
FFTT011450
FFTT011460
FFTT011470
FFTT011480
FFTT011490
FFTT011500
FFTT011510
FFTT011520
FFTT011530
FFTT011540
FFTT011550
FFTT011560
FFTT011570
FFTT011580
FFTT011590
FFTT011600

```

```

KB = (MM - EQ - M) GO TO 15
IF (MM - EQ - M) GO TO 15
K2 = IWK (MM + 1) + KB
K0 = K2 - 1
K0 = K0 - 1
AK2 = A (K2)
A (K2) = A (K0) - AK2
A (K0) = A (K0) + AK2
IF (K0 - GE. KB) GO TO 10
15 C1 = ONE
S1 = ZERO
JJ = 0
K = MM - 1
J = 4
IF (K - GE. 1) GO TO 30
GO TO 70
IF (IWK (J) - GT. JJ) GO TO 25
JJ = JJ - IWK (J)
J = J - 1
IF (IWK (J) - GT. JJ) GO TO 25
JJ = JJ - IWK (J)
J = J - 1
K = K + 2
GO TO 20
25 JJ = 4
ISP = IWK (J) + JJ
IF (JJ - EQ. 0) GO TO 40
C2 = JJ * ISP * RAD
C1 = DCOS (C2)
S1 = DSIN (C2)
35 C2 = C1 * ISP - S1 * S1
S2 = C1 * (S1 + S1) - S1 * S1
C3 = C2 * (C1 - S1) - S2 * S1
S3 = C2 * (S1 + S1) + S2 * C1
40 JSP = ISP + KB
DO I = 1, ISP - I
K0 = JSP + ISP
K1 = K0 + ISP
K2 = K1 + ISP
K3 = K2 + ISP
ZA0 = A (K0)
ZA1 = A (K1)
ZA2 = A (K2)
ZA3 = A (K3)
IF (S1 - EQ. ZERO) GO TO 45

```

RESET TRIGONOMETRIC PARAMETERS

DETERMINE FOURIER COEFFICIENTS  
IN GROUPS OF 4

```

TEMP = A1 * C1 - B1 * S1 * C1
A1 = TEMP
B1 = A2 * C2 - B2 * S2 * C2
TEMP = A3 * C3 - B3 * S3 * C3
A2 = TEMP
B2 = A0 - A2
TEMP = A1 - A3
A3 = TEMP
B0 = B0 - B2
TEMP = B1 - B3
B1 = TEMP
B3 = DCMPPLX (A0+A1, B0+B1)
A (K1) = DCMPPLX (A0-A1, B0-B1)
A (K2) = DCMPPLX (A2-E3, B2+A3)
A (K3) = DCMPPLX (A2+B3, B2-A3)
50 CONTINUE
IF (K .LE. 1) GO TO 55
K = K - 2
GO TO 30
55 KB = K3 + ISE

```

45

C  
C  
CHECK FOR COMPLETION OF FINAL  
ITERATION

```

IF (KN .LE. KB) GO TO 70
IF (J .NE. 1) GO TO 60
K = MK
J = J - 1
GO TO 20
60 C2 = C1
IF (J .NE. 2) GO TO 65
C1 = C1 * CK + S1 * SK
S1 = S1 * CK - C2 * SK
GO TO 35
65 C1 = (C1 - S1) * SQ
S1 = (C2 + S1) * SQ
GO TO 35
70 CONTINUE

```

C  
C  
C  
PERMUTE THE COMPLEX VECTOR IN  
REVERSE BINARY ORDER TO NORMAL  
ORDER

FFT01610  
FFT01620  
FFT01630  
FFT01640  
FFT01650  
FFT01660  
FFT01670  
FFT01680  
FFT01690  
FFT01700  
FFT01710  
FFT01720  
FFT01730  
FFT01740  
FFT01750  
FFT01760  
FFT01770  
FFT01780  
FFT01790  
FFT01800  
FFT01810  
FFT01820  
FFT01830  
FFT01840  
FFT01850  
FFT01860  
FFT01870  
FFT01880  
FFT01890  
FFT01900  
FFT01910  
FFT01920  
FFT01930  
FFT01940  
FFT01950  
FFT01960  
FFT01970  
FFT01980  
FFT01990  
FFT02000  
FFT02010  
FFT02020  
FFT02030  
FFT02040  
FFT02050  
FFT02060  
FFT02070  
FFT02080  
FFT02090  
FFT02100



```

C
IF (M .LE. 1) GO TO 9005
MP = M+1
JJ = 1
INITIALIZE WORK VECTOR
IWK(1) = 1
DO 75 I = 2, MP
  IWK(I) = IWK(I-1) * 2
75 CONTINUE
N4 = IWK(MP-2)
IF (M .GT. 2) N8 = IWK(MP-3)
N2 = IWK(MP-1)
LN = N2
NN = IWK(MP) + 1
MP = MP - 4
DETERMINE INDICES AND SWITCH A
J = 2
JK = JJ + N2
AK2 = A(J)
A(J) = A(JK)
A(JK) = AK2
J = J + 1
IF (JJ .GT. N4) GO TO 85
JJ = JJ + N4
GO TO 105
85 JJ = JJ - N4
IF (JJ .GT. N8) GO TO 90
JJ = JJ + N8
GO TO 105
90 JJ = JJ - N8
K = MP
95 IF (IWK(K) .GE. JJ) GO TO 100
JJ = JJ - IWK(K)
K = K - 1
GO TO 95
100 JJ = IWK(K) + JJ
105 IF (JJ .LE. J) GO TO 110
K = NN - J
JK = NN - JJ
AK2 = A(J)
A(J) = A(JJ)
A(JJ) = AK2
AK2 = A(K)
A(K) = A(JK)
A(JK) = AK2
J = J + 1
110
CYCLE REPEATED UNTIL LIMITING NUMBER
OF CHANGES IS ACHIEVED
IF (J .LE. LM) GO TO 80
C
C
C 9005 RETURN

```

```

FFT02110
FFT02120
FFT02130
FFT02140
FFT02150
FFT02160
FFT02170
FFT02180
FFT02190
FFT02200
FFT02210
FFT02220
FFT02230
FFT02240
FFT02250
FFT02260
FFT02270
FFT02280
FFT02290
FFT02300
FFT02310
FFT02320
FFT02330
FFT02340
FFT02350
FFT02360
FFT02370
FFT02380
FFT02390
FFT02400
FFT02410
FFT02420
FFT02430
FFT02440
FFT02450
FFT02460
FFT02470
FFT02480
FFT02490
FFT02500
FFT02510
FFT02520
FFT02530
FFT02540
FFT02550
FFT02560
FFT02570
FFT02580
FFT02590
FFT02600

```

FFT02610

END

NON-IMSL ROUTINE

```

.....
SUBROUTINE UTPLOT
PURPOSE
    PRINTS GRAPHS ON THE STANDARD OUTPUT PRINTER
FEATURES
    1) FULL CONTROL OVER SCALING
    2) ABILITY TO PLOT SINGLE OR DOUBLE PRECISION VECTORS
CALLING SEQUENCE
CALL UTPLOT (X, Y, N, RANGE, K, MODCUR)
DESCRIPTION OF ARGUMENTS
X      VECTOR OF ABSCISSAE
Y      VECTOR OF ASSOCIATED ORDINATES
N      NUMBER OF (X, Y) PAIRS
RANGE  4 WORD SCALING VECTOR WHERE
        RANGE (1) = MAXIMUM X TO BE PLOTTED
        RANGE (2) = MINIMUM X TO BE PLOTTED
        RANGE (3) = MAXIMUM Y TO BE PLOTTED
        RANGE (4) = MINIMUM Y TO BE PLOTTED
K      ALL (X, Y) POINTS OUTSIDE THE ABOVE RANGE WILL BE PLOTTED
        IN THE BORDER OF THE GRAPH.
        EVERY KTH ELEMENT OF X & Y WILL BE PLOTTED, E.G.,
        FOR REAL*4 DATA (SINGLE PRECISION) K=1
        FOR REAL*8 DATA (DOUBLE PRECISION) K=2.
MODCUR CONTROLS THE NUMBER OF CURVES ON ONE GRAPH
        =0 THERE IS ONLY 1 CURVE ON THIS GRAPH
        =1 THIS IS THE FIRST OF TWO OR MORE CURVES ON THIS GRAPH
        =2 THIS IS AN INTERMEDIATE CURVE ON THIS GRAPH
        =3 THIS IS THE LAST CURVE ON THIS GRAPH
SCALING
        SCALING IS PERFORMED ONLY ON THE FIRST SET OF POINTS (WHEN
.....
UTP00110
UTP00120
UTP00130
UTP00140
UTP00150
UTP00160
UTP00170
UTP00180
UTP00190
UTP00200
UTP00210
UTP00220
UTP00230
UTP00240
UTP00250
UTP00260
UTP00270
UTP00280
UTP00290
UTP00300
UTP00310
UTP00320
UTP00330
UTP00340
UTP00350
UTP00360
UTP00370
UTP00380
UTP00390
UTP00400
UTP00410
UTP00420
UTP00430
UTP00440
UTP00450
UTP00460
UTP00470
UTP00480
UTP00490
UTP00500
UTP00510
UTP00520
UTP00530
UTP00540
UTP00550
UTP00560
UTP00570

```



MODCUR = 0 OR 1.) ARRAY RANGE IS USED TO SET UP THE SCALE FACTORS AND NEED ONLY BE DEFINED FOR THE FIRST CALL TO UTPLOT.

GRID LABELLING

THE DATA TO BE GRAPHED WILL BE FIT INTO AN 80 COLUMN BY 60 ROW GRID. THE GRID WILL BE LABELLED THUSLY:

IN THE X DIRECTION (COLUMN-WISE), THERE WILL BE 5 VALUES: THE MAXIMUM, THE MINIMUM, AND 3 INTERMEDIATE AT INCREMENTS OF (RANGE(2) - RANGE(1)) / 4. FROM THE MINIMUM.

IN THE Y DIRECTION (ROW-WISE) THERE WILL BE 7 VALUES: THE MAXIMUM, THE MINIMUM AND 5 INTERMEDIATE AT INCREMENTS OF (RANGE(4) - RANGE(3)) / 6. FROM THE MINIMUM.

IF THE LABELS HAVE A VALUE BETWEEN 1. AND 10\*\*8, THEY WILL BE PRINTED IN AN F11.2 FORMAT, OTHERWISE THEY WILL BE PRINTED IN A 1PE10.3 FORMAT.

PLOTTING

FOUR CHARACTERS ARE USED FOR PLOTTING CURVES, ". "+", "\*\*", AND "X". WHEN MORE THAN 4 CURVES ARE PLOTTED THE CHARACTERS ARE USED REPEATEDLY. IF A NEW CURVE IS TO BE PLACED IN THE PLOTTING GRID WHERE AN OLD CURVE EXISTS, THE NEW CURVE REPLACES THE OLD ONE. THUS, IF 3 IDENTICAL CURVES ARE PLOTTED, THEY WILL APPEAR AS ONE CURVE COMPOSED OF "\*\*'S.

MESSAGES

UNDER CERTAIN CIRCUMSTANCES A PLOT WILL NOT BE OUTPUT AND ONE OF THE FOLLOWING MESSAGES WILL BE PRINTED ON THE STANDARD OUTPUT IN PLACE OF THE PLOT.

"ALL Y-VALUES=0. CANNOT SETUP PLOT GRID. CHECK MAX & MIN Y WHEN MODCUR=0 OR 1."

"ALL X VALUES=0. CANNOT SETUP PLOT GRID. CHECK MAX AND MIN X WHEN MODCUR=0 OR 1."

"GRID NOT SETUP WHEN MODCUR LAST 0 OR 1. NO PLOT UNTIL GRID PROPERLY SETUP."

NOTE

THE USER IS EXPECTED TO PROVIDE THE NECESSARY CARRIAGE CONTROLS TO PLACE THE GRAPH PROPERLY ON THE PAGE. BEFORE CALLING UTPLOT THE USER SHOULD ISSUE A PRINT STATEMENT WHICH EJECTS A PAGE SO THAT THE GRAPH WILL BE PLOTTED AT THE TOP

UTP 00580  
UTP 00590  
UTP 00600  
UTP 00610  
UTP 00620  
UTP 00630  
UTP 00640  
UTP 00650  
UTP 00660  
UTP 00670  
UTP 00680  
UTP 00690  
UTP 00700  
UTP 00710  
UTP 00720  
UTP 00730  
UTP 00740  
UTP 00750  
UTP 00760  
UTP 00770  
UTP 00780  
UTP 00790  
UTP 00800  
UTP 00810  
UTP 00820  
UTP 00830  
UTP 00840  
UTP 00850  
UTP 00860  
UTP 00870  
UTP 00880  
UTP 00890  
UTP 00900  
UTP 00910  
UTP 00920  
UTP 00930  
UTP 00940  
UTP 00950  
UTP 00960  
UTP 00970  
UTP 00980  
UTP 00990  
UTP 01000  
UTP 01010  
UTP 01020  
UTP 01030  
UTP 01040  
UTP 01050  
UTP 01060  
UTP 01070

CC





UTP 01580  
 UTP 01590  
 UTP 01600  
 UTP 01610  
 UTP 01620  
 UTP 01630  
 UTP 01640  
 UTP 01650  
 UTP 01660  
 UTP 01670  
 UTP 01680  
 UTP 01690  
 UTP 01700  
 UTP 01710  
 UTP 01720  
 UTP 01730  
 UTP 01740  
 UTP 01750  
 UTP 01760  
 UTP 01770  
 UTP 01780  
 UTP 01790  
 UTP 01800  
 UTP 01810  
 UTP 01820  
 UTP 01830  
 UTP 01840  
 UTP 01850  
 UTP 01860  
 UTP 01870  
 UTP 01880  
 UTP 01890  
 UTP 01900  
 UTP 01910  
 UTP 01920  
 UTP 01930  
 UTP 01940  
 UTP 01950  
 UTP 01960  
 UTP 01970  
 UTP 01980  
 UTP 01990  
 UTP 02000  
 UTP 02010  
 UTP 02020  
 UTP 02030  
 UTP 02040  
 UTP 02050  
 UTP 02060  
 UTP 02070

```

IF(YMIN.EQ.0.) GO TO 889
YMIN=0.
YRANGE=YMAX
GO TO 299
298 IF (XRANGE.NE.0.) GO TO 299
IF(XMIN.EQ.0.) GO TO 887
XMIN=0.
XRANGE=XMAX

C
C
C BLANKING OUT MATRIX-(GRID)
DO 300 I=1,61
DO 301 JJ=1,81
GRID(I,JJ)=BLANK
301 CONTINUE
300 IF(XMAX*XMIN.GE.0.) GO TO 222
IYAXIS=80.*(-XMIN)/XRANGE+1.5
DO 40 I=1,61
40 GRID(I,IYAXIS)=DOT
222 IF(YMAX*YMIN.GE.0.) GC TO 333
IXAXIS=60.*YMAX/YRANGE+1.5
DO 60 I=1,81
60 GRID(IXAXIS,I)=DOT

C
C
C COMPUTE PROPER SCALE NUMBERS
333 XINCR=XRANGE/4.
YINCR=YRANGE/6.
XSCALE(1)=XMAX
XSCALE(5)=XMIN
DO 80 I=2,4
80 XSCALE(I)=XSCALE(I-1)-XINCR
IF(ABS(XSCALE(I)).LT.1.E-4) XSCALE(I)=0.
CONTINUE
YSCALE(1)=YMAX
YSCALE(7)=YMIN
DO 81 I=2,6
81 YSCALE(I)=YSCALE(I-1)-YINCR
IF(ABS(YSCALE(I)).LT.1.E-4) YSCALE(I)=0.
CONTINUE
DO 85 II=1,2
85 JJ=6-II
XT=XSCALE(JJ)
XSCALE(JJ)=XSCALE(II)
XSCALE(II)=XT

C
C
C PLACING POINTS IN THEIR PROPER GRID POSITIONS
444 IF(MODCUR.LT.2) JSET=0
  
```



```

UTP02080
UTP02090
UTP02100
UTP02110
UTP02120
UTP02130
UTP02140
UTP02150
UTP02160
UTP02170
UTP02180
UTP02190
UTP02200
UTP02210
UTP02220
UTP02230
UTP02240
UTP02250
UTP02260
UTP02270
UTP02280
UTP02290
UTP02300
UTP02310
UTP02320
UTP02330
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UTP02350
UTP02360
UTP02370
UTP02380
UTP02390
UTP02400
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UTP02440
UTP02450
UTP02460
UTP02470
UTP02480
UTP02490
UTP02500
UTP02510
UTP02520
UTP02530
UTP02540
UTP02550
UTP02560
UTP02570

IF(JERR.GT.0) GO TO 885
JSET=JSET+1
IF(JSET.GT.4) JSET=1
DO 700 I=1,MAX(I)} /YRANGE+1.5
IPTX=60.*(X(I)-XMIN)} /YRANGE+1.5
IF(IPTX.GT.61) OR IPTY.GT.81) GO TO 70
IF(IPTX.LE.0) OR IPTY.LE.0) GO TO 70
GRID(IPTX,IPTY) = XCHAR(JSET)
GO TO 700
IERR=IERR+1
CONTINUE
OUTPUT SECTION WITH GRAPH
C
C
IF(MODCUR.EQ.1) OR MODCUR.EQ.2) RETURN
AXR=ABS(XRANGE)
AYR=ABS(YRANGE)
IF(AXR.LT.1.E+8) AND(AYR.GE..95) GO TO 400
WRITE(6,17) XSCALE
FORMAT(12X,1PE10.3,4(10X,E10.3)/15X,***,8(++++*****),'+***)
17 GO TO 401
WRITE(6,117) XSCALE
FORMAT(6,8X,F11.2,4(9X,F11.2)/15X,***,8(++++*****),'+***)
400 WRITE(6,117) XSCALE
117 FORMAT(6,8X,F11.2,4(9X,F11.2)/15X,***,8(++++*****),'+***)
401 II=1
DO 101 IK=1,61
IF(MOD(IK-1,10).NE.0) GO TO 92
IF(AYR.LT.1.E+8) AND(AYR.GE..95) GO TO 404
WRITE(6,18) YSCALE(II),GRID(IK,IX),IX=1,81, YSCALE(II)
18 FORMAT(3X,1PE10.3,2X,1H+,1X,81A1,1X,1H+,2X,E10.3)
GO TO 405
WRITE(6,118) YSCALE(II),GRID(IK,IX),IX=1,81, YSCALE(II)
404 WRITE(6,118) YSCALE(II),GRID(IK,IX),IX=1,81, YSCALE(II)
118 FORMAT(2X,F11.2, YSCALE(II),8A1, YSCALE(II)
405 II=II+1
GO TO 101
WRITE(6,19) (GRID(IK,IX),IX=1,81)
19 FORMAT(15X,*,8A1,*)
CONTINUE
IF(AXR.LT.1.E+8) AND(AYR.GE..95) GO TO 402
WRITE(6,22) XSCALE
FORMAT(15X,***,8(++++*****),'+***/12X,1PE10.3,4(10X,E10.3),//)
22 GO TO 403
WRITE(6,217) XSCALE
FORMAT(6,15X,***,8(++++*****),'+***/8X,F11.2,4(9X,F11.2),//)
402 WRITE(6,217) XSCALE
217 FORMAT(6,15X,***,8(++++*****),'+***/8X,F11.2,4(9X,F11.2),//)
403 IF(IERR.GT.0) WRITE(6,20) IERR
20 FORMAT(10X,'NUMBER OF POINTS OUT OF RANGE =',I4)
1000 RETURN
C
889 WRITE(6,888)
888 FORMAT(' ALL Y VALUES=0. CANNOT SETUP PLOT GRID. CHECK MAX & MIN Y
1 WHEN MODCUR=0 OR 1.')

```

```

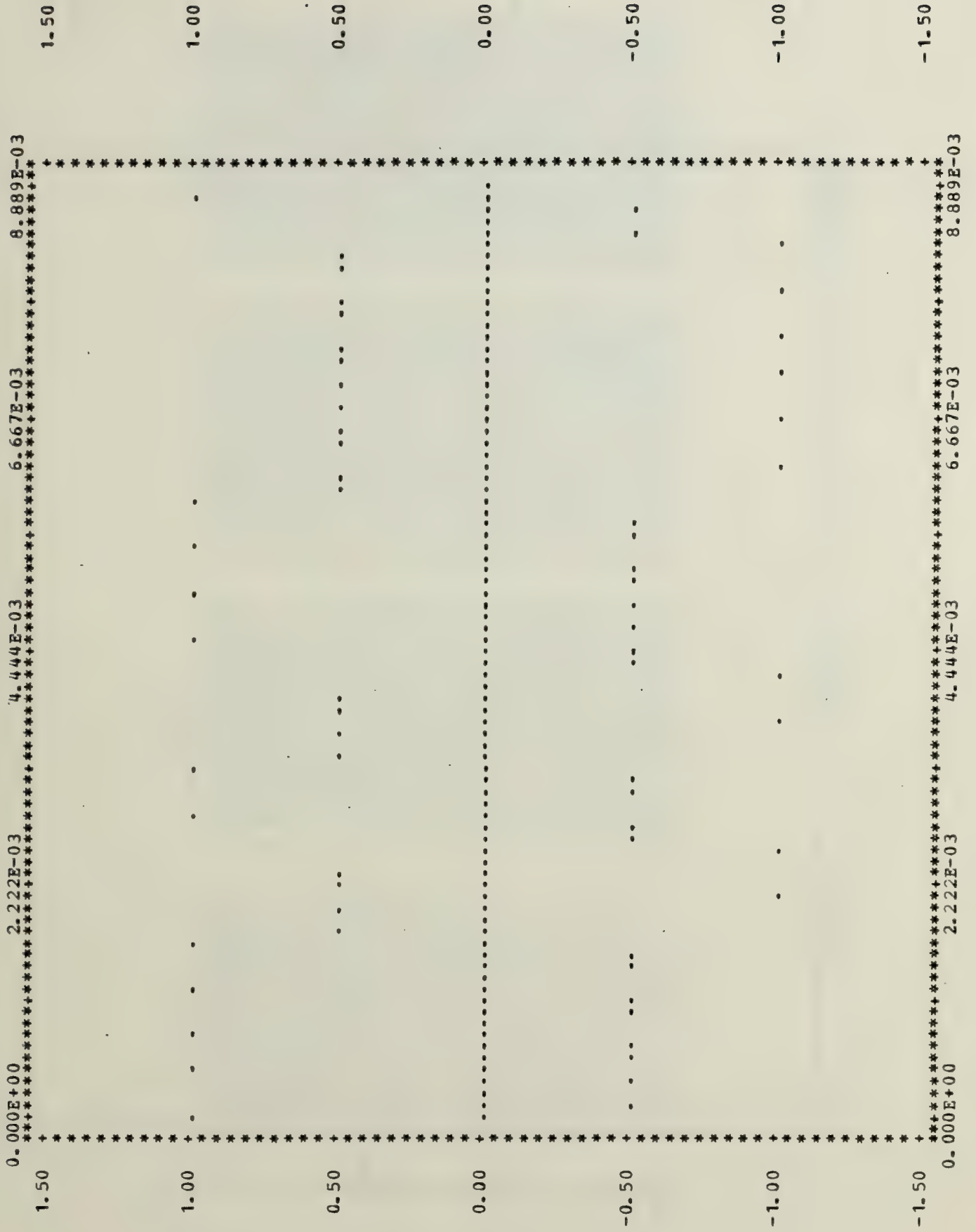
JERR=10
RETURN(6, 886)
887 WRITE(6, 886) ALL X VALUES=0.- CANNOT SETUP PLOT GRID. CHECK MAX & MIN
886 FORMAT(1 WHEN MODCUR=0 OR 1.)
JERR=10
RETURN(6, 884)
885 WRITE(6, 884) GRID NOT SETUP WHEN MODCUR LAST 0 OR 1. NO PLOT UNTIL GRID
884 FORMAT(1D PROPERLY SETUP!)
RETURN
END
UTP02580
UTP02590
UTP02600
UTP02610
UTP02620
UTP02630
UTP02640
UTP02650
UTP02660
UTP02670
UTP02680
UTP02690

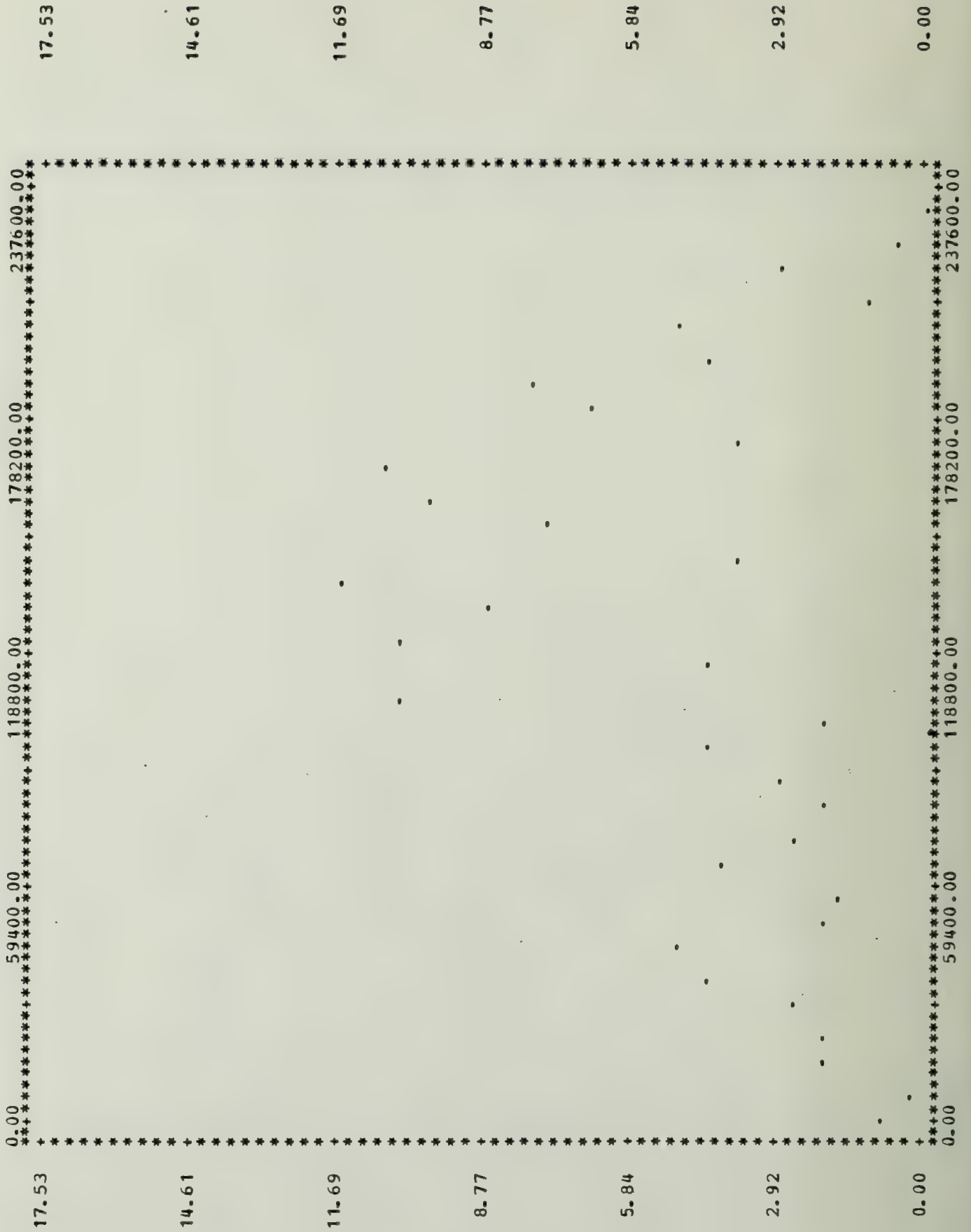
```

APPENDIX C  
 REPRESENTATIVE RESULTS OF TRIAL SIMULATIONS

MODULATION TECHNIQUE = BPSK  
 BIFOLIAR LOGIC  
 BAUD OR SYMBOL RATE = 1200 HZ  
 BITS PER BINARY CODE WORD = 1  
 BIT RATE = 0.1200000000000000E+04  
 CARRIER FREQUENCY = 2400 HZ  
 MAXIMUM CARRIER AMPLITUDE = 1 VOLT(S)  
 INITIAL PHASE ANGLE = 0 DEGREES  
 TIME BETWEEN SAMPLES = 0.1388888888888889E-03 SEC  
 NUMBER OF SAMPLES GENERATED = 64  
 SEED FOR RANDOM NUMBER GENERATOR = 1  
 NUMBER OF TIMES SIMULATION REPEATS = 100  
 SUM OF VARIANCES SUM OF VARIANCES\*\*2  
 0.2015734723265312E+03 0.2783281622639908E+04  
 SUM OF SKEWNESS SUM OF SKEWNESS\*\*2  
 -0.1039295264723739E+05 0.1054665065832765E+08  
 SUM OF KURTOSIS SUM OF KURTOSIS\*\*2  
 0.7097901826525830E+04 0.5645466161770898E+07  
 MEAN VARIANCE VARIANCE OF THE VARIANCES  
 0.6108287040197914E+01 0.4850040606187702E+02  
 MEAN SKEWNESS VARIANCE OF THE SKEWNESS  
 -0.3149379590071935E+03 0.2272973551108844E+06  
 MEAN KURTOSIS VARIANCE OF THE KURTOSIS  
 0.2150879341371464E+03 0.1287122850373493E+06





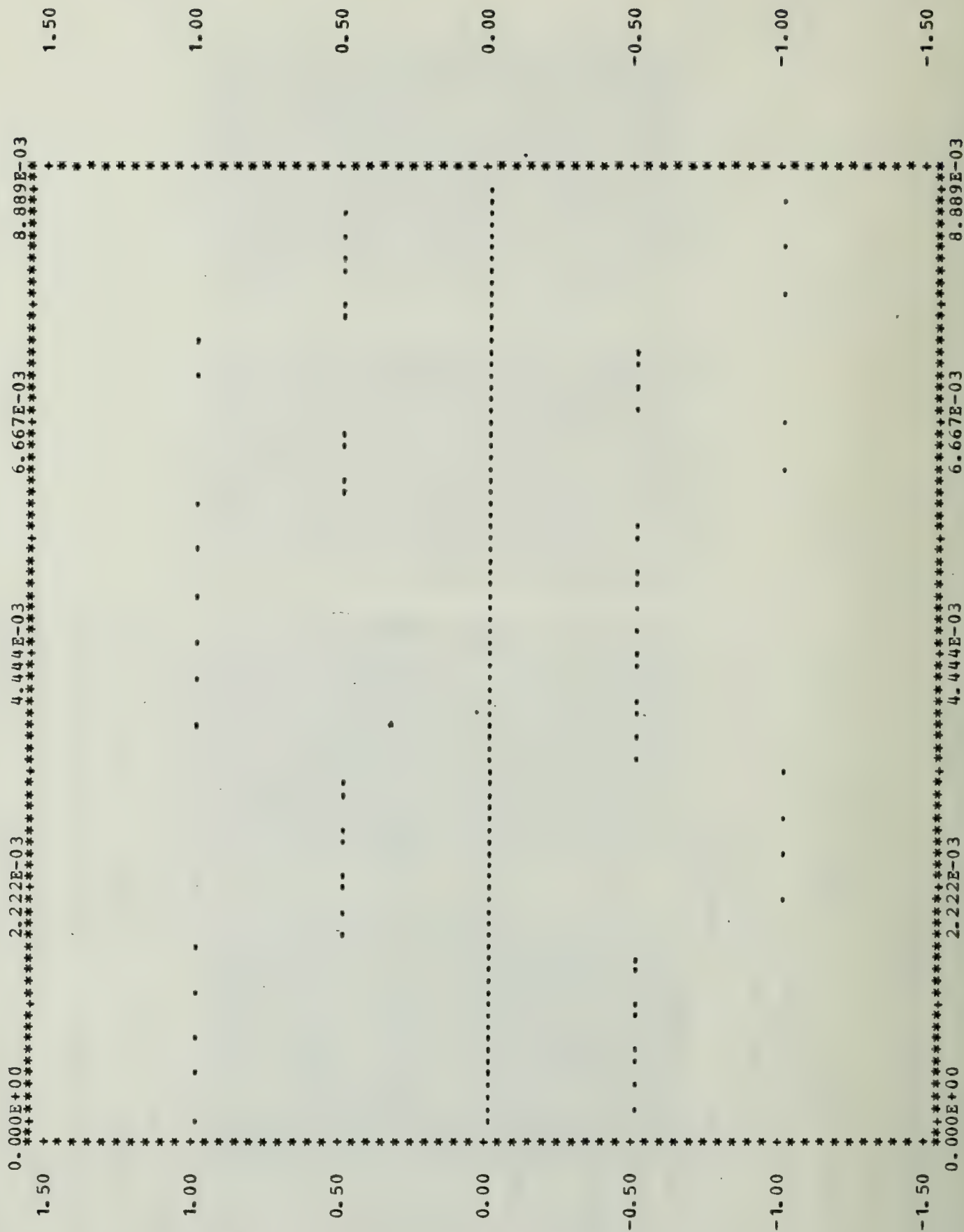




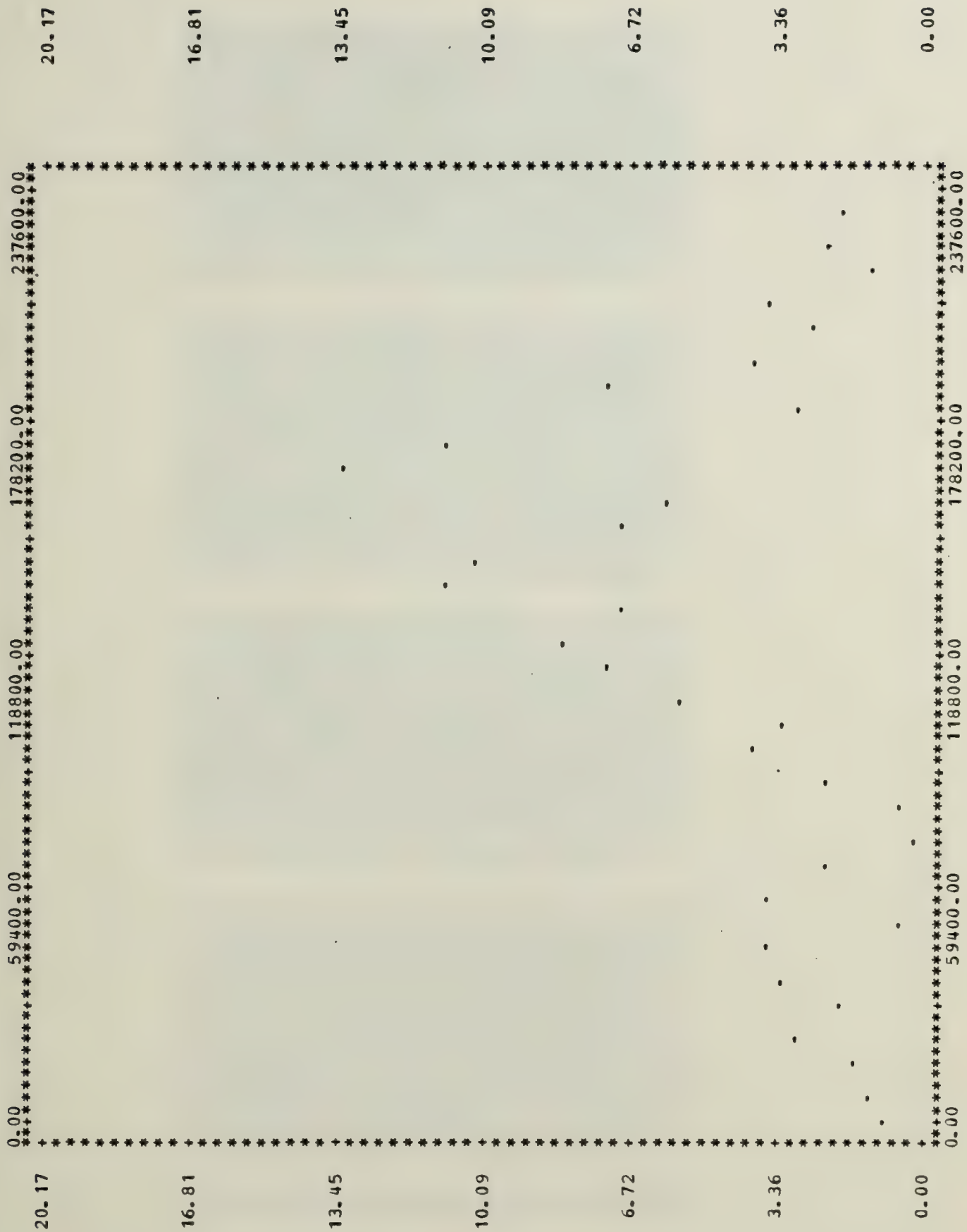


N	X	X*	X**2	X**3	X**4
1	1000	0000	0000	0000	0000
2	1047	0873	0762	0729	0700
3	1781	9901	4950	2979	1371
4	2374	1333	1942	1150	0525
5	2634	1586	2072	1480	0401
6	2930	1785	2409	1793	0259
7	2745	1586	2072	1480	0401
8	2050	1025	1300	0625	0025
9	1456	0809	0840	0512	0033
10	1456	0809	0840	0512	0033
11	1456	0809	0840	0512	0033
12	1920	1440	2720	1536	0000
13	3224	5552	1808	0576	0000
14	4083	7557	2687	0974	0000
15	5391	1087	3905	1521	0000
16	5177	1031	3521	1333	0000
17	7437	1777	5400	2267	0000
18	8281	2033	6144	2621	0000
19	9168	2333	7008	3000	0000
20	8774	2094	6453	2652	0000
21	9524	2333	7008	3000	0000
22	8962	2196	6453	2652	0000
23	7991	1369	4900	1255	0000
24	8132	1358	4867	1239	0000
25	9294	2930	5588	3066	0000
26	4057	828	1568	0589	0000
27	3416	828	1568	0589	0000
28	2867	333	0883	0222	0000
29	2020	571	0339	0088	0000
30	1365	174	0075	0025	0000
31	1080	000	0000	0000	0000
32	1080	000	0000	0000	0000
33	1080	000	0000	0000	0000

MODULATION TECHNIQUE = DBPSK  
 BIFOLAR LOGIC  
 BAUD OR SYMBOL RATE = 1200 HZ  
 BITS PER BINARY CODE WORD = 1  
 BIT RATE = 0.1200000000000000E+04  
 CARRIER FREQUENCY = 2400 HZ  
 MAXIMUM CARRIER AMPLITUDE = 1 VOLT(S)  
 INITIAL PHASE ANGLE = 0 DEGREES  
 TIME BETWEEN SAMPLES = 0.13888888888888889E-03 SEC  
 NUMBER OF SAMPLES GENERATED = 64  
 SEED FOR RANDOM NUMBER GENERATOR = 1  
 NUMBER OF TIMES SIMULATION REPEATS = 100  
 SUM OF VARIANCES SUM OF VARIANCES\*\*2  
 0.2149282745632018E+03 0.3048987837599078E+04  
 SUM OF SKEWNESS SUM OF SKEWNESS\*\*2  
 -0.9719603900974997E+04 0.8598103883304453E+07  
 SUM OF KURTOSIS SUM OF KURTOSIS\*\*2  
 0.8665643426712656E+04 0.8293411900734219E+07  
 MEAN VARIANCE VARIANCE OF THE VARIANCES  
 0.6512978017066721E+01 0.5153639719133948E+02  
 MEAN SKEWNESS VARIANCE OF THE SKEWNESS  
 -0.2945334515446968E+03 0.1792298562094682E+06  
 MEAN KURTOSIS VARIANCE OF THE KURTOSIS  
 0.2625952553549290E+03 0.1880579703838077E+06







N	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	24	25	26	27	28	29	30	31	32	33																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																											
MEAN	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999	99999																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																											
STDEV	120699	161427	226069	319389	382292	455228	532633	615277	707733	800044	899099	100000	110000	120000	130000	140000	150000	160000	170000	180000	190000	200000	210000	220000	230000	240000	250000	260000	270000	280000	290000	300000	310000	320000	330000																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																									
VARIANCE	1469890	3394579	1028688	1860771	2773607	3892255	5588110	7960922	10496015	14095170	19683966	27665020	38503692	53223754	73154951	99549313	133954077	186204196	259946174	357307444	494430810	675328326	913374999	122905182	167485652	233848871	323792095	442959540	607716688	822078877	110332009	150198663	200000000																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																											
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X**12	85632000	214800000	509485760	68848824	148655527	315982409	672172522	1441172522	272402525	518712525	75372525	98837525	122382525	1470272525	17382525	20222525	23222525	26472525	29972525	33722525	37722525	41972525	46522525	51372525	56522525	62072525	67922525	74072525	80422525	87022525	93822525	100822525	108022525	115422525	12322525
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X**13	100368000	250800000	609485760	82848824	178655527	385982409	832172522	1721172522	312402525	688712525	98372525	129382525	1608272525	19382525	22822525	26422525	30172525	34172525	38422525	42922525	47672525	52622525	57872525	63422525	69272525	75322525	81672525	88322525	95222525	102422525	109822525	117422525	12522525	13322525	
SUM	100368000	250800000	609485760	82848824	178655527	385982409	832172522	1721172522	312402525	688712525	98372525	129382525	1608272525	19382525	22822525	26422525	30172525	34172525	38422525	42922525	47672525	52622525	57872525	63422525	69272525	75322525	81672525	88322525	95222525	102422525	10				



MODULATION TECHNIQUE = ORTHOGONAL BPSK  
 BIFOLAR LOGIC

BAUD OR SYMBOL RATE = 1200 HZ  
 BITS PER BINARY CODE WORD = 3

BIT RATE = 0.9600000000000000E+04

CARRIER FREQUENCY = 19200 HZ

MAXIMUM CARRIER AMPLITUDE = 1 VOLT(S)

INITIAL PHASE ANGLE = 0 DEGREES

TIME BETWEEN SAMPLES = 0.2450980392156863E-04 SEC

NUMBER OF SAMPLES GENERATED = 64

SEED FOR RANDOM NUMBER GENERATOR = 1

NUMBER OF TIMES SIMULATION REPEATS = 100

SUM OF VARIANCES SUM OF VARIANCES\*\*2

0.3565282677437312E+03 0.1043543362358570E+05

SUM OF SKEWNESS SUM OF SKEWNESS\*\*2

-0.5381895541377615E+04 0.3582631135045827E+07

SUM OF KURTOSIS SUM OF KURTOSIS\*\*2

0.3658999158877697E+05 0.3330606974397919E+09

MEAN VARIANCE VARIANCE OF THE VARIANCES

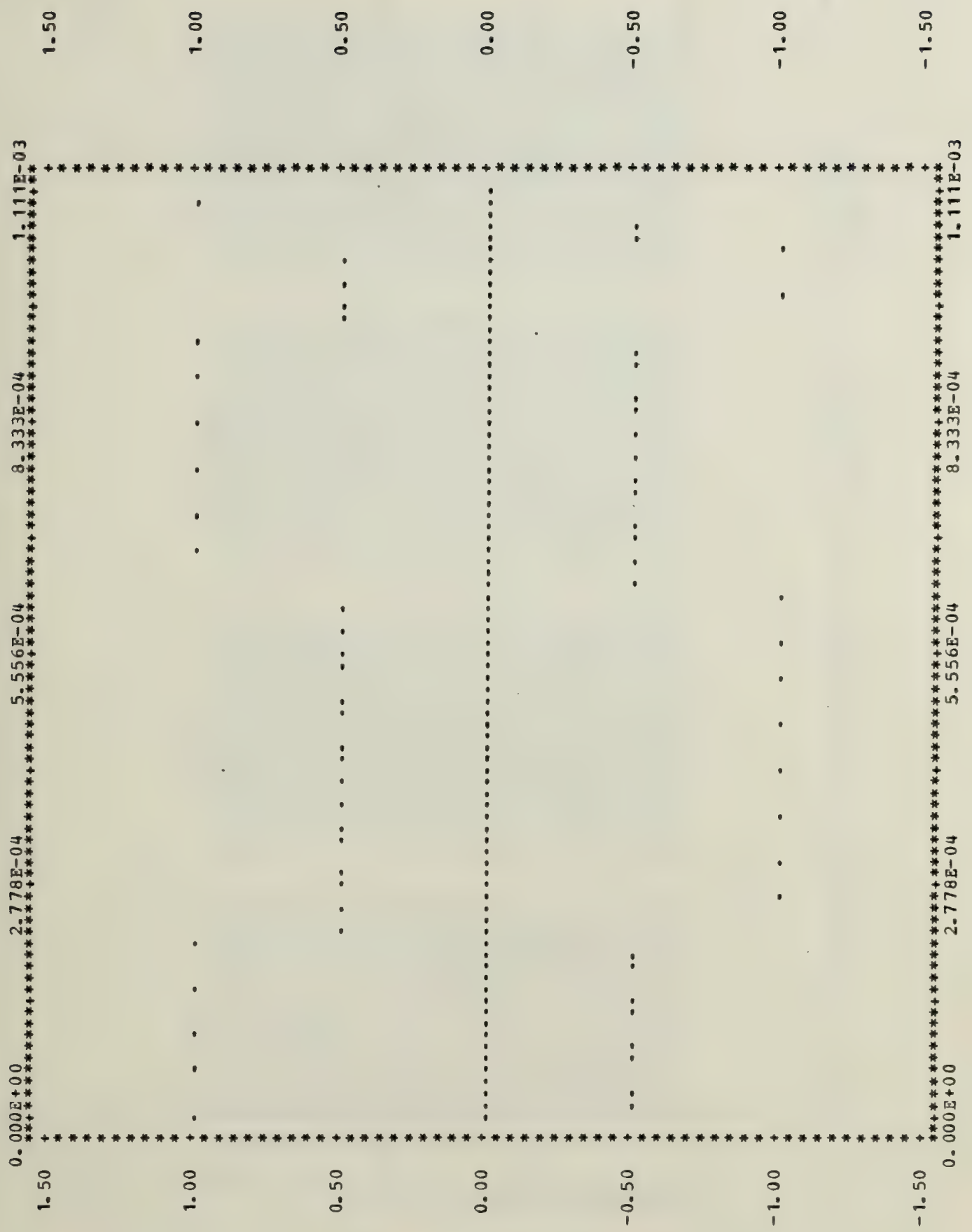
0.1080388690132519E+02 0.2057357044299072E+03

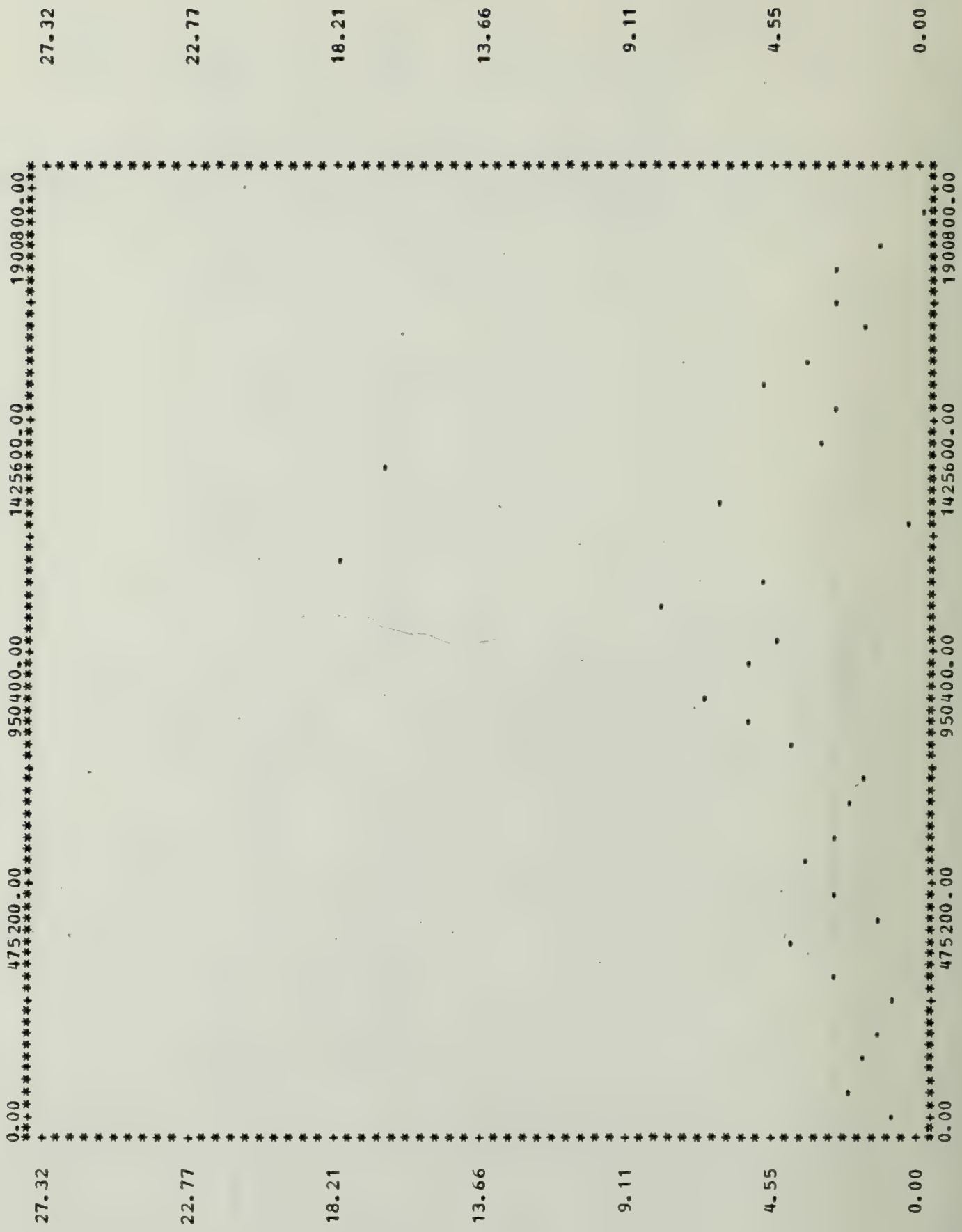
MEAN SKEWNESS VARIANCE OF THE SKEWNESS

-0.1630877436781095E+03 0.8452843545285224E+05

MEAN KURTOSIS VARIANCE OF THE KURTOSIS

0.1108787623902332E+04 0.9140317737733299E+07







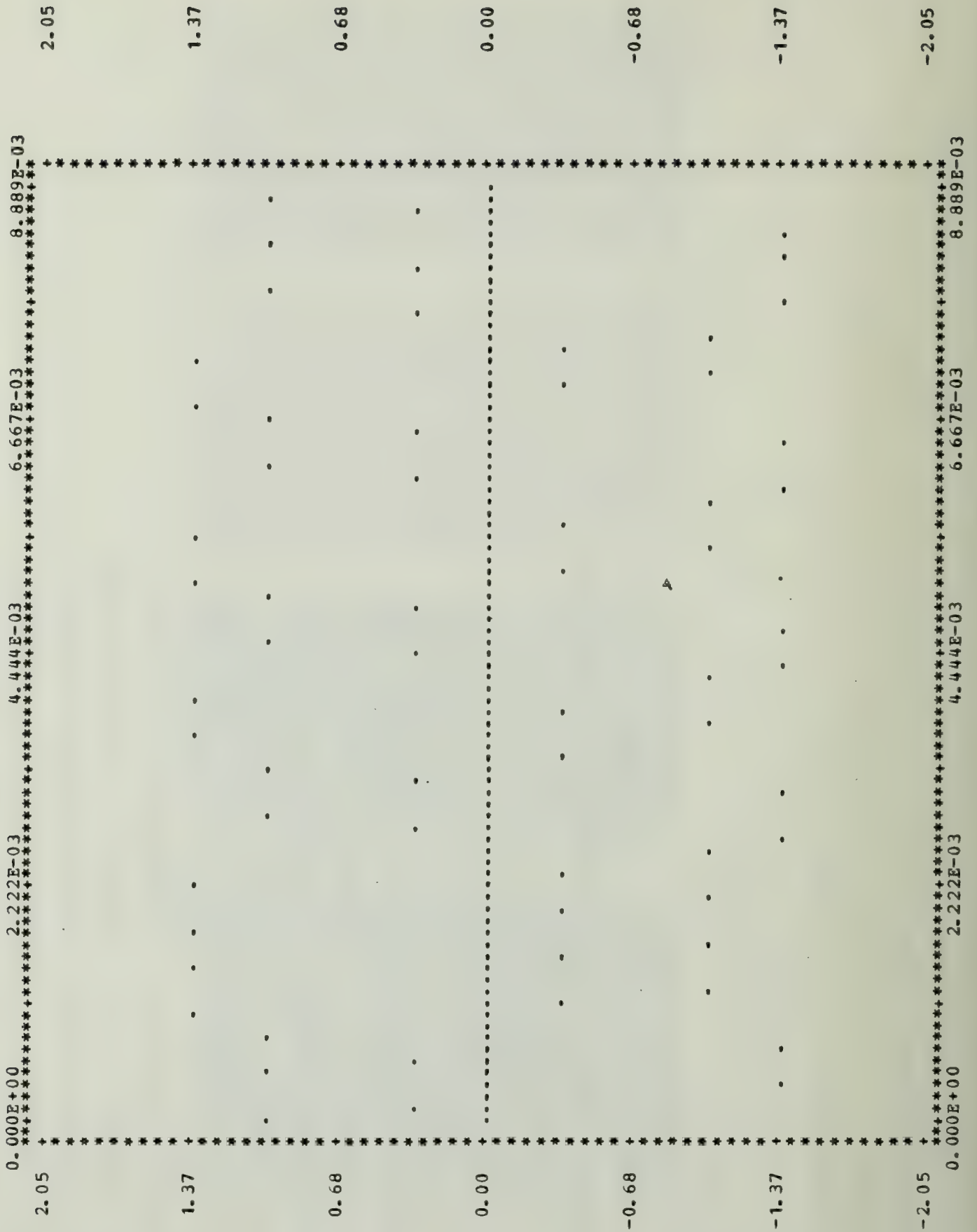






MODULATION TECHNIQUE = QPSK  
 BIFOLAR LOGIC  
 BAUD OR SYMBOL RATE = 1200 HZ  
 BITS PER BINARY CODE WORD = 2  
 BIT RATE = 0.2400000000000000E+04  
 CARRIER FREQUENCY = 2400 HZ  
 MAXIMUM CARRIER AMPLITUDE = 1 VOLT(S)  
 INITIAL PHASE ANGLE = 0 DEGREES  
 TIME BETWEEN SAMPLES = 0.13888888888888889E-03 SEC  
 NUMBER OF SAMPLES GENERATED = 64  
 SEED FOR RANDOM NUMBER GENERATOR = 1  
 NUMBER OF TIMES SIMULATION REPEATS = 100  
 SUM OF VARIANCES SUM OF VARIANCES\*\*2  
 0.4090708290354776E+03 0.1062919801427033E+05  
 SUM OF SKEWNESS SUM OF SKEWNESS\*\*2  
 -0.2867982085342969E+05 0.7352270919634266E+08  
 SUM OF KURTOSIS SUM OF KURTOSIS\*\*2  
 0.3172130538617988E+05 0.9751577105721891E+08  
 MEAN VARIANCE VARIANCE OF THE VARIANCES  
 0.1239608572834781E+02 0.1736975296431322E+03  
 MEAN SKEWNESS VARIANCE OF THE SKEWNESS  
 -0.8690854804069604E+03 0.1518671665998567E+07  
 MEAN KURTOSIS VARIANCE OF THE KURTOSIS  
 0.9612516783690871E+03 0.2094487906709223E+07







N	MEAN	VARIANCE	SKETCHES	KURTOSIS
1	999999	1469890	2199433	2442386
2	999999	8047543	3202742	3833606
3	999999	4087495	4622839	8331833
4	999999	302963	5448833	4922492
5	999999	2201428	6222193	3166300
6	999999	4578714	3309457	7050465
7	999999	3045787	4334533	2804818
8	999999	4839599	5334533	9881952
9	999999	936237	6068429	2477281
10	999999	562139	7068429	6679435
11	999999	456213	8598733	2661541
12	999999	358155	9398733	894158
13	999999	186706	1068596	169278
14	999999	54055	1259311	46799
15	999999	388906	1559311	894158
16	999999	478840	1859651	169278
17	999999	72628	2268005	46799
18	999999	46110	2929979	894158
19	999999	15507	356788	169278
20	999999	39547	4495271	46799
21	999999	38379	533590	894158
22	999999	19404	626987	169278
23	999999	32158	733449	46799
24	999999	28397	833053	894158
25	999999	56530	937790	169278
26	999999	66927	105338	46799
27	999999	31649	118363	894158
28	999999	18360	133604	169278
29	999999	84316	149705	46799
30	999999	36916	167248	894158
31	999999	89364	185999	169278
32	999999	44033	206636	46799
33	999999	18151	235866	894158
34	999999	9484	27154	169278
35	999999	9484	31676	46799
36	999999	9484	36384	894158
37	999999	9484	41492	169278
38	999999	9484	46600	46799
39	999999	9484	51703	894158
40	999999	9484	56811	169278
41	999999	9484	61919	46799
42	999999	9484	67027	894158
43	999999	9484	72135	169278
44	999999	9484	77243	46799
45	999999	9484	82351	894158
46	999999	9484	87459	169278
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82	999999	9484	271363	169278
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86	999999	9484	291797	46799
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88	999999	9484	302014	169278
89	999999	9484	307123	46799
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91	999999	9484	317340	169278
92	999999	9484	322448	46799
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 0- 335493333891581056E+05  
 0- 14213333891581056E+05  
 0- 19233405871340475E+05  
 0- 4869186070483972E+05  
 0- 5807188523954959E+05  
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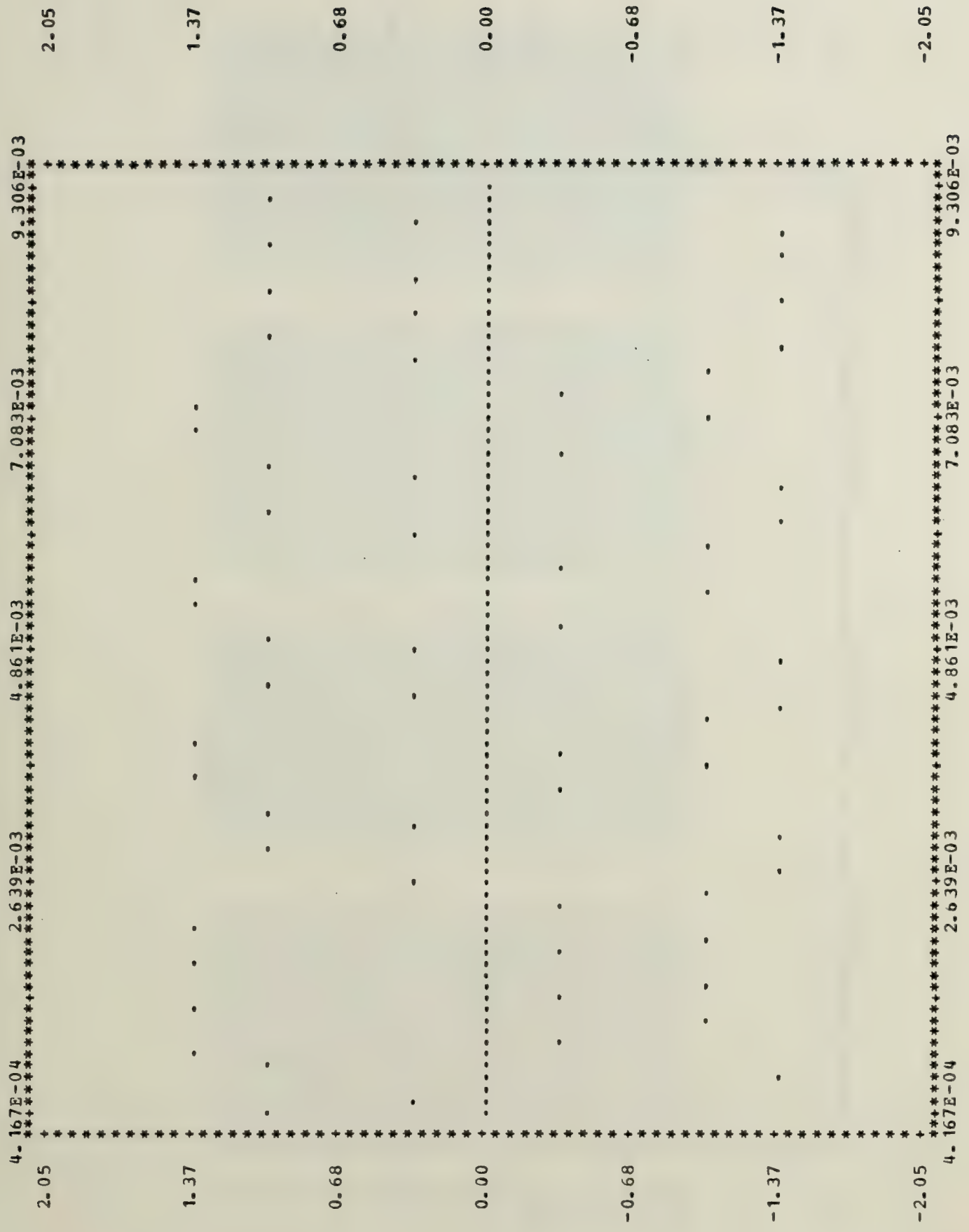
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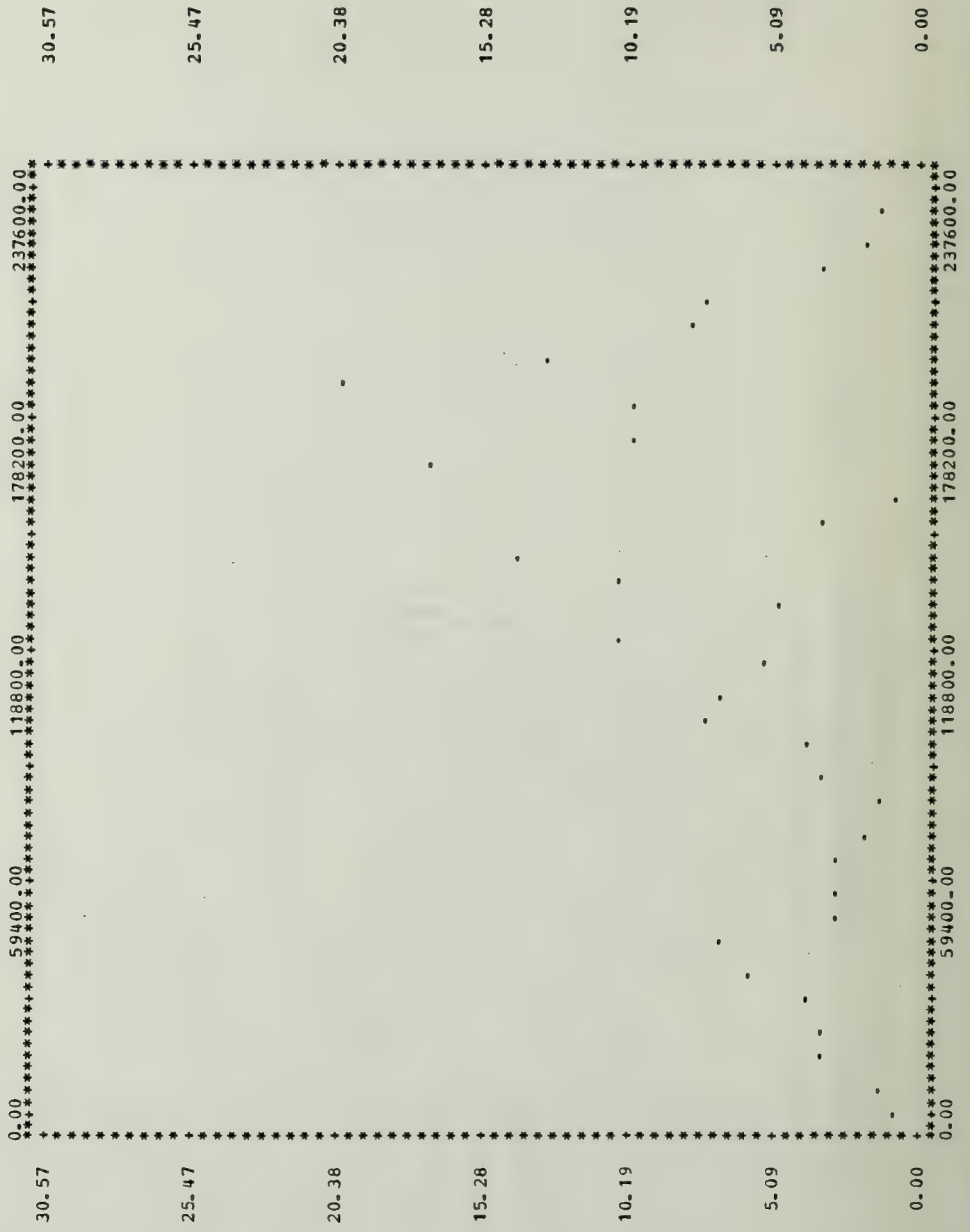
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N 1 2 3 4 5 6 7 8 9 10 11 12 13 14 15 16 17 18 19 20 21 22 23 24 25 26 27 28 29 30 31 32 33

MODULATION TECHNIQUE = OQPSK  
 BIFOLAR LOGIC  
 BAUD OR SYMBOL RATE = 1200 HZ  
 BITS PER BINARY CODE WORD = 2  
 BIT RATE = 0.2400000000000000E+04  
 CARRIER FREQUENCY = 2400 HZ  
 MAXIMUM CARRIER AMPLITUDE = 1 VOLT(S)  
 INITIAL PHASE ANGLE = 0 DEGREES  
 TIME BETWEEN SAMPLES = 0.1388888888888889E-03 SEC  
 NUMBER OF SAMPLES GENERATED = 64  
 SEED FOR RANDOM NUMBER GENERATOR = 1  
 NUMBER OF TIMES SIMULATION REPEATS = 100  
 SUM OF VARIANCES SUM OF VARIANCES\*\*2  
 0.4302799744926702E+03 0.1192673508267923E+05  
 SUM OF SKEWNESS SUM OF SKEWNESS\*\*2  
 -0.2766466990808353E+05 0.6316161942013623E+08  
 SUM OF KURTOSIS SUM OF KURTOSIS\*\*2  
 0.3332128031283272E+05 0.9992700833249297E+08  
 MEAN VARIANCE VARIANCE OF THE VARIANCES  
 0.1303878710583849E+02 0.1973876906051148E+03  
 MEAN SKEWNESS VARIANCE OF THE SKEWNESS  
 -0.8383233305479857E+03 0.1249052537633781E+07  
 MEAN KURTOSIS VARIANCE OF THE KURTOSIS  
 0.1009735767055537E+04 0.2071291243641945E+07





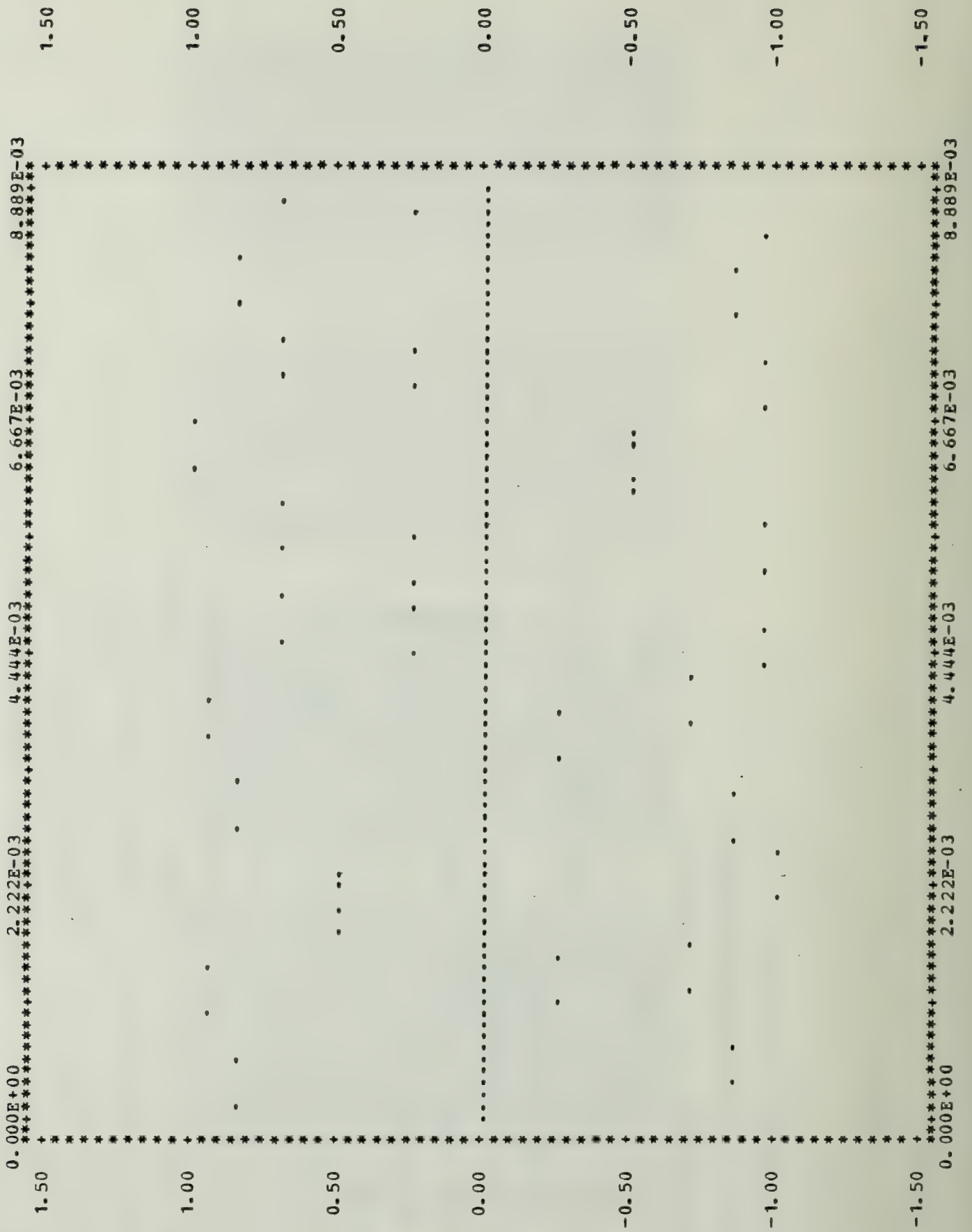


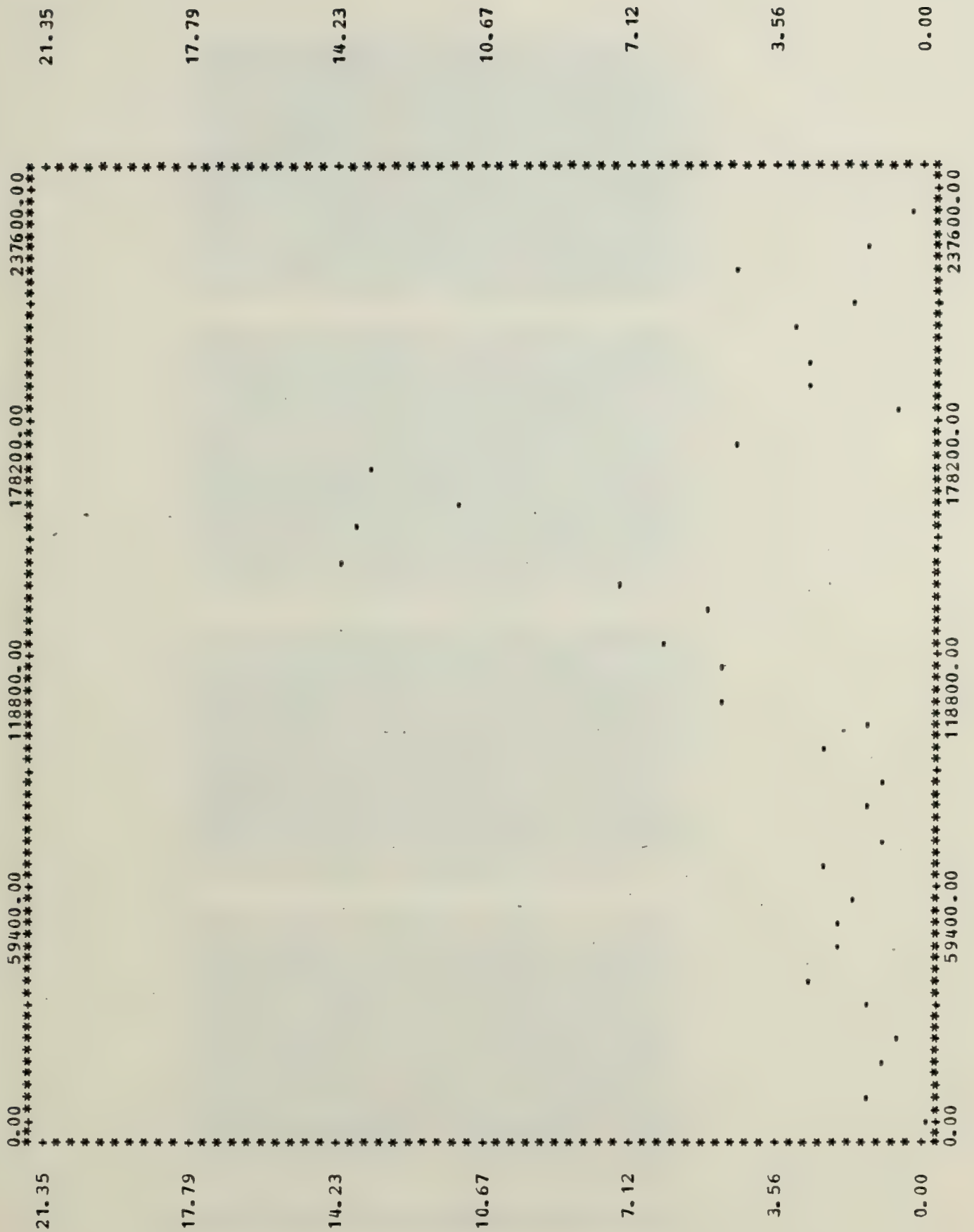






MODULATION TECHNIQUE = MPSK  
 BIFOLAR LOGIC  
 BAUD OR SYMBOL RATE = 1200 HZ  
 BITS PER BINARY CODE WORD = 3  
 BIT RATE = 0.3600000000000000E+04  
 CARRIER FREQUENCY = 2400 HZ  
 MAXIMUM CARRIER AMPLITUDE = 1 VOLT(S)  
 INITIAL PHASE ANGLE = 0 DEGREES  
 TIME BETWEEN SAMPLES = 0.13888888888888889E-03 SEC  
 NUMBER OF SAMPLES GENERATED = 64  
 SEED FOR RANDOM NUMBER GENERATOR = 1  
 NUMBER OF TIMES SIMULATION REPEATS = 100  
 SUM OF VARIANCES SUM OF VARIANCES\*\*2  
 0.2100020931141337E+03 0.2899383925937582E+04  
 SUM OF SKEWNESS SUM OF SKEWNESS\*\*2  
 -0.960095280888690E+04 0.8204154937490084E+07  
 SUM OF KURTOSIS SUM OF KURTOSIS\*\*2  
 0.8272550682771584E+04 0.7152001918873632E+07  
 MEAN VARIANCE VARIANCE OF THE VARIANCES  
 0.6363699791337385E+01 0.4884355155646109E+02  
 MEAN SKEWNESS VARIANCE OF THE SKEWNESS  
 -0.2909392509360209E+03 0.1690890175696266E+06  
 MEAN KURTOSIS VARIANCE OF THE KURTOSIS  
 0.2506833540233813E+03 0.1586940989808761E+06







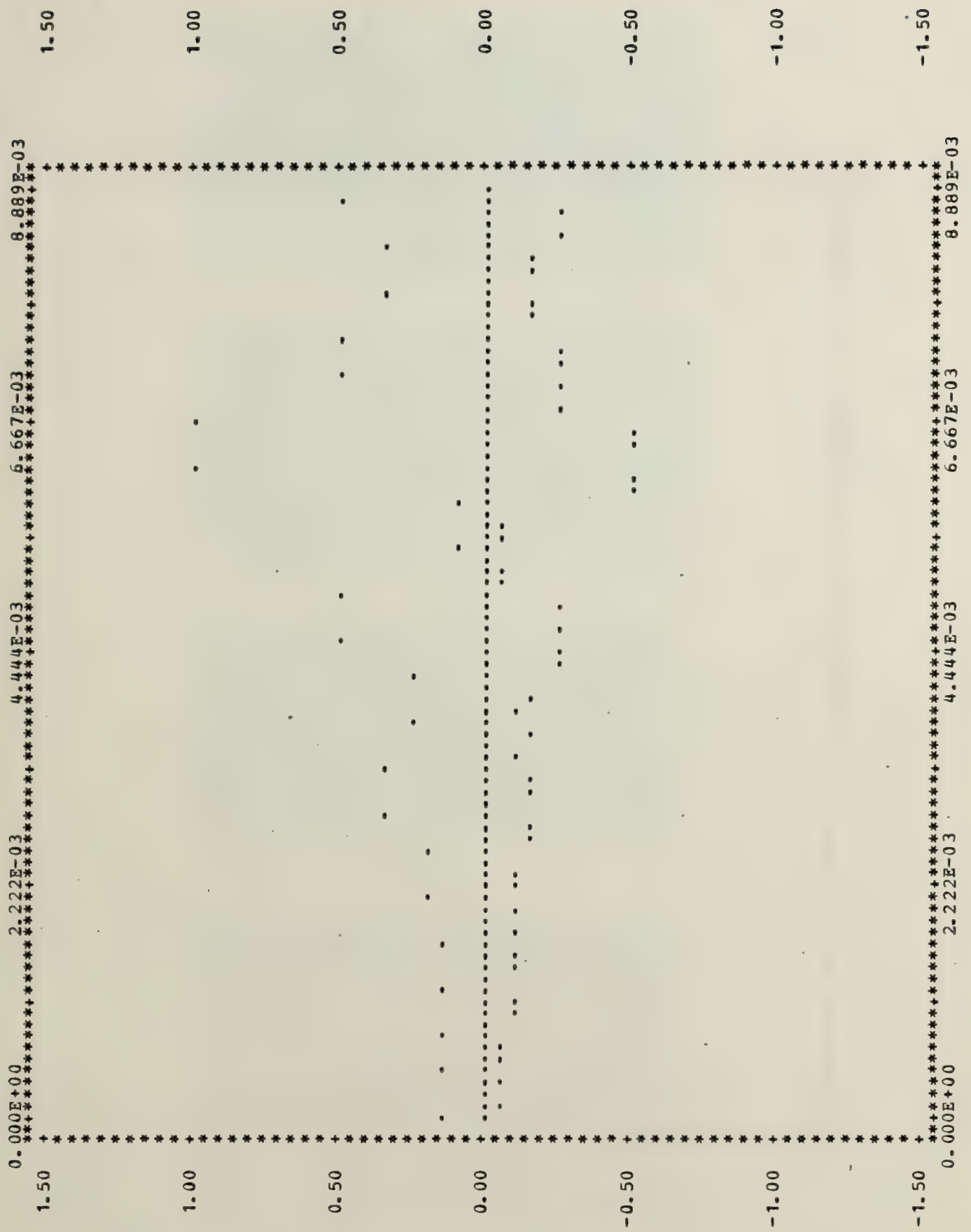
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4	123780	1477847	12149	10759
5	23797	1246925	1119	64714
6	77880	2098145	1453	66723
7	45493	1038070	326	43412
8	33854	8172058	2859	93597
9	54948	4602410	2143	26306
10	19433	4651434	2154	23368
11	19433	4651434	2154	23368
12	17014	3465143	1872	49343
13	79421	8238855	2870	18452
14	15065	1823885	1352	10256
15	5361	3367091	1842	27644
16	10295	5878224	7684	77338
17	19521	7336843	8599	15769
18	97904	1179056	1093	47171
19	17581	4332643	6637	38943
20	48175	2437175	4932	83691
21	33481	1838659	4272	14233
22	16800	1838659	4272	84023
23	41880	1666933	4087	51863
24	18800	1666933	4087	16110
25	99753	1376973	3716	32777
26	18800	8531145	2947	10561
27	67363	2289980	1508	22285
28	96187	4948268	2225	16643
29	33961	8943393	947	92527
30	82339	2289980	1508	17439
31	67363	2289980	1508	22285
32	96187	4948268	2225	16643
33	33961	8943393	947	92527
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36	96187	4948268	2225	16643
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40	96187	4948268	2225	16643
41	33961	8943393	947	92527
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45	33961	8943393	947	92527
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48	96187	4948268	2225	16643
49	33961	8943393	947	92527
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54	82339	2289980	1508	17439
55	67363	2289980	1508	22285
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71	67363	2289980	1508	22285
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79	67363	2289980	1508	22285
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81	33961	8943393	947	92527
82	82339	2289980	1508	17439
83	67363	2289980	1508	22285
84	96187	4948268	2225	16643
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94	82339	2289980	1508	17439
95	67363	2289980	1508	22285
96	96187	4948268	2225	16643
97	33961	8943393	947	92527
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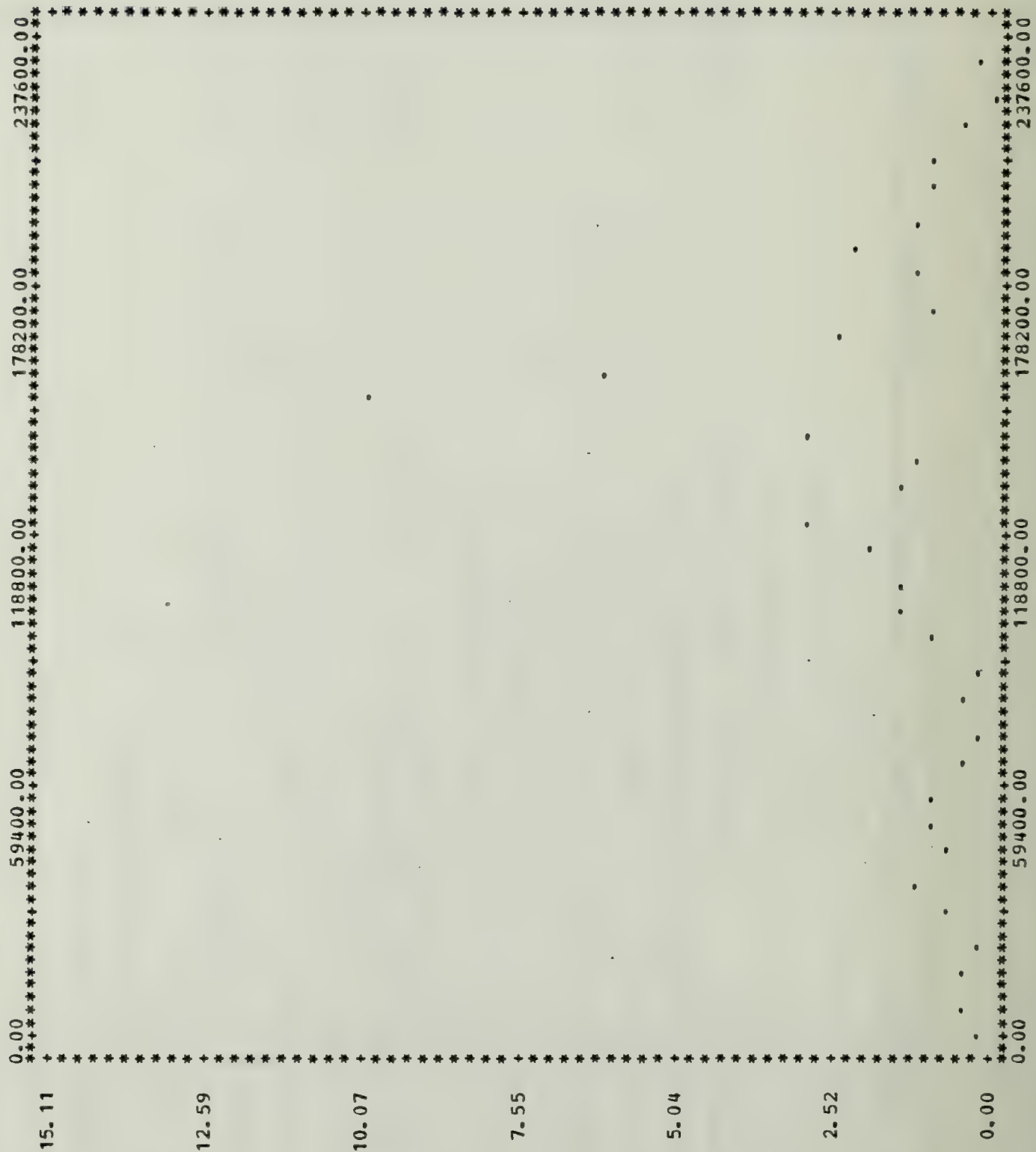
N	Y	SUM	X	Y**2	X**2	XY	Y**3	X**3	XY**2	Y**4	X**4	XY**3		
1	5852	6911	934	45831	17E+02	414	706	9E+02	414	706	9E+02	414	706	9E+02
2	8991	3368	044	1706	9E+03	13	286	4E+03	13	286	4E+03	13	286	4E+03
3	1273	759	207	613	9E+03	15	97	24E+03	15	97	24E+03	15	97	24E+03
4	2493	303	343	163	6E+03	3	43	6E+03	3	43	6E+03	3	43	6E+03
5	2478	804	54	605	6E+03	3	56	7E+03	3	56	7E+03	3	56	7E+03
6	1792	265	82	408	2E+03	3	194	3E+03	3	194	3E+03	3	194	3E+03
7	1338	421	96	551	5E+03	3	272	7E+03	3	272	7E+03	3	272	7E+03
8	1206	794	10	945	5E+03	3	594	3E+03	3	594	3E+03	3	594	3E+03
9	1701	079	214	725	0E+03	3	415	0E+03	3	415	0E+03	3	415	0E+03
10	2834	150	61	551	4E+03	3	61	5E+03	3	61	5E+03	3	61	5E+03
11	3743	807	55	361	8E+03	3	86	3E+03	3	86	3E+03	3	86	3E+03
12	4695	195	02	102	9E+03	3	54	0E+03	3	54	0E+03	3	54	0E+03
13	5610	648	95	155	6E+03	3	74	8E+03	3	74	8E+03	3	74	8E+03
14	6706	908	95	155	6E+03	3	89	7E+03	3	89	7E+03	3	89	7E+03
15	7233	528	19	790	4E+03	3	153	2E+03	3	153	2E+03	3	153	2E+03
16	8344	623	48	153	2E+03	3	210	5E+03	3	210	5E+03	3	210	5E+03
17	8944	133	4	160	9E+03	3	30	4E+03	3	30	4E+03	3	30	4E+03
18	8874	950	97	160	9E+03	3	225	3E+03	3	225	3E+03	3	225	3E+03
19	8248	418	80	159	6E+03	3	263	6E+03	3	263	6E+03	3	263	6E+03
20	7208	201	80	000	7E+03	3	361	7E+03	3	361	7E+03	3	361	7E+03
21	6574	426	94	206	1E+03	3	45	7E+03	3	45	7E+03	3	45	7E+03
22	4540	960	06	187	9E+03	3	34	7E+03	3	34	7E+03	3	34	7E+03
23	3600	823	39	311	1E+03	3	11	1E+03	3	11	1E+03	3	11	1E+03
24	2979	637	26	851	3E+03	3	137	9E+03	3	137	9E+03	3	137	9E+03
25	1788	872	68	851	3E+03	3	157	9E+03	3	157	9E+03	3	157	9E+03
26	1438	892	58	298	1E+03	3	155	7E+03	3	155	7E+03	3	155	7E+03
27	3750	000	000	000	0E+03	3	000	0E+03	3	000	0E+03	3	000	0E+03
28	3750	000	000	000	0E+03	3	000	0E+03	3	000	0E+03	3	000	0E+03
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35	3750	000	000	000	0E+03	3	000	0E+03	3	000	0E+03	3	000	0E+03
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97	3750	000	000	000	0E+03	3	000	0E+03	3					



MODULATION TECHNIQUE = MASK  
 BIFOLAR LOGIC  
 BAUD OR SYMBOL RATE = 1200 HZ  
 BITS PER BINARY CODE WORD = 3  
 BIT RATE = 0.3600000000000000E+04  
 CARRIER FREQUENCY = 2400 HZ  
 MAXIMUM CARRIER AMPLITUDE = 1 VOLT(S)  
 INITIAL PHASE ANGLE = 0 DEGREES  
 TIME BETWEEN SAMPLES = 0.13888888888888889E-03 SEC  
 NUMBER OF SAMPLES GENERATED = 64  
 SEED FOR RANDOM NUMBER GENERATOR = 1  
 NUMBER OF TIMES SIMULATION REPEATS = 100  
 SUM OF VARIANCES SUM OF VARIANCES\*\*2  
 0.2672011399771625E+02 0.6236464612091026E+02  
 SUM OF SKEWNESS SUM OF SKEWNESS\*\*2  
 -0.1657760855944905E+04 0.1880354503890174E+07  
 SUM OF KURTOSIS SUM OF KURTOSIS\*\*2  
 0.1906467874763258E+03 0.7184370776848387E+04  
 MEAN VARIANCE VARIANCE OF THE VARIANCES  
 0.8097004241732196E+00 0.1272792452593831E+01  
 MEAN SKEWNESS VARIANCE OF THE SKEWNESS  
 -0.5023517745287592E+02 0.5615864353491720E+05  
 MEAN KURTOSIS VARIANCE OF THE KURTOSIS  
 0.5777175378070478E+01 0.1900928390728725E+03







N	MEAN	VARIANCE	STDEVES	KURTOSIS
1	339.33	8384.46	92.28	274.27
2	400.40	7668.19	87.54	289.14
3	567.96	14570.17	120.72	155.64
4	716.10	15862.38	126.00	109.05
5	841.35	25063.68	158.32	85.53
6	875.05	24755.48	157.33	36.99
7	881.06	14551.12	120.64	185.24
8	993.22	18574.08	136.35	97.61
9	220.63	16264.02	127.20	185.24
10	695.74	15746.33	125.49	104.85
11	201.31	15805.99	126.14	225.89
12	289.42	22299.14	149.33	168.17
13	259.12	7062.87	84.07	477.97
14	608.95	11450.04	107.00	168.17
15	259.12	13575.07	116.51	747.79
16	580.64	12424.77	111.48	450.68
17	365.98	23046.55	151.82	184.00
18	798.06	49314.57	222.49	440.03
19	817.83	33786.16	184.99	650.79
20	482.22	27933.00	167.24	1752.02
21	365.98	13176.00	114.75	556.22
22	664.73	8497.90	92.28	1879.38
23	282.40	7665.10	87.54	1869.75
24	338.74	15151.86	123.11	343.50
25	993.22	28065.02	167.60	784.61
26	400.40	17633.08	132.74	1669.33
27	800.80	17633.08	132.74	1778.81
28	548.05	6396.25	79.91	2052.39
29	182.33	3396.14	58.28	3922.27
30	338.74	10465.08	102.30	3922.27
31	190.73	8396.14	91.60	2052.39
32	338.74	10465.08	102.30	3922.27
33	338.74	10465.08	102.30	3922.27



N	X	SUM	X**2	X**3	SUM	X*4
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3	3	9	9	27	81	81
4	4	16	16	64	256	256
5	5	25	25	125	625	625
6	6	36	36	216	1296	1296
7	7	49	49	343	2401	2401
8	8	64	64	512	4096	4096
9	9	81	81	729	6561	6561
10	10	100	100	1000	10000	10000
11	11	121	121	1331	14641	14641
12	12	144	144	1728	20736	20736
13	13	169	169	2197	28549	28549
14	14	196	196	2744	38416	38416
15	15	225	225	3375	50625	50625
16	16	256	256	4096	65536	65536
17	17	289	289	4913	83521	83521
18	18	324	324	5832	104976	104976
19	19	361	361	6859	129679	129679
20	20	400	400	8000	160000	160000
21	21	441	441	9261	194481	194481
22	22	484	484	10648	232416	232416
23	23	529	529	12167	274279	274279
24	24	576	576	13824	320064	320064
25	25	625	625	15625	369775	369775
26	26	676	676	17568	423456	423456
27	27	729	729	19659	481127	481127
28	28	784	784	21904	542800	542800
29	29	841	841	24309	608401	608401
30	30	900	900	26970	678000	678000
31	31	961	961	29881	751619	751619
32	32	1024	1024	32048	829264	829264
33	33	1089	1089	34485	911049	911049
34	34	1156	1156	37208	997024	997024
35	35	1225	1225	40235	1097255	1097255
36	36	1296	1296	43568	1211808	1211808
37	37	1369	1369	47217	1340841	1340841
38	38	1444	1444	51184	1484336	1484336
39	39	1521	1521	55471	1642455	1642455
40	40	1600	1600	60080	1815280	1815280
41	41	1681	1681	65011	1902921	1902921
42	42	1764	1764	70264	2005376	2005376
43	43	1849	1849	75849	2122745	2122745
44	44	1936	1936	81764	2255128	2255128
45	45	2025	2025	88009	2402625	2402625
46	46	2116	2116	94584	2565336	2565336
47	47	2209	2209	101489	2743361	2743361
48	48	2304	2304	108724	2936800	2936800
49	49	2401	2401	116289	3145855	3145855
50	50	2500	2500	124194	3370624	3370624
51	51	2601	2601	132449	3612205	3612205
52	52	2704	2704	141054	3870704	3870704
53	53	2809	2809	150009	4146225	4146225
54	54	2916	2916	159324	4438872	4438872
55	55	3025	3025	169009	4748745	4748745
56	56	3136	3136	179064	5075944	5075944
57	57	3249	3249	189499	5420565	5420565
58	58	3364	3364	200324	5782704	5782704
59	59	3481	3481	211549	6162455	6162455
60	60	3600	3600	223184	6569920	6569920
61	61	3721	3721	235229	7005205	7005205
62	62	3844	3844	247694	7468416	7468416
63	63	3969	3969	260589	7959645	7959645
64	64	4096	4096	273924	8479000	8479000
65	65	4225	4225	287709	9027585	9027585
66	66	4356	4356	301954	9605504	9605504
67	67	4489	4489	316669	10213855	10213855
68	68	4624	4624	331864	10852864	10852864
69	69	4761	4761	347549	11522635	11522635
70	70	4900	4900	363734	12224272	12224272
71	71	5041	5041	380429	12958875	12958875
72	72	5184	5184	397644	13727544	13727544
73	73	5329	5329	415379	14530375	14530375
74	74	5476	5476	433644	15368464	15368464
75	75	5625	5625	452449	16241905	16241905
76	76	5776	5776	471704	17151892	17151892
77	77	5929	5929	491429	18098525	18098525
78	78	6084	6084	511634	19081904	19081904
79	79	6241	6241	532329	20103125	20103125
80	80	6400	6400	553524	21162296	21162296
81	81	6561	6561	575229	22260515	22260515
82	82	6724	6724	597454	23398880	23398880
83	83	6889	6889	620209	24578485	24578485
84	84	7056	7056	643504	25799424	25799424
85	85	7225	7225	667349	27061895	27061895
86	86	7396	7396	691744	28366000	28366000
87	87	7569	7569	716689	29712845	29712845
88	88	7744	7744	742194	31112524	31112524
89	89	7921	7921	768269	32565135	32565135
90	90	8100	8100	794924	34070780	34070780
91	91	8281	8281	822169	35639565	35639565
92	92	8464	8464	849904	37271592	37271592
93	93	8649	8649	878149	38966965	38966965
94	94	8836	8836	906904	40725784	40725784
95	95	9025	9025	936179	42548145	42548145
96	96	9216	9216	965984	44434152	44434152
97	97	9409	9409	996329	46383905	46383905
98	98	9604	9604	1027214	48397408	48397408
99	99	9801	9801	1058629	50474765	50474765
100	100	10000	10000	1090554	52616072	52616072

MODULATION TECHNIQUE = QASK  
BIFOLAR LOGIC

BAUD OR SYMBOL RATE = 1200 HZ

BITS PER BINARY CODE WORD = 3

BIT RATE = 0.3600000000000000E+04

CARRIER FREQUENCY = 2400 HZ

MAXIMUM CARRIER AMPLITUDE = 1 VOLT(S)

INITIAL PHASE ANGLE = 0 DEGREES

TIME BETWEEN SAMPLES = 0.13888888888888889E-03 SEC

NUMBER OF SAMPLES GENERATED = 64

SEED FOR RANDOM NUMBER GENERATOR = 1

NUMBER OF TIMES SIMULATION REPEATS = 100

SUM OF VARIANCES SUM OF VARIANCES\*\*2

0.4885105807865026E+02 0.1949104911522201E+03

SUM OF SKEWNESS SUM OF SKEWNESS\*\*2

-0.5438333348597472E+04 0.1951490781083542E+08

SUM OF KURTOSIS SUM OF KURTOSIS\*\*2

0.6866043785868234E+03 0.9002560164061202E+05

MEAN VARIANCE VARIANCE OF THE VARIANCES

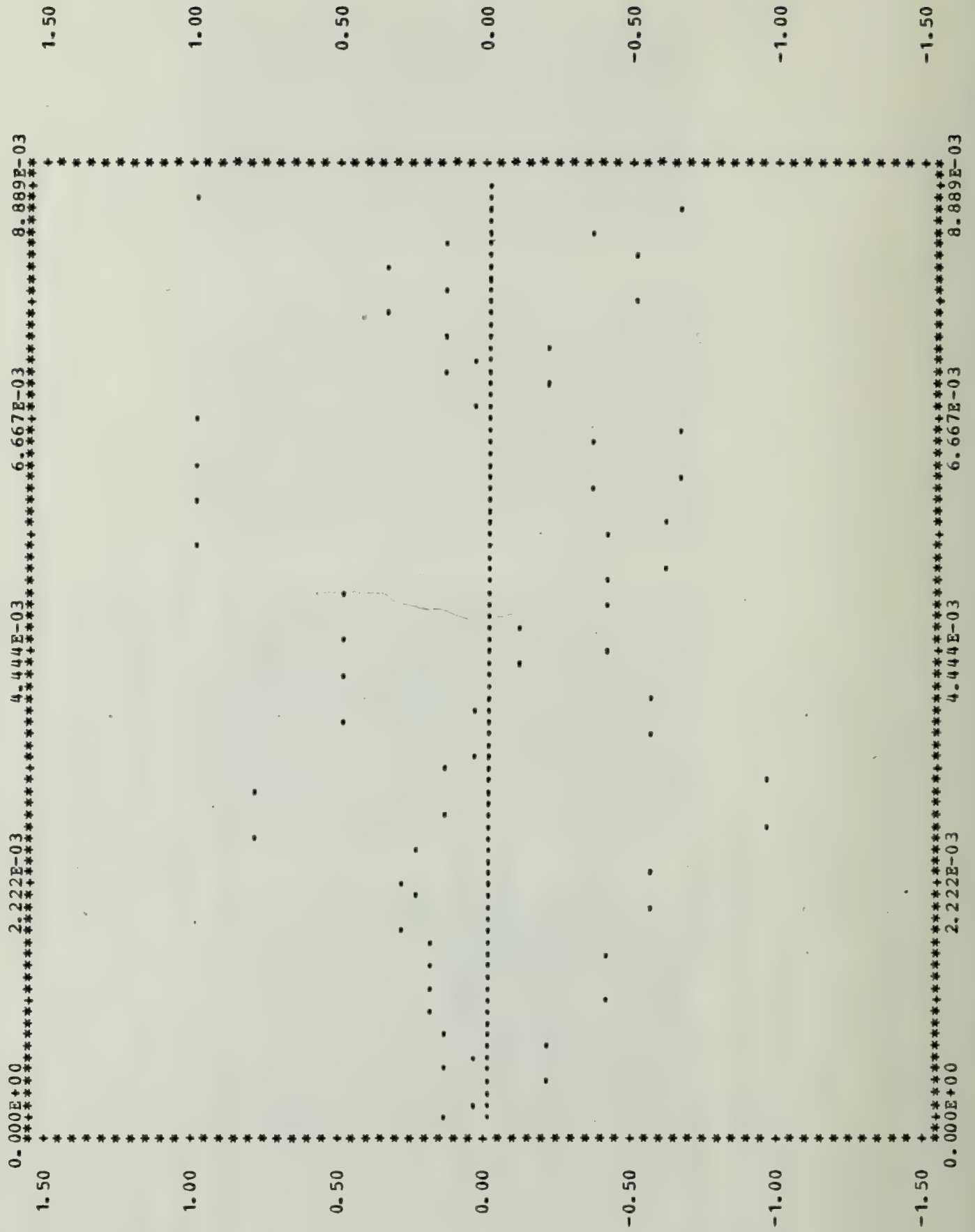
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MEAN SKEWNESS VARIANCE OF THE SKEWNESS

-0.1647979802605294E+03 0.5818337955938460E+06

MEAN KURTOSIS VARIANCE OF THE KURTOSIS

0.2080619329050980E+02 0.2366874319550756E+04







N	MEAN	VARIANCE	SKENNESS	KURTOSIS
1	363.1071	9.100003	4.487644	27.365509
2	434.8756	10.774029	4.499936	30.993481
3	457.6835	10.447298	4.399348	31.876680
4	562.3573	13.986694	2.812110	45.286947
5	662.4733	22.370622	1.777804	88.719883
6	665.3553	28.219210	1.275999	125.919883
7	605.3546	28.219210	1.275999	125.919883
8	605.3546	28.219210	1.275999	125.919883
9	605.3546	28.219210	1.275999	125.919883
10	605.3546	28.219210	1.275999	125.919883
11	605.3546	28.219210	1.275999	125.919883
12	605.3546	28.219210	1.275999	125.919883
13	605.3546	28.219210	1.275999	125.919883
14	605.3546	28.219210	1.275999	125.919883
15	605.3546	28.219210	1.275999	125.919883
16	605.3546	28.219210	1.275999	125.919883
17	605.3546	28.219210	1.275999	125.919883
18	605.3546	28.219210	1.275999	125.919883
19	605.3546	28.219210	1.275999	125.919883
20	605.3546	28.219210	1.275999	125.919883
21	605.3546	28.219210	1.275999	125.919883
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23	605.3546	28.219210	1.275999	125.919883
24	605.3546	28.219210	1.275999	125.919883
25	605.3546	28.219210	1.275999	125.919883
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31	605.3546	28.219210	1.275999	125.919883
32	605.3546	28.219210	1.275999	125.919883
33	605.3546	28.219210	1.275999	125.919883

N	Y	SUM	X	X**2	SUM	Y	SUM	X**3	SUM	X**4	SUM
1	1	3631	0714	2857	19368	33900	1313	99386	10336	1696	5808
2	2	6750	8992	6835	5799	5430	3737	47906	6559	2969	1083
3	3	7480	7862	5863	4733	6931	2222	26227	8536	3444	1587
4	4	9221	2664	6605	3553	0770	5587	56586	2013	2146	2861
5	5	1026	1035	2308	7546	2052	4329	10475	8888	3215	3297
6	6	1020	2814	1710	7785	2052	4329	46587	1643	2907	3297
7	7	9039	6039	1012	9196	5861	3433	56599	9048	5507	4808
8	8	6921	1235	6938	1457	4052	1648	5910	7008	4808	4808
9	9	6486	2673	5693	2249	4052	1648	7769	1481	3333	1250
10	10	8754	0973	5693	2249	4052	1648	10682	3662	5133	1068
11	11	1209	7219	5933	2602	4052	1648	15186	6624	6555	1965
12	12	1588	1095	2214	8007	4052	1648	18465	9690	1945	1945
13	13	1906	9746	0908	4911	4052	1648	20925	1250	4668	1017
14	14	2352	8799	1704	1219	4052	1648	23350	1807	1017	1990
15	15	3101	2768	5151	7939	4052	1648	25232	2599	1615	1615
16	16	4176	9672	1899	3768	4052	1648	27272	3362	2523	6886
17	17	3698	6348	1545	2042	4052	1648	29272	4242	4003	1003
18	18	4176	9672	1899	3768	4052	1648	31305	5067	5120	5120
19	19	3101	2768	5151	7939	4052	1648	33357	5991	6886	1038
20	20	4176	9672	1899	3768	4052	1648	35454	6990	8862	1400
21	21	1333	6459	6182	4457	4052	1648	37525	7754	1038	1038
22	22	6989	5735	3017	1479	4052	1648	39699	8707	1400	1400
23	23	4042	8553	3842	4669	4052	1648	41922	9794	1769	1769
24	24	3311	5450	4591	5114	4052	1648	44205	1095	2220	3044
25	25	3373	0867	0960	4516	4052	1648	46548	1224	3044	3044
26	26	2677	6251	1047	1693	4052	1648	48951	1367	3521	5120
27	27	1969	7481	6718	4020	4052	1648	51418	1527	4154	1311
28	28	1366	9607	3050	6631	4052	1648	53952	1704	5120	1704
29	29	1184	9962	6750	7274	4052	1648	56559	1899	6252	1004
30	30	8700	9769	4709	1932	4052	1648	59233	2099	7273	1311
31	31	6506	2288	2227	2112	4052	1648	61977	2322	8252	1260
32	32	3631	0714	2857	19368	33900	1313	64790	2599	9999	1260
33	33	6750	8992	6835	5799	5430	3737	67669	2922	1260	1260



MODULATION TECHNIQUE = MSK  
BIFOLAR LOGIC

BAUD OR SYMBOL RATE = 1200 HZ

BITS PER BINARY CODE WORD = 1

BIT RATE = 0.1200000000000000E+04

CARRIER FREQUENCY = 3000 HZ

MAXIMUM CARRIER AMPLITUDE = 1 VOLT(S)

INITIAL PHASE ANGLE = 0 DEGREES

TIME BETWEEN SAMPLES = 0.1111111111111111E-03 SEC

NUMBER OF SAMPLES GENERATED = 64

SEED FOR RANDOM NUMBER GENERATOR = 1

NUMBER OF TIMES SIMULATION REPEATS = 100

SUM OF VARIANCES SUM OF VARIANCES\*\*2

0.1525890150844788E+03 0.2084564025759793E+04

SUM OF SKEWNESS SUM OF SKEWNESS\*\*2

-0.1665578258850635E+05 0.9704853819233804E+08

SUM OF KURTOSIS SUM OF KURTOSIS\*\*2

0.5423666237610907E+04 0.6118691548293858E+07

MEAN VARIANCE VARIANCE OF THE VARIANCES

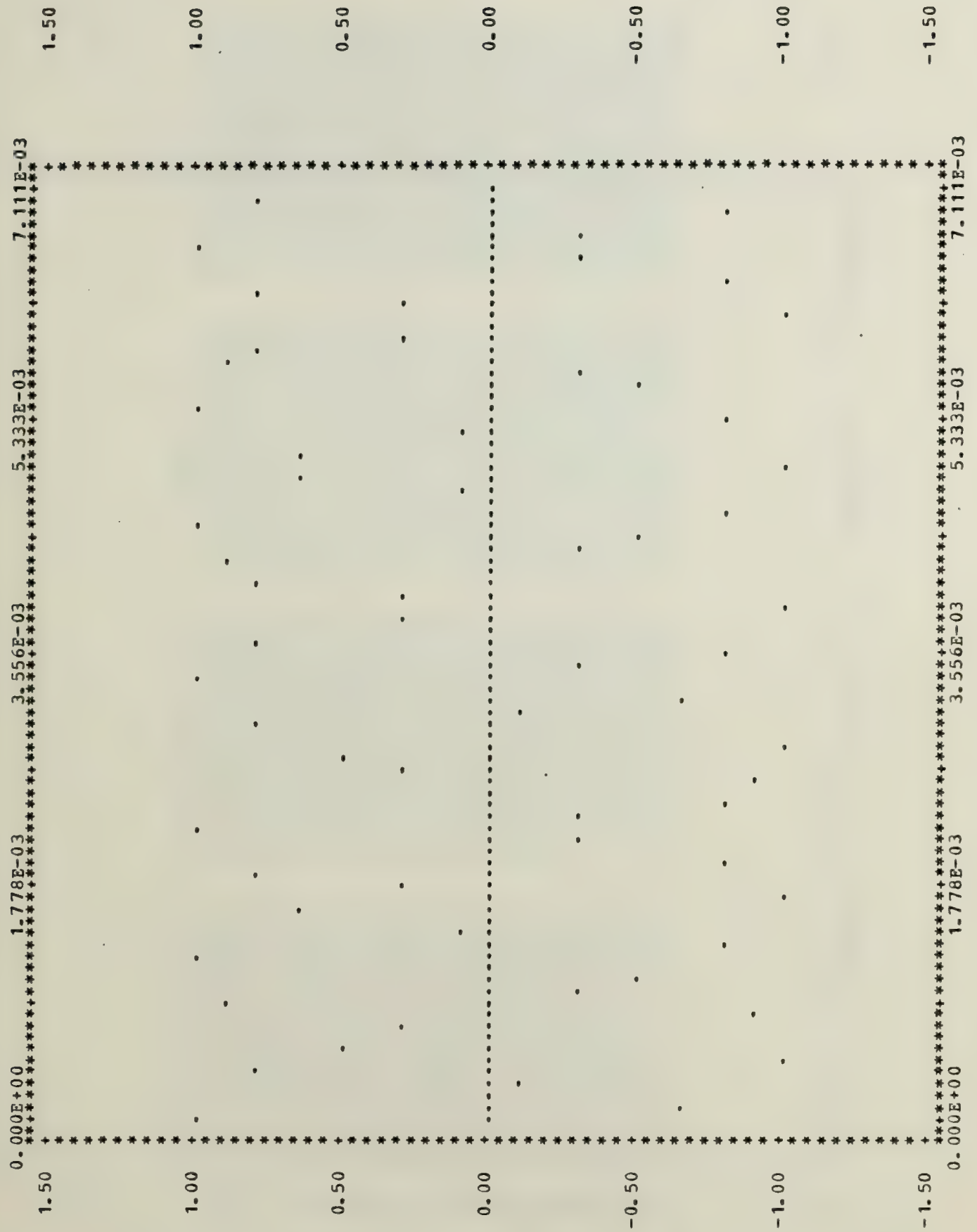
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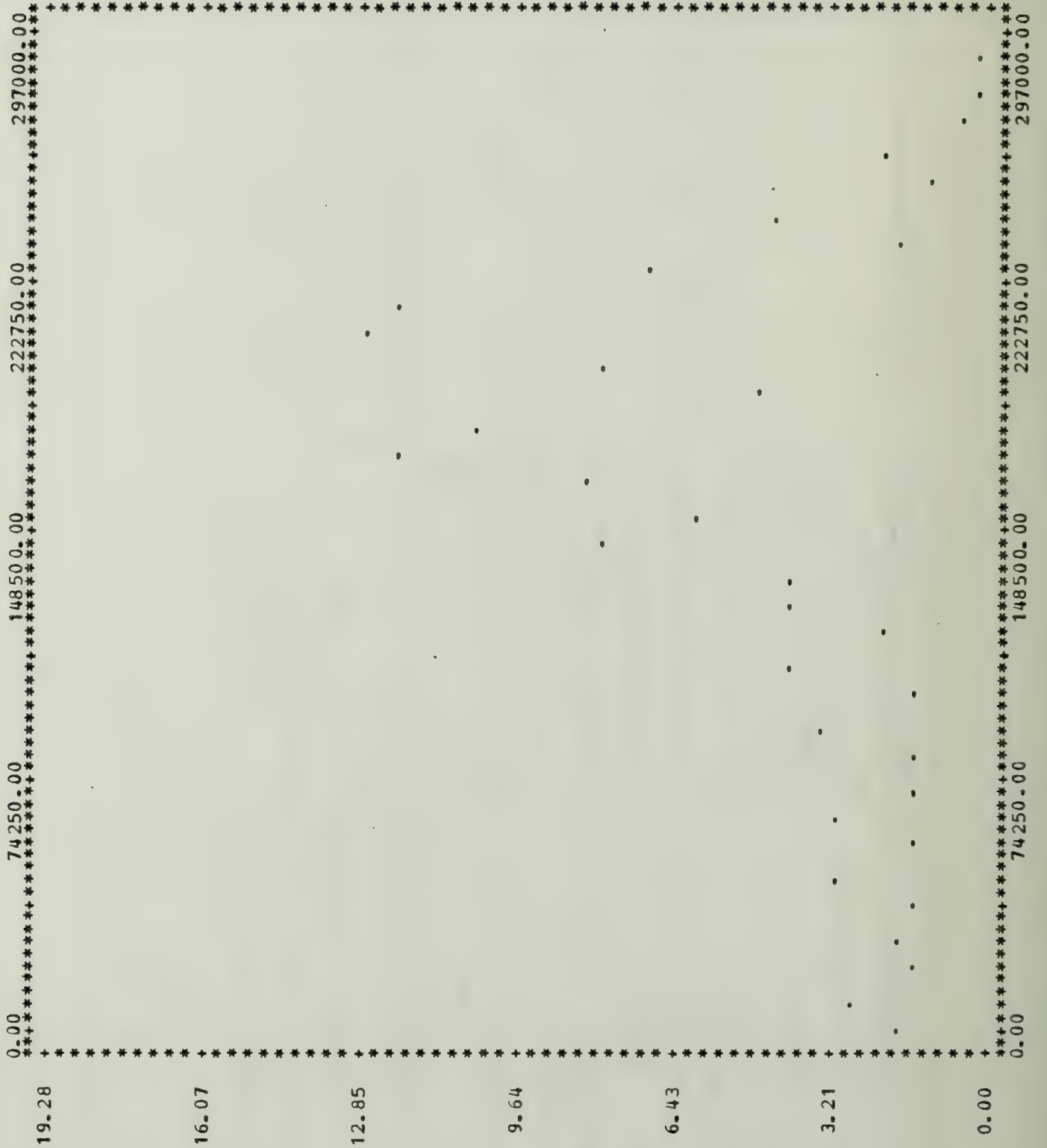
MEAN SKEWNESS VARIANCE OF THE SKEWNESS

-0.5047206845001924E+03 0.2770063131355647E+07

MEAN KURTOSIS VARIANCE OF THE KURTOSIS

0.1643535223518457E+03 0.1633529030650538E+06





19.28  
16.07  
12.85  
9.64  
6.43  
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0.00

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N	X	SUM	X**2	SUM	X**3	SUM	X**4
1	1745381057261094E+03	4388378241030927E+03	30761659528E+03	1346110434953155E+04	4708163227179921E+04		
2	1670893117553454E+03	33761733061659528E+03	110286843353743E+03	445903233160757E+03	2034919190335989E+04		
3	1661199892947162E+03	343119113663228693E+03	1191136628693E+03	447784476187219E+03	2026599445555989E+04		
4	17891680985469851E+03	3376884874075052E+03	4376884874075052E+03	1283686155661075E+04	0-2132708178344390E+04		
5	1793980985469851E+03	38822425787684695E+03	4376884874075052E+03	0-4297079886762252E+04	0-4297079886762252E+04		
6	174419801285398217E+03	3874892151236307E+03	4376884874075052E+03	0-2394998604752789E+04	0-2394998604752789E+04		
7	20534884746733196E+03	51683383574529951E+03	4376884874075052E+03	0-42224573323385694E+04	0-42224573323385694E+04		
8	2088844746733196E+03	23121263939239922E+03	4376884874075052E+03	0-47726253224795818E+04	0-47726253224795818E+04		
9	2078335081298121E+03	516903233939239922E+03	4376884874075052E+03	0-4284140702659404E+04	0-4284140702659404E+04		
10	1846376751847116E+03	45003233939239922E+03	4376884874075052E+03	0-4050916330073661E+04	0-4050916330073661E+04		
11	224594649482377E+03	6227184396690914E+04	120424635892330E+04	0-6729137619289132E+04	0-6729137619289132E+04		
12	2963675540203763E+03	1251601271582480E+04	170798006002793E+04	0-2377476116996190E+05	0-2377476116996190E+05		
13	3173784304747807E+03	170798006002793E+04	261264880821186E+04	0-2854218554507767E+05	0-2854218554507767E+05		
14	346563279745810E+03	4614191970595573E+03	6056467485580745E+03	0-88878144309945281E+05	0-88878144309945281E+05		
15	4614191970595573E+03	6056467485580745E+03	4682979409916067E+04	0-1322897499994962E+06	0-1322897499994962E+06		
16	6056467485580745E+03	48787758748546135E+04	48787758748546135E+04	0-4402476375446830E+06	0-4402476375446830E+06		
17	64968791308735019E+04	3168438719176228E+04	3168438719176228E+04	0-3915963851799988E+06	0-3915963851799988E+06		
18	668321329902640E+03	521451582724333E+04	521451582724333E+04	0-5722380556849635E+06	0-5722380556849635E+06		
19	6465666718341274E+04	4910870028811991E+04	4910870028811991E+04	0-4972868019579191E+06	0-4972868019579191E+06		
20	116454812650914E+04	146054259297951E+05	146054259297951E+05	0-3879520293056626E+06	0-3879520293056626E+06		
21	1044373719208894E+03	3198338530726825E+04	3198338530726825E+04	0-21622824361492207E+07	0-21622824361492207E+07		
22	498105111985246E+03	139669130804281E+04	139669130804281E+04	0-1814011642212024E+06	0-1814011642212024E+06		
23	41233269596834735E+03	98017963297371694E+03	98017963297371694E+03	0-8978341400784285E+05	0-8978341400784285E+05		
24	286193571105750E+03	4147963297371694E+03	4147963297371694E+03	0-1569116396929776E+05	0-1569116396929776E+05		
25	1309968810623288E+02	2085390938171418E+03	2085390938171418E+03	0-33776698865106972E+04	0-33776698865106972E+04		
26	86126838968396291E+02	9649084057365937E+02	9649084057365937E+02	0-7582535161578267E+03	0-7582535161578267E+03		
27	6043391628396291E+02	4436495121812613E+02	4436495121812613E+02	0-30977912706858261E+03	0-30977912706858261E+03		
28	5680265225039E+02	371233360301779964E+02	371233360301779964E+02	0-1928732784350291E+02	0-1928732784350291E+02		



MODULATION TECHNIQUE = MFSK

BIPOLAR LOGIC

BAUD OR SYMBOL RATE = 1200 HZ

BITS PER BINARY CODE WORD = 3

BIT RATE = 0.3600000000000000E+04

CARRIER FREQUENCY = 10800 HZ

MAXIMUM CARRIER AMPLITUDE = 1 VOLT(S)

INITIAL PHASE ANGLE = 0 DEGREES

TIME BETWEEN SAMPLES = 0.3086419753086419E-04 SEC

NUMBER OF SAMPLES GENERATED = 64

SEED FOR RANDOM NUMBER GENERATOR = 1

NUMBER OF TIMES SIMULATION REPEATS = 100

SUM OF VARIANCES SUM OF VARIANCES\*\*2

0.4806181141925261E+03 0.1352273679574422E+05

SUM OF SKEWNESS SUM OF SKEWNESS\*\*2

-0.1654247621695240E+04 0.5890267815807785E+06

SUM OF KURTOSIS SUM OF KURTOSIS\*\*2

0.5341287673934422E+05 0.2790612658473240E+09

MEAN VARIANCE VARIANCE OF THE VARIANCES

0.1456418527856140E+02 0.2038414228878589E+03

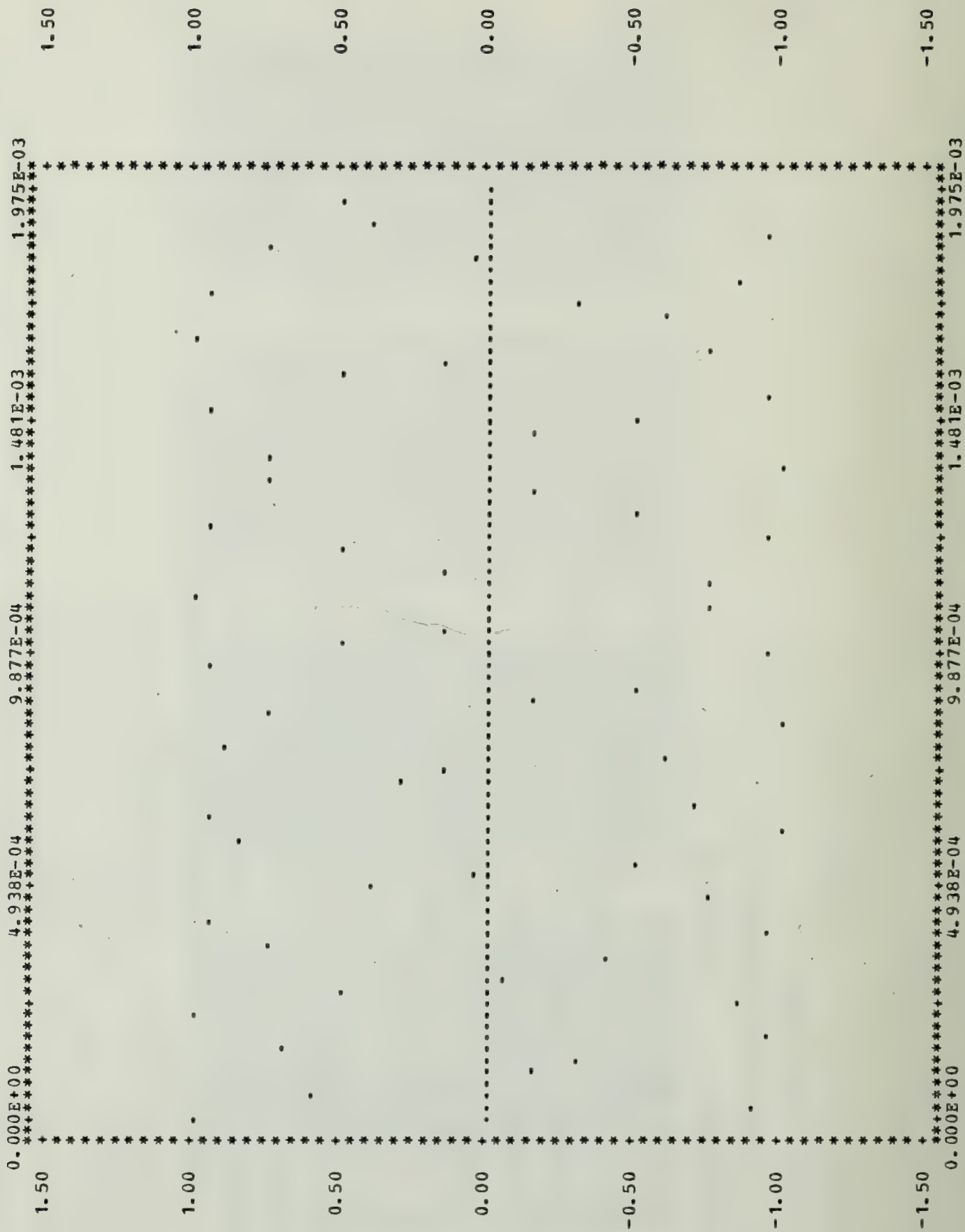
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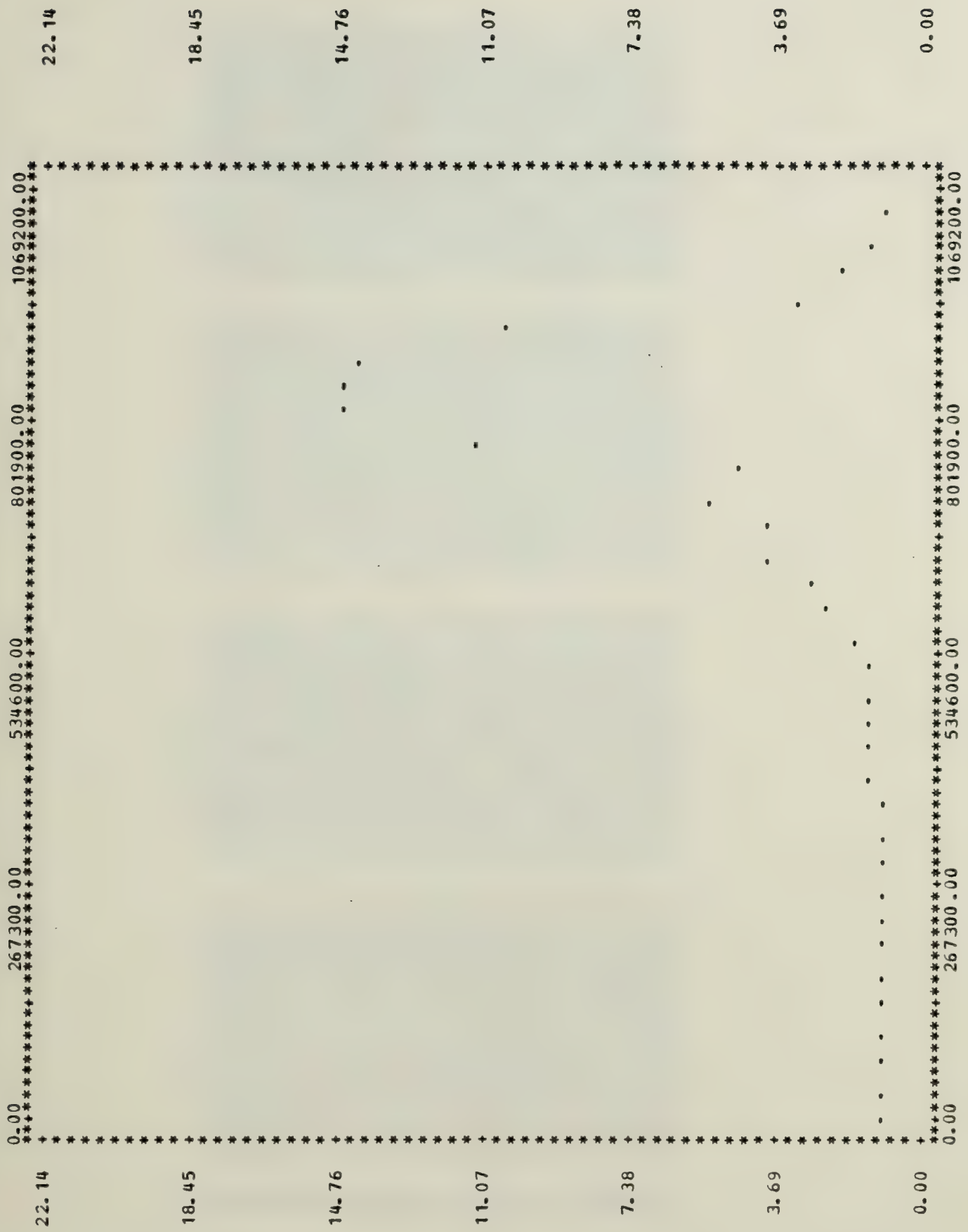
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MEAN KURTOSIS VARIANCE OF THE KURTOSIS

0.1618572022404370E+04 0.6019021185027759E+07





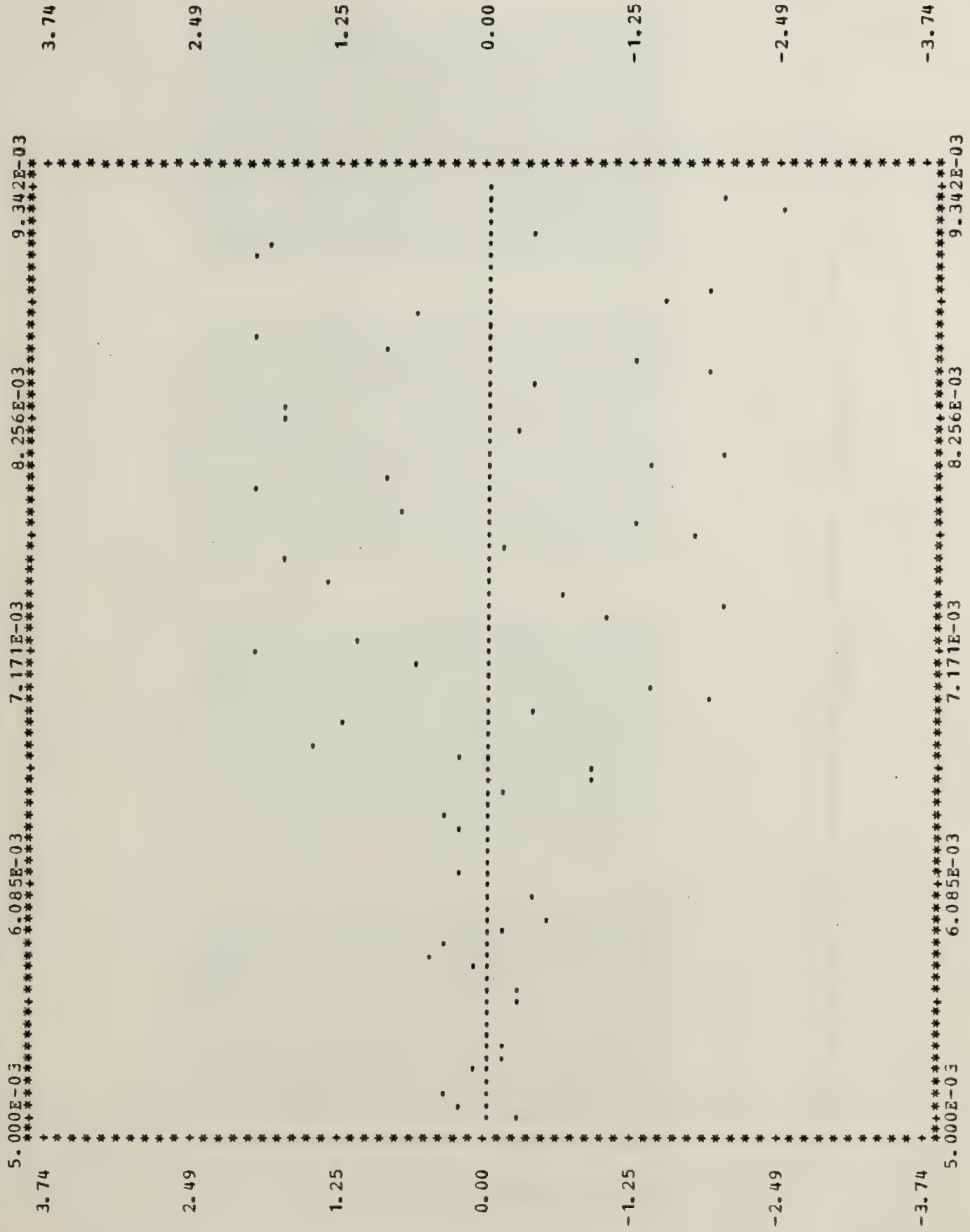


N	MEAN	VARIANCE	SKENNESS	KURTOSIS
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2	19943	1692179	187009	165777
3	159049	308445	209809	771429
4	75085	43860	71176	99494
5	45775	33654	11769	23902
6	38126	151743	2293	40605
7	73358	154131	2279	53725
8	58691	162527	3015	33873
9	11352	198945	15803	90794
10	23357	215853	16717	40007
11	18079	20633	16258	75947
12	40239	32533	17030	4112
13	36641	59014	14728	4444
14	16891	10064	4522	3080
15	17891	44989	5259	4853
16	33091	86046	2225	2698
17	17891	44989	5259	4853
18	33091	86046	2225	2698
19	17891	44989	5259	4853
20	33091	86046	2225	2698
21	17891	44989	5259	4853
22	33091	86046	2225	2698
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29	17891	44989	5259	4853
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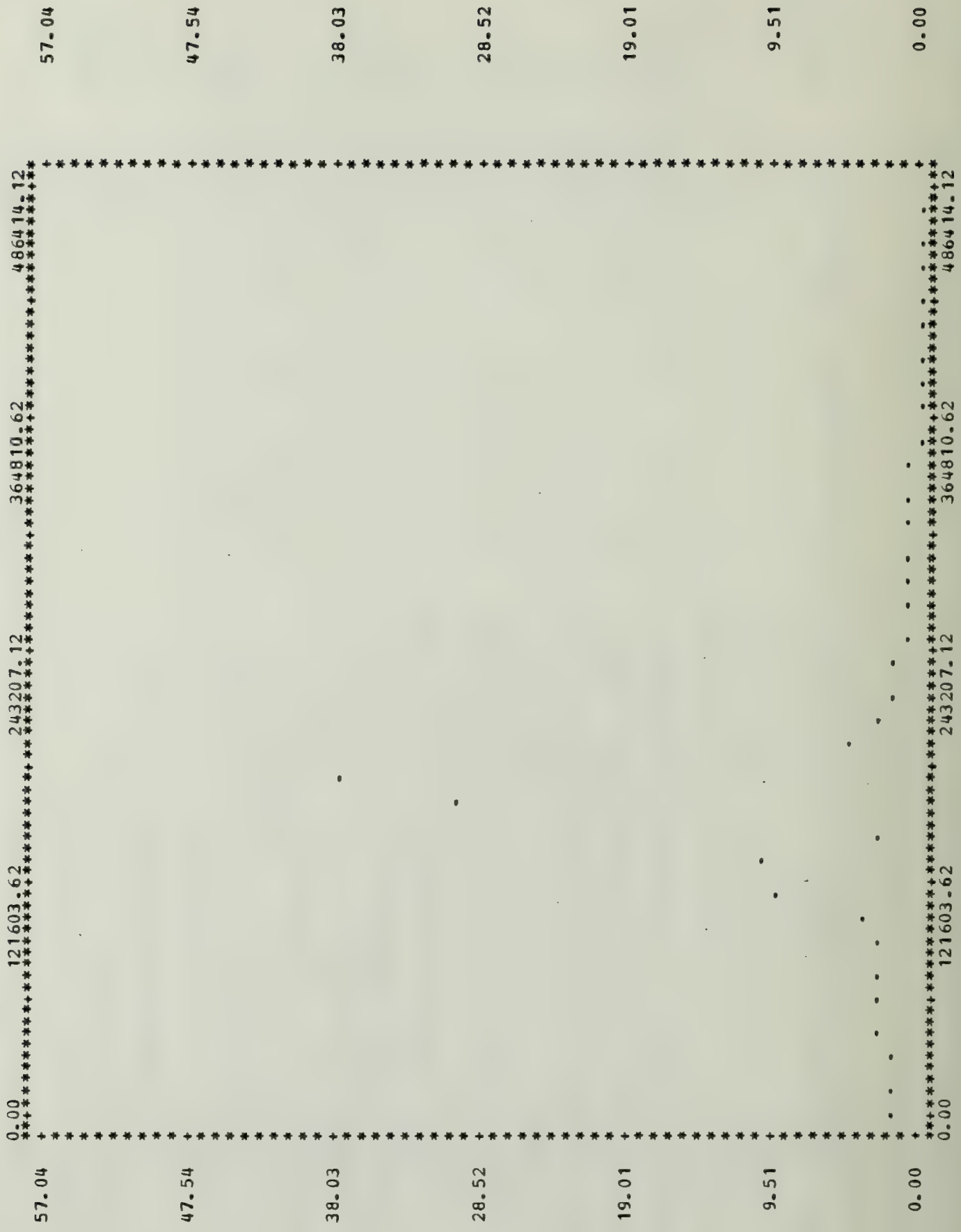


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4	30	900	5500	18210	536280
5	31	960	5999	19575	583959
6	32	1020	6536	21104	634368
7	33	1080	7111	22803	687527
8	34	1140	7724	24678	743456
9	35	1200	8375	26735	802265
10	36	1260	9064	28980	864072
11	37	1320	9791	31419	928985
12	38	1380	10556	34058	997112
13	39	1440	11363	36903	1068463
14	40	1500	12212	40060	1143040
15	41	1560	13103	43535	1220971
16	42	1620	14036	47344	1302368
17	43	1680	15011	51493	1387341
18	44	1740	16028	55998	1475992
19	45	1800	17087	60865	1568433
20	46	1860	18188	66100	1664672
21	47	1920	19331	71719	1764817
22	48	1980	20516	77728	1868972
23	49	2040	21743	84143	1977243
24	50	2100	23012	90970	2089728
25	51	2160	24323	98215	2206535
26	52	2220	25676	105886	2327768
27	53	2280	27071	114005	2453533
28	54	2340	28508	122580	2583944
29	55	2400	29987	131619	2719107
30	56	2460	31508	141130	2859136
31	57	2520	33071	151121	2994137
32	58	2580	34676	161590	3134216
33	59	2640	36323	172545	3279481

MODULATION TECHNIQUE = QPRS  
 QPS CLASS 1 FILTER  
 BIFOLAR LOGIC  
 BAUD OR SYMBOL RATE = 1200 HZ  
 BITS PER BINARY CODE WORD = 2  
 BIT RATE = 0.2400000000000000E+04  
 CARRIER FREQUENCY = 2400 HZ  
 MAXIMUM CARRIER AMPLITUDE = 1 VOLT(S)  
 INITIAL PHASE ANGLE = 0 DEGREES  
 TIME BETWEEN SAMPLES = 0.6784342273548912E-04 SEC  
 NUMBER OF SAMPLES GENERATED = 64  
 SEED FOR RANDOM NUMBER GENERATOR = 1  
 NUMBER OF TIMES SIMULATION REPEATS = 100  
 SUM OF VARIANCES SUM OF VARIANCES\*\*2  
 0.7925891201989905E+03 0.1512135267846902E+06  
 SUM OF SKEWNESS SUM OF SKEWNESS\*\*2  
 -0.1732359062524069E+06 0.9850778236273860E+10  
 SUM OF KURTOSIS SUM OF KURTOSIS\*\*2  
 0.3533192079165041E+06 0.4536060891356758E+11  
 MEAN VARIANCE VARIANCE OF THE VARIANCES  
 0.2401785212724214E+02 0.4130538703065308E+04  
 MEAN SKEWNESS VARIANCE OF THE SKEWNESS  
 -0.5249572916739602E+04 0.2794176160813870E+09  
 MEAN KURTOSIS VARIANCE OF THE KURTOSIS  
 0.1070664266413649E+05 0.1299304575250932E+10





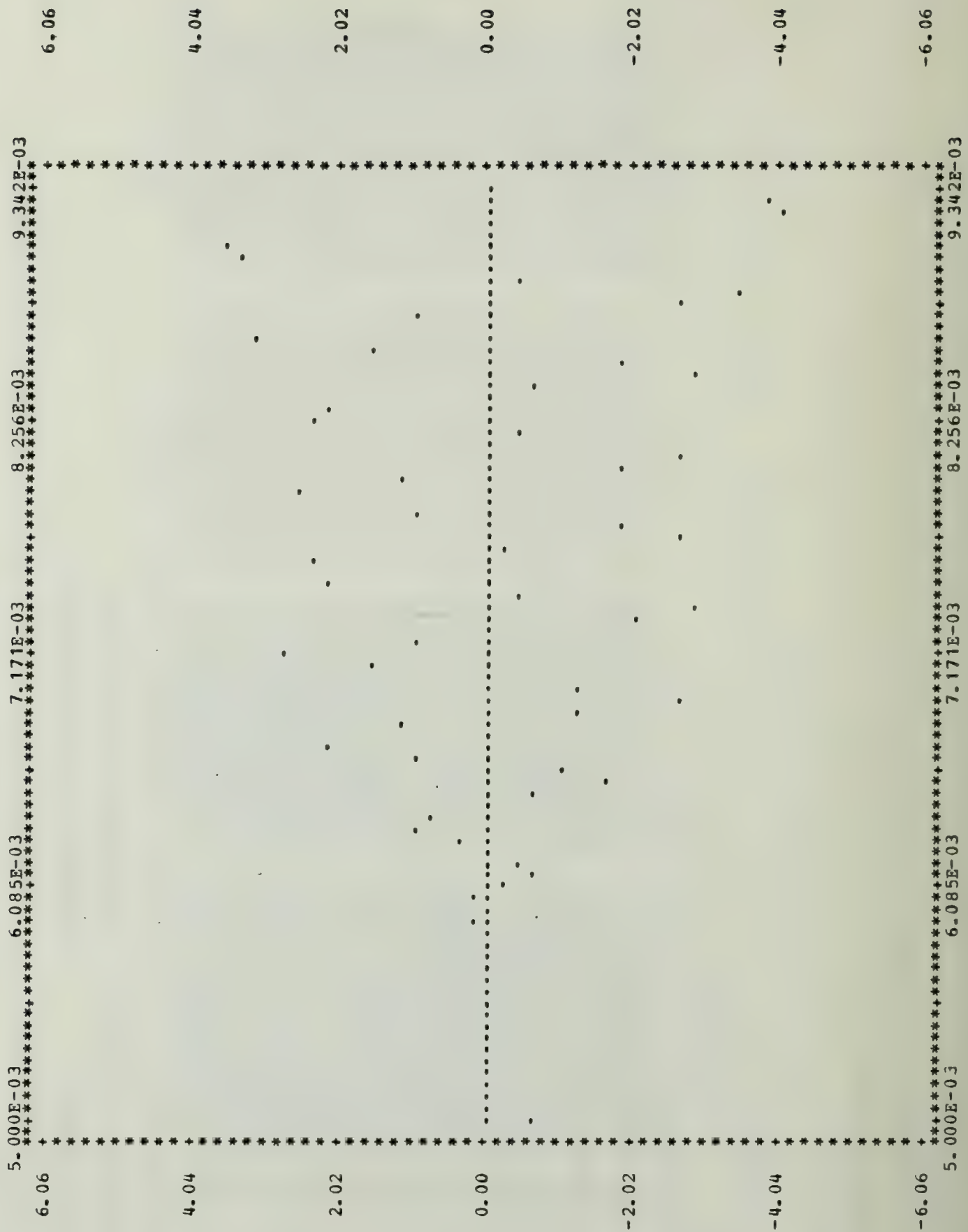


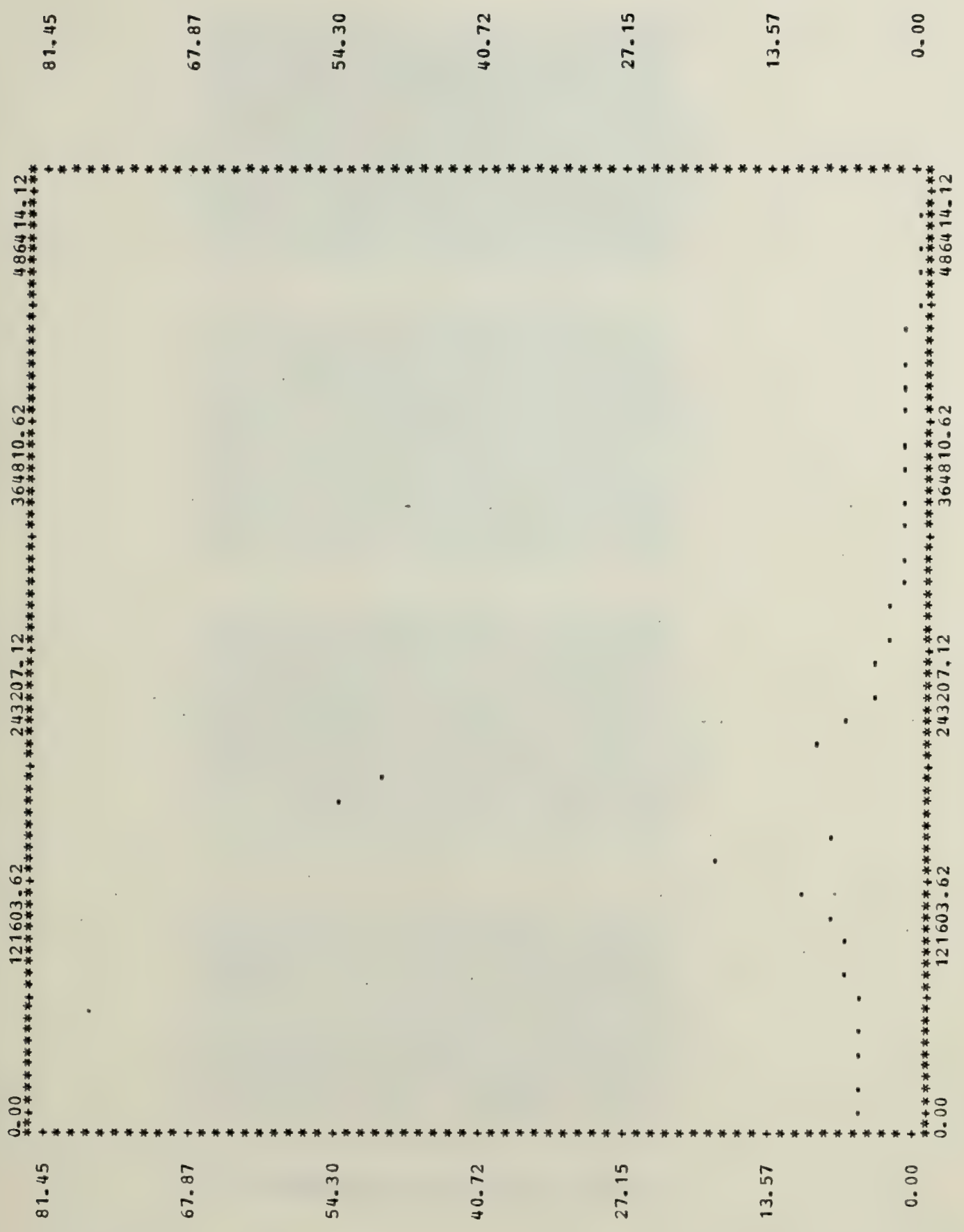
N	MEAN	VARIANCE	SKWENESS	KURTOSIS
1	1746.8113	1388348922	2282770578	263080072
2	1783.9347	1373390059	2283836957	263096047
3	1881.6866	1391153714	2283639517	263076327
4	2049.7388	1468678508	2281963351	263090676
5	2283.3526	1644047473	2281209353	263043522
6	3314.7555	1999006627	2281182331	263029241
7	3536.5555	4811195995	2280899317	263028556
8	3795.5855	1129481209	2280519680	263037022
9	3723.2140	2402922648	2280155976	263033204
10	3732.0840	2623380860	2279784342	263031804
11	3885.0840	1051549869	2279494814	263034536
12	3885.0840	1396155253	2279209481	263035353
13	3885.0840	4194625326	2278945150	263036222
14	3970.0840	2376936995	2278677160	263037177
15	3970.0840	1616678403	2278425858	263038080
16	3970.0840	1211509389	2278172087	263038933
17	3970.0840	814815822	2277933393	263039777
18	3970.0840	7084187569	2277693595	263040600
19	3970.0840	6327165569	2277453333	263041422
20	3970.0840	5373557484	2277213120	263042244
21	3970.0840	5052688758	2276972933	263043066
22	3970.0840	4821330846	2276732769	263043888
23	3970.0840	4650482001	2276492600	263044711
24	3970.0840	4528007126	2276252433	263045533
25	3970.0840	4439521405	2276012266	263046355
26	3970.0840	4372540540	2275772099	263047177
27	3970.0840	4371419567	2275531932	263048000
28	3970.0840	4382132490	2275291765	263048822
29	3970.0840	4382132490	2275051598	263049644
30	3970.0840	4382132490	2274811431	263050466
31	3970.0840	4382132490	2274571264	263051288
32	3970.0840	4382132490	2274331097	263052111
33	3970.0840	4382132490	2274090930	263052933

N	X	Y	X**2	Y**2	X**3	Y**3	SUM	X**4	Y**4
1	1746	8113	3038516	658169	45782424	3593707	206400000	54425825	158956003
2	1783	9349	317889	873081	4812537	6016671	13945984	4822344	12665942
3	1845	6865	3403225	4711625	116653125	3867369	1979648	5476630	1979648
4	2297	5885	5277009	3462425	27512500	3353681	1955782	9655930	1077361
5	2683	5056	7238489	2556336	19200000	1648716	1921907	1597192	169035
6	4514	7552	20371156	5703072	91224000	4258624	3353247	1095168	228097
7	9536	5551	90935136	3081361	841376000	170391	8097112	767903	3116
8	3232	5855	10448384	3426125	33593728	203335	1095168	228097	3116
9	3232	5855	10448384	3426125	33593728	203335	1095168	228097	3116
10	2012	3686	404824	1358436	806496	50336	806496	326622	331
11	7502	3686	56280004	1358436	421952000	100336	27396624	2067852	331
12	4016	3485	16128336	1214025	64665600	50336	25728000	10448384	331
13	2970	3808	882036	1450064	26000000	10893	26000000	10893	331
14	2390	8746	571241	7607716	13250000	10893	13250000	10893	331
15	2016	8142	4064256	6629284	16512000	5206	16512000	5206	331
16	1558	7357	2426964	5412409	3799200	3555	3799200	3555	331
17	1408	7357	198224	5412409	2771200	3555	2771200	3555	331
18	1290	8266	1664100	6823236	1699200	10893	1699200	10893	331
19	1196	4726	1435216	2237316	1699200	10893	1699200	10893	331
20	1056	5671	1114632	3215881	1114632	10893	1114632	10893	331
21	1004	8541	1008016	7294281	1008016	10893	1008016	10893	331
22	9623	9894	9259329	9788121	88812877	9259329	88812877	9259329	9788121
23	9278	1277	8607524	1633289	80000000	8607524	80000000	8607524	1633289
24	9001	1157	8101801	133849	7294281	7294281	7294281	7294281	133849
25	8786	2885	7717716	831025	6823236	5412409	6823236	5412409	5412409
26	8529	671	7274041	450241	6125000	381175	6125000	381175	381175
27	8490	5275	72072100	27820625	490079200	1490792	4900792	1490792	1490792
28	8490	5275	72072100	27820625	490079200	1490792	4900792	1490792	1490792
29	8490	5275	72072100	27820625	490079200	1490792	4900792	1490792	1490792
30	8490	5275	72072100	27820625	490079200	1490792	4900792	1490792	1490792
31	8490	5275	72072100	27820625	490079200	1490792	4900792	1490792	1490792
32	8490	5275	72072100	27820625	490079200	1490792	4900792	1490792	1490792
33	8490	5275	72072100	27820625	490079200	1490792	4900792	1490792	1490792



MODULATION TECHNIQUE = QPRS  
 QPRS CLASS 2 FILTER  
 BIFOLAR LOGIC  
 BAUD OR SYMBOL RATE = 1200 HZ  
 BITS PER BINARY CODE WORD = 2  
 BIT RATE = 0.2400000000000000E+04  
 CARRIER FREQUENCY = 2400 HZ  
 MAXIMUM CARRIER AMPLITUDE = 1 VOLT(S)  
 INITIAL PHASE ANGLE = 0 DEGREES  
 TIME BETWEEN SAMPLES = 0.6784342273548912E-04 SEC  
 NUMBER OF SAMPLES GENERATED = 64  
 SEED FOR RANDOM NUMBER GENERATOR = 1  
 NUMBER OF TIMES SIMULATION REPEATS = 100  
 SUM OF VARIANCES SUM OF VARIANCES\*\*2  
 0.2406301297164022E+04 0.1712404031156036E+07  
 SUM OF SKEWNESS SUM OF SKEWNESS\*\*2  
 -0.1014071770004887E+07 0.4097280851254902E+12  
 SUM OF KURTOSIS SUM OF KURTOSIS\*\*2  
 0.4062278349001294E+07 0.7316406994491162E+13  
 MEAN VARIANCE VARIANCE OF THE VARIANCES  
 0.7291822112618248E+02 0.4802940065853782E+05  
 MEAN SKEWNESS VARIANCE OF THE SKEWNESS  
 -0.3072944757590566E+05 0.1183019436971622E+11  
 MEAN KURTOSIS VARIANCE OF THE KURTOSIS  
 0.1230993439091301E+06 0.2130107248422762E+12







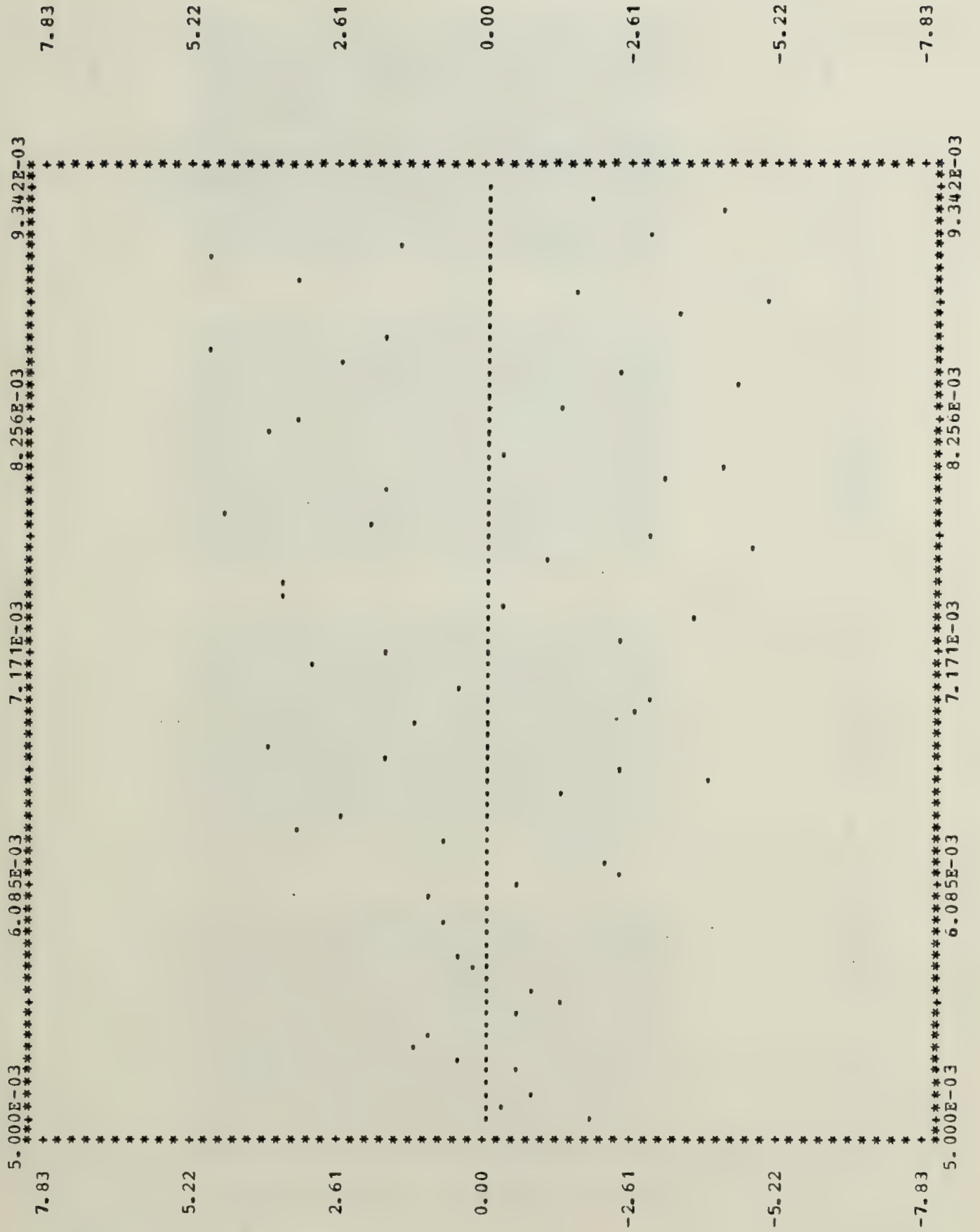
N	MEAN	VARIANCE	SKENNESS	KURTOSIS
1	288374	399220	2532	4693
2	294557	366305	2761	4671
3	112787	136664	3233	4850
4	39210	131239	4380	5378
5	43304	288919	4794	5524
6	47691	296521	8452	8996
7	49618	39535	3821	1516
8	5182	36487	7589	2649
9	2742	6600	2184	1518
10	2640	6415	2781	1681
11	2464	8781	2781	1123
12	2515	2881	3981	2059
13	2765	3373	7651	3921
14	2812	4237	8873	4166
15	3039	6408	1087	6900
16	3396	9520	1408	1082
17	3750	1585	2241	1592
18	4572	2300	3091	2593
19	5945	3769	5063	3992
20	7264	5892	7444	5318
21	8316	7701	9189	7799
22	9967	1187	1379	1096
23	1303	1665	1993	1394
24	1961	2606	3096	2269
25	2806	4264	5211	3380
26	3904	7255	8521	4919
27	5794	1263	1378	7723
28	8066	2240	2263	1094
29	1159	3922	3791	1591
30	1623	6667	6240	2263
31	2324	1057	9333	3383
32	3164	1663	1385	4955
33	4814	2563	2146	7469

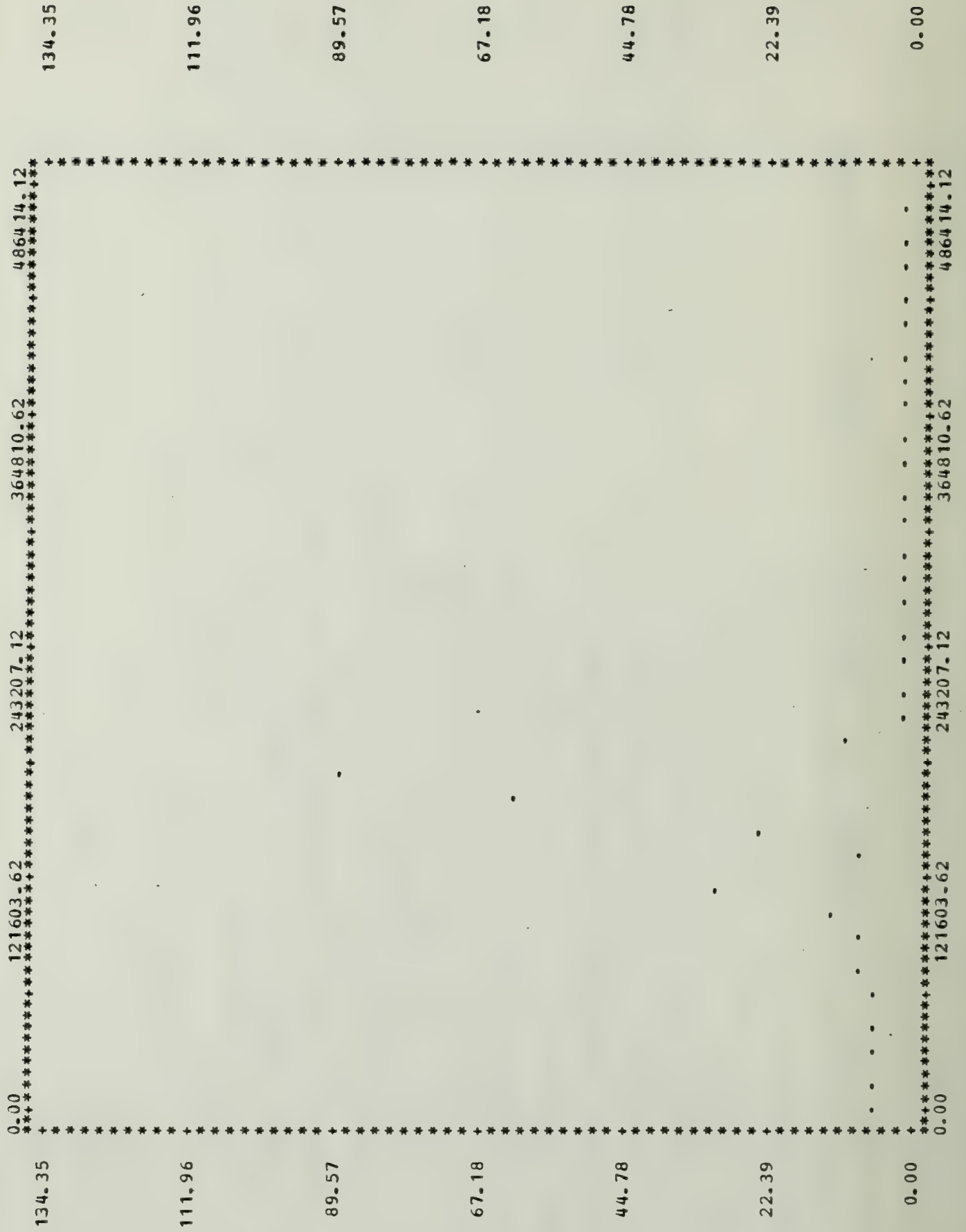
N	X	SUM	X**2	X**3	X**4
1	2883	7442	0527	1784	2051
2	2945	7556	0677	1989	2285
3	3112	7856	0974	2508	2985
4	3392	1056	1366	3598	3985
5	3814	1330	1569	3791	5222
6	4448	1630	1923	4649	5651
7	4646	1769	2145	5173	7938
8	4996	1994	2505	6116	9389
9	5211	2066	2660	6469	9955
10	5544	2254	3077	7373	12975
11	5879	2464	3370	8099	14207
12	6100	2515	3633	8514	14833
13	6281	2550	3773	8769	15233
14	6628	2765	4037	9378	16163
15	5078	2300	2544	6299	3193
16	4154	1750	1754	4399	1821
17	3774	1509	1403	3774	1409
18	3103	1095	1033	2884	919
19	2520	764	603	2049	517
20	2153	599	463	1713	366
21	2019	513	413	1536	316
22	1908	480	362	1418	289
23	1739	447	315	1261	253
24	1623	406	274	1111	224
25	1551	392	245	1011	205
26	1531	385	235	989	200
27	1523	381	232	982	199
28	1523	381	232	982	199
29	1523	381	232	982	199
30	1523	381	232	982	199
31	1523	381	232	982	199
32	1523	381	232	982	199
33	1523	381	232	982	199
34	1523	381	232	982	199
35	1523	381	232	982	199
36	1523	381	232	982	199
37	1523	381	232	982	199
38	1523	381	232	982	199
39	1523	381	232	982	199
40	1523	381	232	982	199
41	1523	381	232	982	199
42	1523	381	232	982	199
43	1523	381	232	982	199
44	1523	381	232	982	199
45	1523	381	232	982	199
46	1523	381	232	982	199
47	1523	381	232	982	199
48	1523	381	232	982	199
49	1523	381	232	982	199
50	1523	381	232	982	199



MODULATION TECHNIQUE = QPRS  
 QPRS CLASS 3 FILTER  
 BIFOLAR LOGIC  
 BAUD OR SYMBOL RATE = 1200 HZ  
 BITS PER BINARY CODE WORD = 2  
 BIT RATE = 0.2400000000000000E+04  
 CARRIER FREQUENCY = 2400 HZ  
 MAXIMUM CARRIER AMPLITUDE = 1 VOLT(S)  
 INITIAL PHASE ANGLE = 0 DEGREES  
 TIME BETWEEN SAMPLES = 0.6784342273548912E-04 SEC  
 NUMBER OF SAMPLES GENERATED = 64  
 SEED FOR RANDOM NUMBER GENERATOR = 1  
 NUMBER OF TIMES SIMULATION REPEATS = 100  
 SUM OF VARIANCES SUM OF VARIANCES\*\*2  
 0.2212313780321216E+04 0.9349388767869366E+06  
 SUM OF SKEWNESS SUM OF SKEWNESS\*\*2  
 -0.7876831322230185E+06 0.1444178694786092E+12  
 SUM OF KURTOSIS SUM OF KURTOSIS\*\*2  
 0.2375351460755018E+07 0.1504008831530188E+13  
 MEAN VARIANCE VARIANCE OF THE VARIANCES  
 0.6703981152488534E+02 0.2458205556000923E+05  
 MEAN SKEWNESS VARIANCE OF THE SKEWNESS  
 -0.2386918582493996E+05 0.3925516075762725E+10  
 MEAN KURTOSIS VARIANCE OF THE KURTOSIS  
 0.7198034729560660E+05 0.4165719401362236E+11





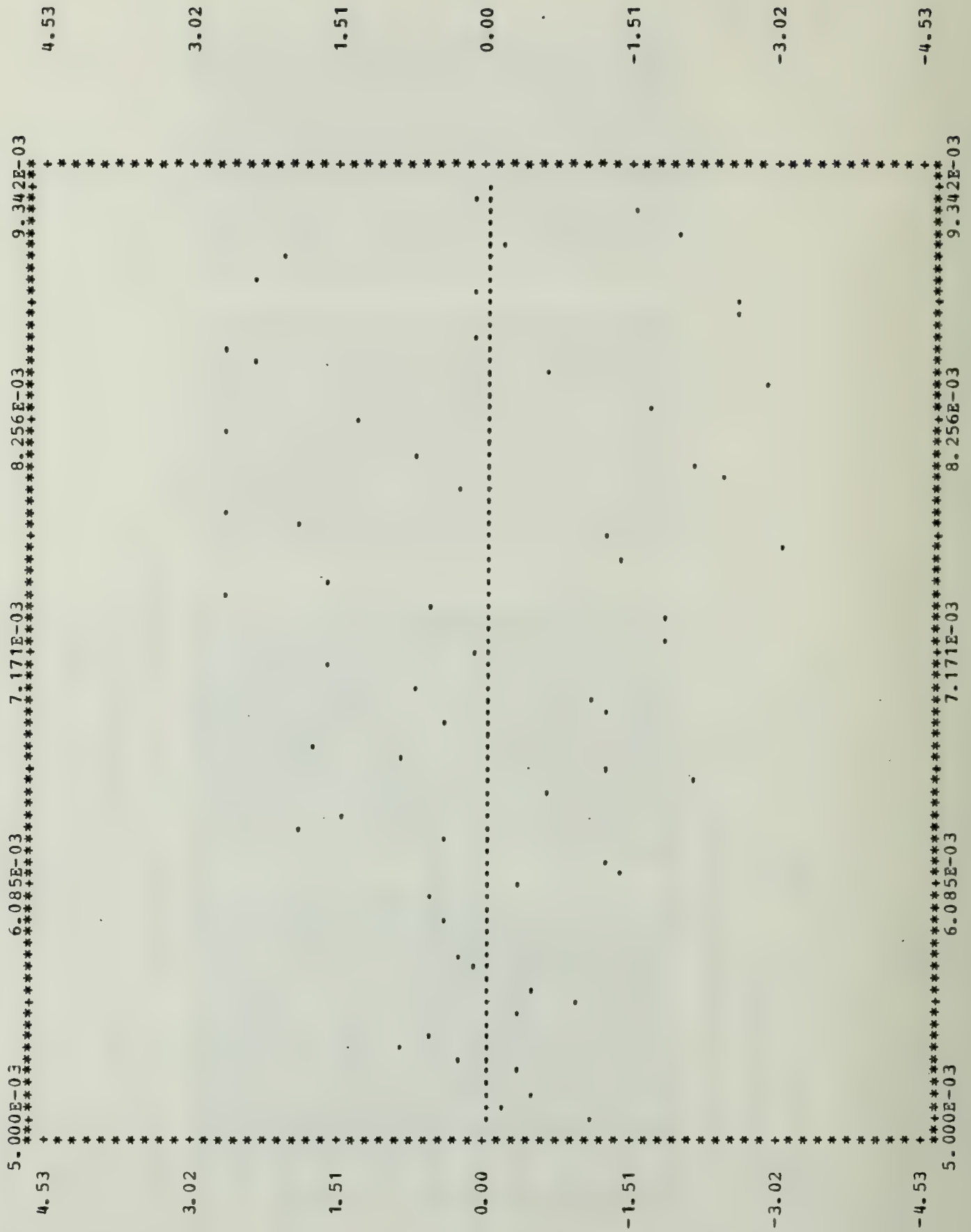


N	MEAN	VARIANCE	SKENWESS	KURTOSIS
1	2859	19334	8822	5158
2	2941	19333	8255	4712
3	3472	29390	29891	2631
4	3969	30638	0017	6225
5	4735	42348	6614	2734
6	6018	53778	0703	3503
7	8654	25866	5708	1304
8	2281	13774	9588	2290
9	4754	54611	3236	3322
10	4244	75014	7741	5033
11	4734	96644	6308	4063
12	4636	50244	7034	9374
13	8010	57171	4065	2299
14	5748	36086	2508	1627
15	5873	99088	8117	3853
16	3356	70424	6710	4812
17	2985	70424	6710	4812
18	2700	28932	9982	3620
19	2474	43615	8878	3331
20	2214	36034	5702	1659
21	2017	35220	8785	2796
22	1909	80210	9242	8610
23	1739	29733	9562	2044
24	1673	01488	7948	5713
25	1618	30967	9485	7132
26	1575	12726	9274	7432
27	1544	22882	7889	5295
28	1530	49827	8895	2538
29	1530	49827	8895	2538
30	1530	49827	8895	2538
31	1530	49827	8895	2538
32	1530	49827	8895	2538
33	1530	49827	8895	2538

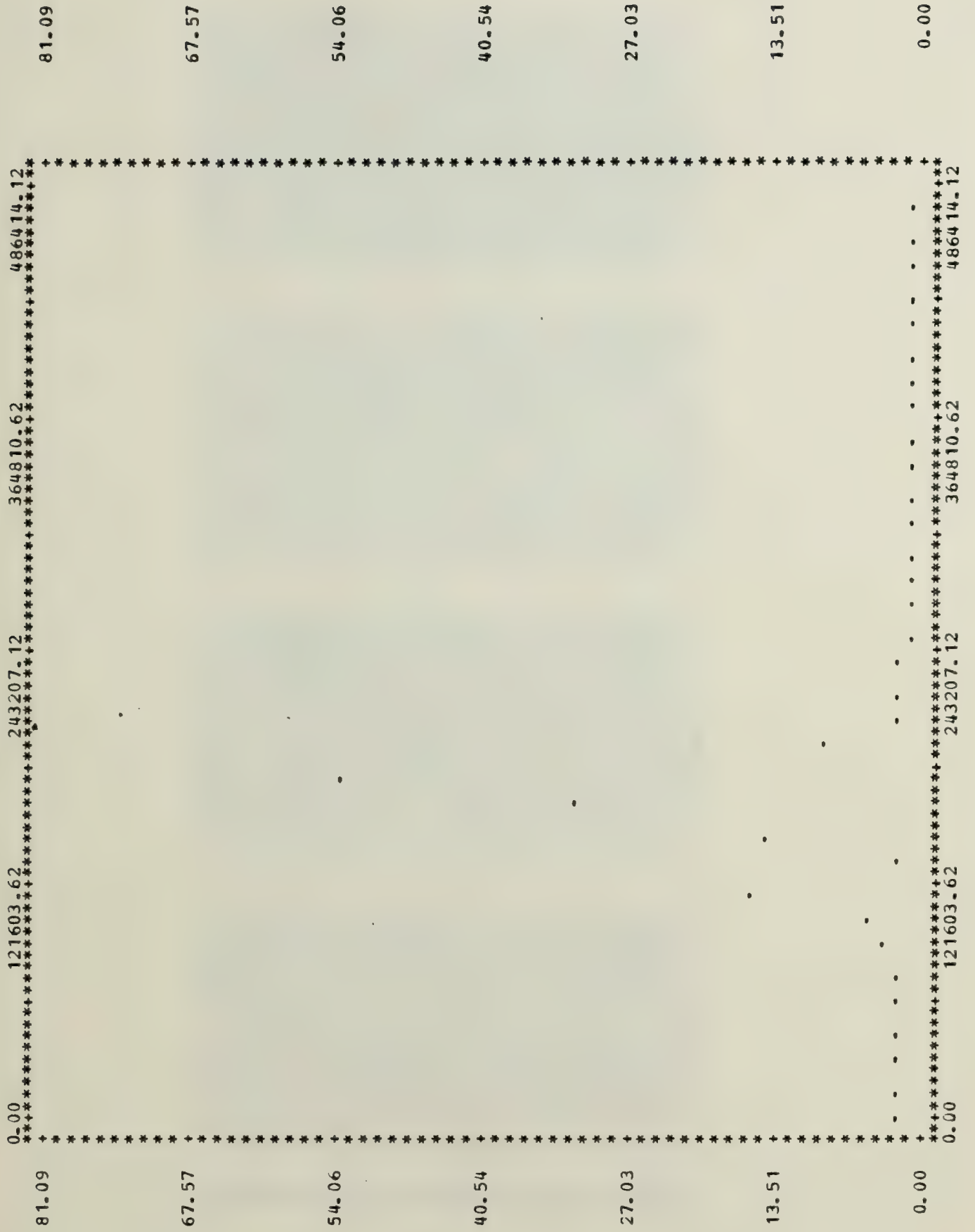




MODULATION TECHNIQUE = QPRS  
 QPS CLASS 4 FILTER  
 BIFOLAR LOGIC  
 BAUD OR SYMBOL RATE = 1200 HZ  
 BITS PER BINARY CODE WORD = 2  
 BIT RATE = 0.2400000000000000E+04  
 CARRIER FREQUENCY = 2400 HZ  
 MAXIMUM CARRIER AMPLITUDE = 1 VOLT (S)  
 INITIAL PHASE ANGLE = 0 DEGREES  
 TIME BETWEEN SAMPLES = 0.6784342273548912E-04 SEC  
 NUMBER OF SAMPLES GENERATED = 64  
 SEED FOR RANDOM NUMBER GENERATOR = 1  
 NUMBER OF TIMES SIMULATION REPEATS = 100  
 SUM OF VARIANCES. SUM OF VARIANCES\*\*2  
 0.7516311188359808E+03 0.1147433777377834E+06  
 SUM OF SKEWNESS SUM OF SKEWNESS\*\*2  
 -0.1502958760239533E+06 0.5926297375617011E+10  
 SUM OF KURTOSIS SUM OF KURTOSIS\*\*2  
 0.3178921864476514E+06 0.3285296126177793E+11  
 MEAN VARIANCE VARIANCE OF THE VARIANCES  
 0.2277670057078729E+02 0.3050740650136574E+04  
 MEAN SKEWNESS VARIANCE OF THE SKEWNESS  
 -0.4554420485574342E+04 0.1638058362173805E+09  
 MEAN KURTOSIS VARIANCE OF THE KURTOSIS  
 0.9633096559019739E+04 0.9309585979490561E+09





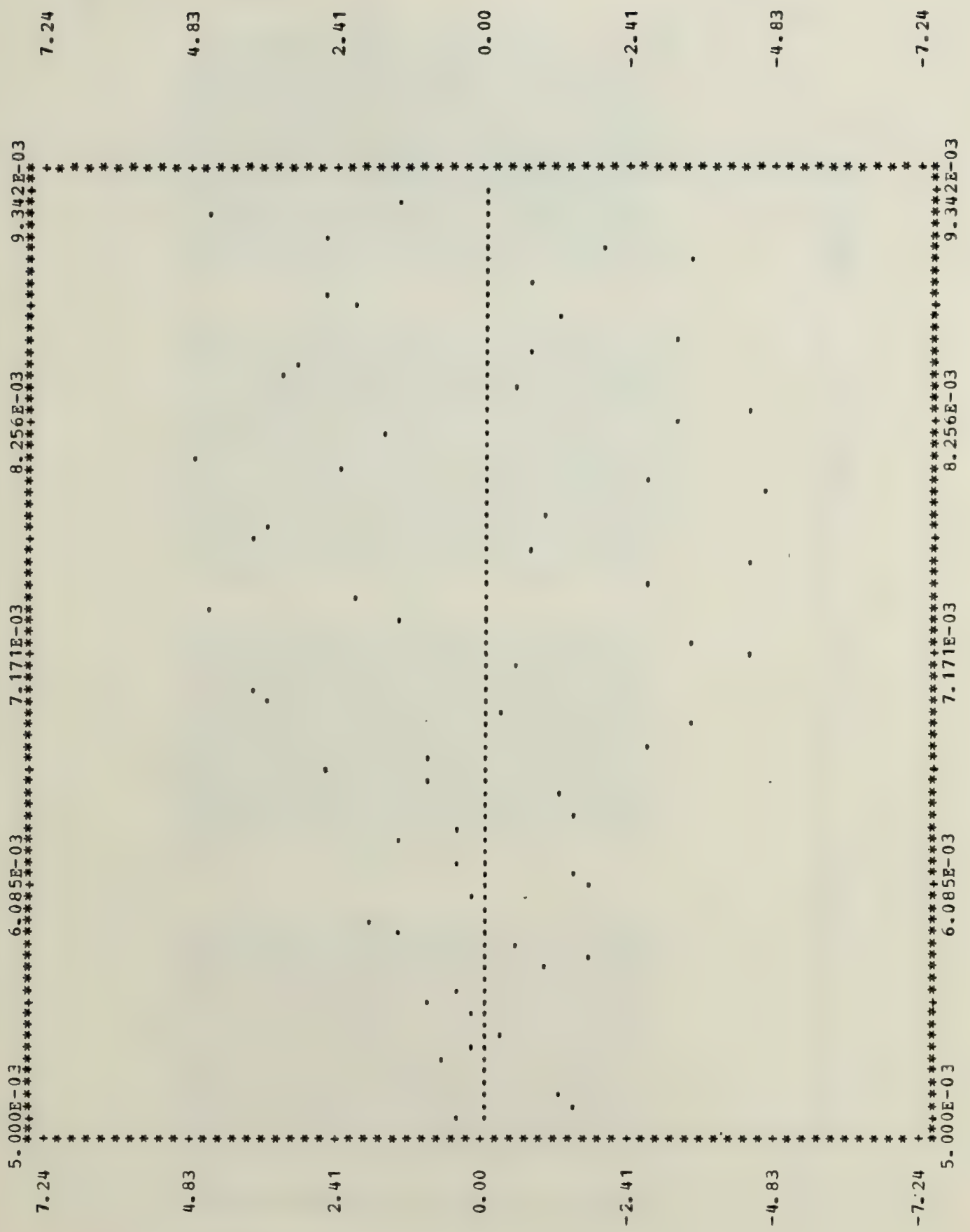


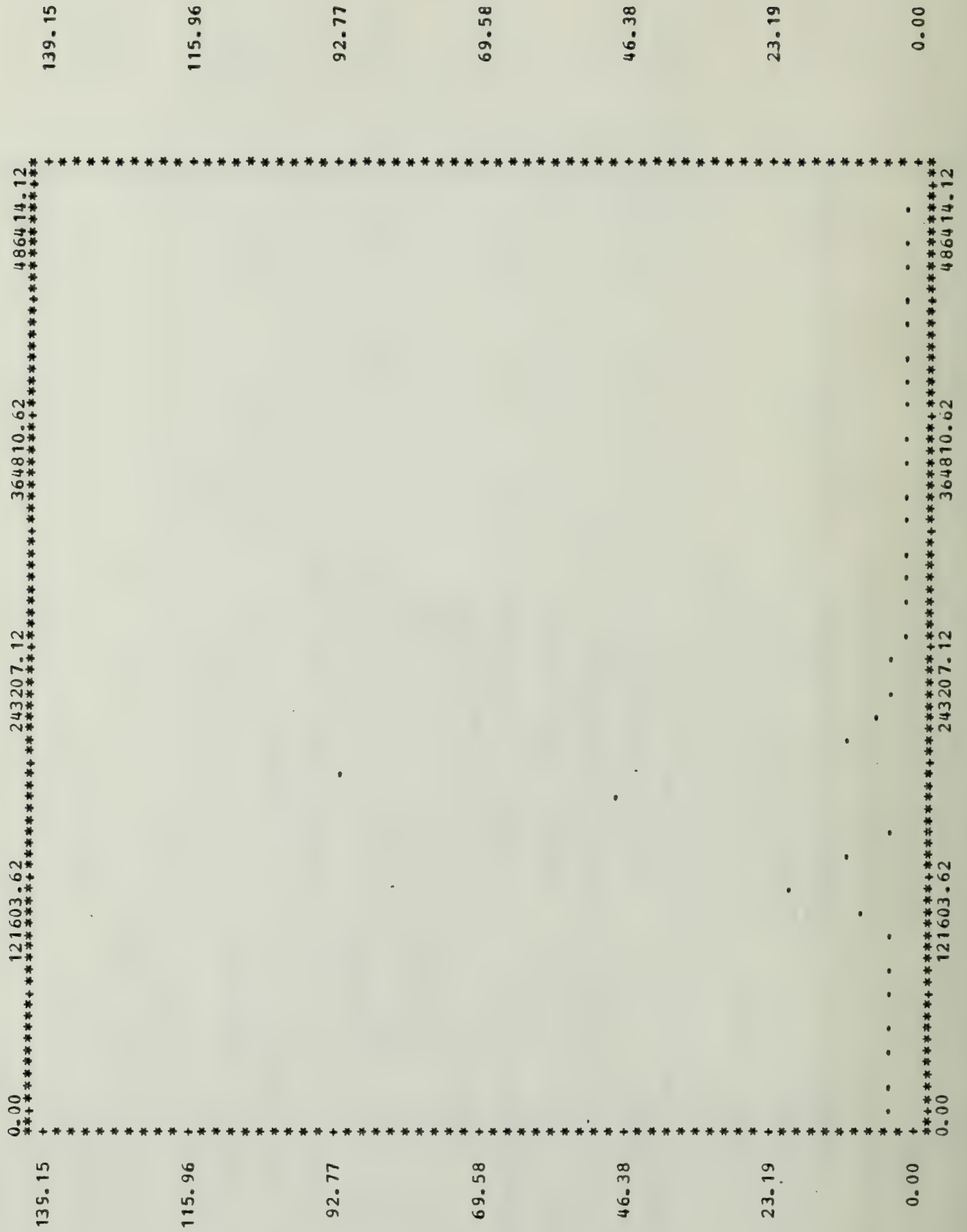
N	MEAN	VARIANCE	SKENNESS	KURTOSIS
1	1639783	1709248	1134099	1062312
2	1694119	1659799	1633184	1093616
3	1822084	1697975	33360	1051811
4	1944483	1696003	527175	1149726
5	2329575	1889410	625566	1396880
6	2799618	2351081	977226	2003885
7	3593606	3179843	226557	3923065
8	4687324	4331933	630633	5965534
9	5319333	515225	711517	7700144
10	5962582	5822507	813819	8875606
11	6319558	6258196	933063	1015887
12	6818585	6654210	953758	1167317
13	7319558	6855421	973375	1262167
14	7818594	6654210	953758	1430962
15	8435948	5007521	286576	2014017
16	9048483	2436613	192695	3959197
17	9710923	2436613	192695	5591976
18	1033535	1543256	111531	7304581
19	1099223	1116116	899345	9081626
20	1165223	873063	308584	1058898
21	1232565	7086980	308584	1275128
22	1300918	522377	308584	1557788
23	1369257	4737584	222515	1721968
24	1437600	4420632	175225	1933599
25	1505943	4116512	155262	2146330
26	1574286	4070568	137297	2359971
27	1642629	4162209	129299	2573749
28	1710972	4305494	115443	2790740
29	1779315	4338672	102753	3007407
30	1847658	4338672	102753	3224095
31	1915999	4338672	102753	3440783
32	1984342	4338672	102753	3657471
33	2052685	4338672	102753	3874159

N	X	SUM	X**2	X**3	X**4
1	16397831	0841063E+03	43811044	8806267950E+03	22449494587657E+04
2	16941190	10472507E+03	45132082	625687009E+03	2242510452042E+04
3	18220844	17905823E+03	49415472	11585657E+03	715162869890E+04
4	232295753	01573337E+03	72974370	41226407E+03	9029491598422E+04
5	27996184	59186339E+03	10165433	390755379E+04	1304690135072E+05
6	35936066	90701501E+03	16396339	386513001E+04	22848244523839E+05
7	52502686	78825291E+03	35168339	529825987E+04	544432206523839E+05
8	14687324	45728772E+04	27213197	319799950061E+05	2395389115966947E+06
9	19715319	33822615E+04	46312323	3334921331E+05	412984839791464E+06
10	24933195	58019601E+04	74073685	98838439E+05	884735841076753E+06
11	29556818	86654210E+04	10943366	202513548E+05	222567420421119E+06
12	49709048	4803092E+03	15733227	769165E+05	423314469472229E+07
13	35327571	03156359E+03	29659224	590723244E+04	1474973714749334E+07
14	28124367	95359599E+03	14892620	66775789E+04	3695846281176221E+05
15	33677526	81199091E+03	9437629	6401034992E+03	1488190681227194E+05
16	20677526	81199091E+03	51399337	17236453E+03	76080533377373551E+04
17	18526645	6563048547E+03	41338978	51788587E+03	4515033391001683E+04
18	16910492	57518267E+03	3450675	643836312E+03	297421844064030E+04
19	15644851	52670945E+03	2965192	3175910359E+03	2118738265791955E+04
20	14627245	37650117E+03	2608601	159823359E+03	1604859418202128E+04
21	13793611	98878539E+03	2340279	896914001E+03	127734943962953624E+04
22	13102129	71454939E+03	2134948	066914001E+03	9081718957376042E+03
23	12524666	06083759E+03	1976207	372245338E+03	8013993822923571E+03
24	12041969	57391062E+03	1853076	586681869E+03	7244971790085427E+03
25	11640703	64847772E+03	1758038	704075860E+03	6686943101747503E+03
26	11311639	35402371E+03	1685838	715559886E+03	6283944137293207E+03
27	11049050	18984923E+03	1622862	857850134E+03	5999720029213764E+03
28	10850949	57871504E+03	1596611	300350684E+03	5810861131486764E+03
29	10720177	99482293E+03	1575466	096633816E+03	5702822834018495E+03
30	10668961	45745031E+03	1568515	828666411E+03	5667653419404097E+03



MODULATION TECHNIQUE = QPRS  
 QPRS CLASS 5 FILTER  
 BIPOLAR LOGIC  
 BAUD OR SYMBOL RATE = 1200 HZ  
 BITS PER BINARY CODE WORD = 2  
 BIT RATE = 0.2400000000000000E+04  
 CARRIER FREQUENCY = 2400 HZ  
 MAXIMUM CARRIER AMPLITUDE = 1 VOLT(S)  
 INITIAL PHASE ANGLE = 0 DEGREES  
 TIME BETWEEN SAMPLES = 0.6784342273548912E-04 SEC  
 NUMBER OF SAMPLES GENERATED = 64  
 SEED FOR RANDOM NUMBER GENERATOR = 1  
 NUMBER OF TIMES SIMULATION REPEATS = 100  
 SUM OF VARIANCES SUM OF VARIANCES\*\*2  
 0.2422177865073176E+04 0.1348538369218457E+07  
 SUM OF SKEWNESS SUM OF SKEWNESS\*\*2  
 -0.8202160755049249E+06 0.2049138949944578E+12  
 SUM OF KURTOSIS SUM OF KURTOSIS\*\*2  
 0.3582703927263425E+07 0.4782889442290025E+13  
 MEAN VARIANCE VARIANCE OF THE VARIANCES  
 0.7339932924464169E+02 0.3658600433158962E+05  
 MEAN SKEWNESS VARIANCE OF THE SKEWNESS  
 -0.2485503259105833E+05 0.5766481178314780E+10  
 MEAN KURTOSIS VARIANCE OF THE KURTOSIS  
 0.1085667856746492E+06 0.1373102122775966E+12







N	MEAN	VARIANCE	SKEWNESS	KURTOSIS
1	317500	979664	377247	639011
2	324061	199444	572217	779029
3	418492	879921	490172	581606
4	372005	986691	129192	486419
5	418807	361150	169979	383790
6	492121	435658	169979	383790
7	164678	3320300	255922	183467
8	673493	874244	198888	503593
9	200145	889591	684580	666961
10	540923	215610	118866	149970
11	307109	172333	609429	431499
12	412506	118058	954188	588371
13	540519	849648	560075	335909
14	156244	696486	964335	374262
15	802742	633915	717240	626262
16	459631	104760	220092	185368
17	386341	179601	288359	905154
18	336315	582533	722475	605035
19	300326	108183	224502	573149
20	273190	428099	131233	157314
21	252123	441845	829145	686683
22	235420	528298	615082	567365
23	219851	85742	159828	220960
24	110603	355400	255638	673201
25	202107	551111	598655	329450
26	194757	703391	261902	669188
27	188757	144657	607621	658534
28	183939	664964	789929	110818
29	180200	719831	290971	208507
30	177490	732841	465967	395902
31	175811	235197	472890	895180
32	175226	333447	704326	551887
33	175226	333447	704326	551887

N	X	SUM	X**2	SUM	X**3	SUM	X**4
1	1	317	1	154	1	565	1
2	2	341	4	155	8	899	16
3	3	418	9	160	27	228	81
4	4	372	16	168	64	506	256
5	5	418	25	192	125	623	625
6	6	492	36	236	216	1224	1296
7	7	616	49	328	343	2111	2401
8	8	734	64	427	512	4820	32768
9	9	873	81	533	729	7867	6729
10	10	1009	100	657	1000	1900	16000
11	11	1205	121	799	1331	2729	14641
12	12	1377	144	958	1728	3583	20736
13	13	1566	169	1138	2197	4613	28242
14	14	1827	196	1343	2744	6199	38416
15	15	2104	225	1578	3375	8550	50625
16	16	2409	256	1847	4096	11889	65536
17	17	2744	289	2155	4913	16227	83521
18	18	3109	324	2507	5832	21671	104976
19	19	3506	361	2911	6859	28423	130321
20	20	3937	400	3378	8000	36793	160000
21	21	4404	441	3911	9261	47027	194481
22	22	4909	484	4514	10648	60023	234368
23	23	5454	529	5191	12277	75783	282429
24	24	6041	576	5938	14160	94527	346624
25	25	6672	625	6761	16300	116877	426976
26	26	7349	676	7668	17800	144823	527296
27	27	8074	729	8657	19683	178987	651521
28	28	8849	784	9738	21952	219903	806496
29	29	9676	841	10911	24609	278423	990529
30	30	10557	900	12198	27660	355207	1216700
31	31	11494	961	13611	31113	451927	1481556
32	32	12489	1024	15162	35168	570283	1770081
33	33	13544	1089	16863	39843	713187	2125441







```

C      X=0. DO
C      Y=0. DO
C      Z=0. DO
C      DO 30 I=1,15
C          X=X+SUM X(I)
C          Y=Y+SUM XSQ(I)
C          Z=Z+SUM X(I)**2
C      CONTINUE
C      XA=(Z/N)-(X**2/(R*N))
C      XE=Y-(Z/N)
C      TOTAL=Y-(X**2/(R*N))
C      DF1=R-1. DO
C      DF2=R*(N-1. DO)
C      FSTAT=(XA/DF1)/(XE/DF2)
C      *** OUTPUT THE RESULTS-THE FORMATS REPRESENT VARIABLE DEFINITIONS
C      WRITE(6,31) J
C      FORMAT(1,' F-STATISTICS FOR THE',I4,' COMPONENT')
C      WRITE(6,40) X
C      FORMAT(10,' SUM OF SUMS =',E23.16)
C      WRITE(6,41) Y
C      FORMAT(10,' SUM OF SUMS**2 =',E23.16)
C      WRITE(6,42) Z
C      FORMAT(10,' SUM OF SQUARE OF SUMS =',E23.16)
C      WRITE(6,50) XA
C      FORMAT(10,' AMONG GROUP SUM OF SQUARES =',E23.16)
C      WRITE(6,51) XE
C      FORMAT(10,' WITHIN GROUP SUM OF SQUARES =',E23.16)
C      WRITE(6,60) TOTAL
C      FORMAT(10,' TOTAL SUM OF SQUARES =',E23.16)
C      WRITE(6,70) DF1
C      FORMAT(10,' DEGREES OF FREEDOM IN NUMERATOR =',E23.16)
C      WRITE(6,71) DF2
C      FORMAT(10,' DEGREES OF FREEDOM IN DENOMINATOR =',E23.16)
FTE 00450
FTE 00460
FTE 00470
FTE 00480
FTE 00490
FTE 00500
FTE 00510
FTE 00520
FTE 00530
FTE 00540
FTE 00550
FTE 00560
FTE 00570
FTE 00580
FTE 00590
FTE 00600
FTE 00610
FTE 00620
FTE 00630
FTE 00640
FTE 00650
FTE 00660
FTE 00670
FTE 00680
FTE 00690
FTE 00700
FTE 00710
FTE 00720
FTE 00730
FTE 00740
FTE 00750
FTE 00760
FTE 00770
FTE 00780
FTE 00790
FTE 00800
FTE 00810
FTE 00820
FTE 00830
FTE 00840
FTE 00850
FTE 00860
FTE 00870
FTE 00880
FTE 00890
FTE 00900
FTE 00910
FTE 00920
FTE 00930
FTE 00940

```

```
80 WRITE(6,80) FSTAT  
C  FORMAT('0.', ' F-STATISTIC =', E23.16)  
C 100 CONTINUE  
C 81 WRITE(6,81)  
C  FORMAT('1.')
```

```
FTE00950  
FTE00960  
FTE00970  
FTE00980  
FTE00990  
FTE01000  
FTE01010  
FTE01020  
FTE01030
```

APPENDIX E  
RESULTS OF F-TEST

F-STATISTICS FOR THE 1 COMPONENT

SUM OF SUMS = 0.2091487882294994D 02  
SUM OF SUMS\*\*2 = 0.1810655469453404D 02  
SUM OF SQUARE OF SUMS = 0.4128097510679048D 02  
AMCNG GROUP SUM OF SQUARES = 0.1211883136154515D 00  
WITHIN GROUP SUM OF SQUARES = 0.1769374494346614D 02  
TOTAL SUM OF SQUARES = 0.1781493325708159D 02  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.7265063214387600D 00

F-STATISTICS<sup>A</sup> FOR THE 2 COMPONENT

SUM OF SUMS = 0.2305792717805020D 02  
SUM OF SUMS\*\*2 = 0.1894701234015291D 02  
SUM OF SQUARE OF SUMS = 0.4663825126655468D 02  
AMCNG GROUP SUM OF SQUARES = 0.1119371755000360D 00  
WITHIN GROUP SUM OF SQUARES = 0.1848062982748736D 02  
TOTAL SUM OF SQUARES = 0.1859256700298739D 02  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.6424746465014737D 00

F-STATISTICS FOR THE 3 COMPONENT

SUM OF SUMS = 0.2693734535827635D 02  
SUM OF SUMS\*\*2 = 0.2124674921974186D 02  
SUM OF SQUARE OF SUMS = 0.5921566381358867D 02  
AMCNG GROUP SUM OF SQUARES = 0.1084095881685183D 00  
WITHIN GROUP SUM OF SQUARES = 0.2065459258160598D 02  
TOTAL SUM OF SQUARES = 0.2076300216977449D 02  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.5567362242775778D 00



F-STATISTICS FOR THE 4 COMPONENT

SUM OF SUMS = 0.3175688631627065D 02  
SUM OF SUMS\*\*2 = 0.2621104542300214D 02  
SUM OF SQUARE OF SUMS = 0.8070513093726266D 02  
AMONG GROUP SUM OF SQUARES = 0.1347180903696015D 00  
WITHIN GROUP SUM OF SQUARES = 0.2540399411362951D 02  
TOTAL SUM OF SQUARES = 0.2553871220399911D 02  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04

F-STATISTIC = 0.5624997485041869D 00

F-STATISTICS FOR THE 5 COMPONENT

SUM OF SUMS = 0.3669104349531116D 02  
SUM OF SUMS\*\*2 = 0.3380633740317695D 02  
SUM OF SQUARE OF SUMS = 0.1069682473356158D 03  
AMONG GROUP SUM OF SQUARES = 0.1721940248396148D 00  
WITHIN GROUP SUM OF SQUARES = 0.3273665492982079D 02  
TOTAL SUM OF SQUARES = 0.3290884895466041D 02  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04

F-STATISTIC = 0.5579331866788867D 00

F-STATISTICS FOR THE 6 COMPONENT

SUM OF SUMS = 0.4160097878948431D 02  
SUM OF SUMS\*\*2 = 0.4313236642020094D 02  
SUM OF SQUARE OF SUMS = 0.1408215358007360D 03  
AMONG GROUP SUM OF SQUARES = 0.2544544005119440D 00  
WITHIN GROUP SUM OF SQUARES = 0.4172415106219358D 02  
TOTAL SUM OF SQUARES = 0.4197860546270552D 02  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04

F-STATISTIC = 0.6468757561623442D 00

F-STATISTICS FOR THE 7 COMPONENT

SUM OF SUMS = 0.4810387092348111D 02  
SUM OF SUMS\*\*2 = 0.5206409381611744D 02  
SUM OF SQUARE OF SUMS = 0.1964653659903027D 03  
AMONG GROUP SUM OF SQUARES = 0.4219987280210724D 00  
WITHIN GROUP SUM OF SQUARES = 0.5009944015621441D 02  
TOTAL SUM OF SQUARES = 0.5052143888423548D 02  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.8934632362547178D 00

F-STATISTICS FOR THE 8 COMPONENT

SUM OF SUMS = 0.5623255941290002D 02  
SUM OF SUMS\*\*2 = 0.8590285622098203D 02  
SUM OF SQUARE OF SUMS = 0.3070018365343552D 03  
AMONG GROUP SUM OF SQUARES = 0.9619512065933320D 00  
WITHIN GROUP SUM OF SQUARES = 0.8283283785563848D 02  
TOTAL SUM OF SQUARES = 0.8379478906223181D 02  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.1231824736914027D 01

F-STATISTICS FOR THE 9 COMPONENT

SUM OF SUMS = 0.9931605295569187D 02  
SUM OF SUMS\*\*2 = 0.4043531056175881D 03  
SUM OF SQUARE OF SUMS = 0.1416642417283150D 04  
AMONG GROUP SUM OF SQUARES = 0.7590638589699638D 01  
WITHIN GROUP SUM OF SQUARES = 0.3901866814447566D 03  
TOTAL SUM OF SQUARES = 0.3977773200344562D 03  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.2063499133280512D 01

F-STATISTICS FOR THE 10 COMPONENT

SUM OF SUMS = 0.2029942628781924D 03  
SUM OF SUMS\*\*2 = 0.1757564757819924D 04  
SUM OF SQUARE OF SUMS = 0.7655692250451960D 04  
AMCNG GROUP SUM OF SQUARES = 0.4908580866354581D 02  
WITHIN GROUP SUM OF SQUARES = 0.1681007835315404D 04  
TOTAL SUM OF SQUARES = 0.1730093643978950D 04  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.3097309684192641D 01

F-STATISTICS FOR THE 11 COMPONENT

SUM OF SUMS = 0.2004773302247246D 03  
SUM OF SUMS\*\*2 = 0.1811875077561617D 04  
SUM OF SQUARE OF SUMS = 0.7980439863108266D 04  
AMCNG GROUP SUM OF SQUARES = 0.5301029200839379D 02  
WITHIN GROUP SUM OF SQUARES = 0.1732070678930534D 04  
TOTAL SUM OF SQUARES = 0.1785080970938928D 04  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.3246332537532912D 01

F-STATISTICS FOR THE 12 COMPONENT

SUM OF SUMS = 0.2210834652773109D 03  
SUM OF SUMS\*\*2 = 0.2165859765218944D 04  
SUM OF SQUARE OF SUMS = 0.9414279932820438D 04  
AMCNG GROUP SUM OF SQUARES = 0.6155753358218842D 02  
WITHIN GROUP SUM OF SQUARES = 0.2071716965890740D 04  
TOTAL SUM OF SQUARES = 0.2133274499472928D 04  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.3151731454585565D 01



F-STATISTICS FOR THE 13 COMPONENT

SUM OF SUMS = 0.1979583434471791D 03

SUM OF SUMS\*\*2 = 0.1904701075347160D 04

SUM OF SQUARE OF SUMS = 0.7185446672840862D 04

AMCNG GROUP SUM OF SQUARES = 0.4572946290150773D 02

WITHIN GROUP SUM OF SQUARES = 0.1832846608618751D 04

TOTAL SUM OF SQUARES = 0.1878576071520259D 04

DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02

DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04

F-STATISTIC = 0.2646478671459864D 01

F-STATISTICS FOR THE 14 COMPONENT

SUM OF SUMS = 0.9227736272614619D 02

SUM OF SUMS\*\*2 = 0.2600827196886254D 03

SUM OF SQUARE OF SUMS = 0.9540726819081072D 03

AMCNG GROUP SUM OF SQUARES = 0.3863985704619236D 01

WITHIN GROUP SUM OF SQUARES = 0.2505419928695444D 03

TOTAL SUM OF SQUARES = 0.2544059785741636D 03

DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02

DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04

F-STATISTIC = 0.1635887377498235D 01

F-STATISTICS FOR THE 15 COMPONENT

SUM OF SUMS = 0.6880926336210145D 02

SUM OF SUMS\*\*2 = 0.1018300362534556D 03

SUM OF SQUARE OF SUMS = 0.3695062863415098D 03

AMCNG GROUP SUM OF SQUARES = 0.5385863804584070D 00

WITHIN GROUP SUM OF SQUARES = 0.9813497339004056D 02

TOTAL SUM OF SQUARES = 0.9867355977049896D 02

DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02

DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04

F-STATISTIC = 0.5821433971075596D 00

F-STATISTICS FOR THE 16 COMPONENT

SUM OF SUMS = 0.6961387326450431D 02  
SUM OF SUMS\*\*2 = 0.1137157033969463D 03  
SUM OF SQUARE OF SUMS = 0.3597053597878584D 03  
AMCNG GROUP SUM OF SQUARES = 0.3663260306209388D 00  
WITHIN GROUP SUM OF SQUARES = 0.1101186497990677D 03  
TOTAL SUM OF SQUARES = 0.1104849758296886D 03  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.3528623485827816D 00

F-STATISTICS FOR THE 17 COMPONENT

SUM OF SUMS = 0.7052854165691613D 02  
SUM OF SUMS\*\*2 = 0.1548042004914454D 03  
SUM OF SQUARE OF SUMS = 0.3820324148469269D 03  
AMCNG GROUP SUM OF SQUARES = 0.5041406896350342D 00  
WITHIN GROUP SUM OF SQUARES = 0.1509838763429761D 03  
TOTAL SUM OF SQUARES = 0.1514880170326111D 03  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.3541763825767676D 00

F-STATISTICS FOR THE 18 COMPONENT

SUM OF SUMS = 0.7950452893326030D 02  
SUM OF SUMS\*\*2 = 0.1454127656122291D 03  
SUM OF SQUARE OF SUMS = 0.5157161626920556D 03  
AMCNG GROUP SUM OF SQUARES = 0.9431815463208067D 00  
WITHIN GROUP SUM OF SQUARES = 0.1402556039853086D 03  
TOTAL SUM OF SQUARES = 0.1411987855316294D 03  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.7133020797581566D 00

F-STATISTICS FOR THE 19 COMPONENT

SUM OF SUMS = 0.8206200282896035D 02

SUM OF SUMS\*\*2 = 0.1642503769158388D 03

SUM OF SQUARE OF SUMS = 0.5788746875066006D 03

AMCNG GROUP SUM OF SQUARES = 0.1299298669532475D 01

WITHIN GROUP SUM OF SQUARES = 0.1584616300407728D 03

TOTAL SUM OF SQUARES = 0.1597609287103052D 03

DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02

DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04

F-STATISTIC = 0.8697276809711277D 00

F-STATISTICS FOR THE 20 COMPONENT

SUM OF SUMS = 0.9520048302882186D 02

SUM OF SUMS\*\*2 = 0.2094086558414699D 03

SUM OF SQUARE OF SUMS = 0.8713219218815827D 03

AMCNG GROUP SUM OF SQUARES = 0.2671131239535162D 01

WITHIN GROUP SUM OF SQUARES = 0.2006954366226541D 03

TOTAL SUM OF SQUARES = 0.2033665678621893D 03

DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02

DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04

F-STATISTIC = 0.1411744637781583D 01

F-STATISTICS FOR THE 21 COMPONENT

SUM OF SUMS = 0.8994139952150146D 02

SUM OF SUMS\*\*2 = 0.2335862791765991D 03

SUM OF SQUARE OF SUMS = 0.7245755635510632D 03

AMCNG GROUP SUM OF SQUARES = 0.1852785403586404D 01

WITHIN GROUP SUM OF SQUARES = 0.2263405235410884D 03

TOTAL SUM OF SQUARES = 0.2281933089446748D 03

DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02

DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04

F-STATISTIC = 0.8682828488687504D 00



F-STATISTICS FOR THE 22 COMPONENT

SUM OF SUMS = 0.1035648300735959D 03  
SUM OF SUMS\*\*2 = 0.2414810868436007D 03  
SUM OF SQUARE OF SUMS = 0.9427341493765320D 03  
AMCNG GROUP SUM OF SQUARES = 0.2276892141650119D 01  
WITHIN GROUP SUM OF SQUARES = 0.2320537453498354D 03  
TOTAL SUM OF SQUARES = 0.2343306374914855D 03  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.1040764077320930D 01

F-STATISTICS FOR THE 23 COMPONENT

SUM OF SUMS = 0.9031234717154746D 02  
SUM OF SUMS\*\*2 = 0.2047768116906234D 03  
SUM OF SQUARE OF SUMS = 0.7149379755971323D 03  
AMCNG GROUP SUM OF SQUARES = 0.1711833054881912D 01  
WITHIN GROUP SUM OF SQUARES = 0.1976274319346521D 03  
TOTAL SUM OF SQUARES = 0.1993392649895340D 03  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.9187822552243536D 00

F-STATISTICS FOR THE 24 COMPONENT

SUM OF SUMS = 0.9171706807335522D 02  
SUM OF SUMS\*\*2 = 0.1900438745495707D 03  
SUM OF SQUARE OF SUMS = 0.8139562480340389D 03  
AMCNG GROUP SUM OF SQUARES = 0.2531548763025405D 01  
WITHIN GROUP SUM OF SQUARES = 0.1819043120692303D 03  
TOTAL SUM OF SQUARES = 0.1844358608322557D 03  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.1476188171339999D 01

F-STATISTICS FOR THE 25 COMPONENT

SUM OF SUMS = 0.8440126623444366D 02  
SUM OF SUMS\*\*2 = 0.1782254895437147D 03  
SUM OF SQUARE OF SUMS = 0.6892901073945840D 03  
AMCNG GROUP SUM OF SQUARES = 0.2143851912627546D 01  
WITHIN GROUP SUM OF SQUARES = 0.1713325884697689D 03  
TOTAL SUM OF SQUARES = 0.1734764403823964D 03  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.1327251499840134D 01

F-STATISTICS FOR THE 26 COMPONENT

SUM OF SUMS = 0.6923579472492147D 02  
SUM OF SUMS\*\*2 = 0.1481985058366125D 03  
SUM OF SQUARE OF SUMS = 0.4452980334185913D 03  
AMCNG GROUP SUM OF SQUARES = 0.1257250153391605D 01  
WITHIN GROUP SUM OF SQUARES = 0.1437455255024266D 03  
TOTAL SUM OF SQUARES = 0.1450027756558182D 03  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.9277389287476910D 00

F-STATISTICS FOR THE 27 COMPONENT

SUM OF SUMS = 0.6366092822129744D 02  
SUM OF SUMS\*\*2 = 0.1177906819912730D 03  
SUM OF SQUARE OF SUMS = 0.3935159041535181D 03  
AMCNG GROUP SUM OF SQUARES = 0.1233349853537058D 01  
WITHIN GROUP SUM OF SQUARES = 0.1138555229497378D 03  
TOTAL SUM OF SQUARES = 0.1150888728032749D 03  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.1149027974258136D 01

F-STATISTICS FOR THE 28 COMPONENT

SUM OF SUMS = 0.5387592494868525D 02  
SUM OF SUMS\*\*2 = 0.1186133442790129D 03  
SUM OF SQUARE OF SUMS = 0.2792595957915495D 03  
AMCNG GROUP SUM OF SQUARES = 0.8575190985312517D 00  
WITHIN GROUP SUM OF SQUARES = 0.1158207483210974D 03  
TOTAL SUM OF SQUARES = 0.1166782674196286D 03  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.7853366268738306D 00

F-STATISTICS FOR THE 29 COMPONENT

SUM OF SUMS = 0.4536144151906700D 02  
SUM OF SUMS\*\*2 = 0.6564305139058856D 02  
SUM OF SQUARE OF SUMS = 0.1919053830677748D 03  
AMONG GROUP SUM OF SQUARES = 0.5472802462192574D 00  
WITHIN GROUP SUM OF SQUARES = 0.6372399755991081D 02  
TOTAL SUM OF SQUARES = 0.6427127780613007D 02  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.9109723144851782D 00

F-STATISTICS FOR THE 30 COMPONENT

SUM OF SUMS = 0.3732568029970994D 02  
SUM OF SUMS\*\*2 = 0.5705765228634419D 02  
SUM OF SQUARE OF SUMS = 0.1284820507967475D 03  
AMCNG GROUP SUM OF SQUARES = 0.3560162347433721D 00  
WITHIN GROUP SUM OF SQUARES = 0.5577283177837672D 02  
TOTAL SUM OF SQUARES = 0.5612884801312009D 02  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.6770886363437515D 00



F-STATISTICS FOR THE 31 COMPONENT

SUM OF SUMS = 0.2973876286269883D 02  
SUM OF SUMS\*\*2 = 0.4217833610659068D 02  
SUM OF SQUARE OF SUMS = 0.7726205204918280D 02  
AMONG GROUP SUM OF SQUARES = 0.1830245094226048D 00  
WITHIN GROUP SUM OF SQUARES = 0.4140571558609885D 02  
TOTAL SUM OF SQUARES = 0.4158874009552145D 02  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.4688645251806333D 00

F-STATISTICS FOR THE 32 COMPONENT

SUM OF SUMS = 0.2142367689889270D 02  
SUM OF SUMS\*\*2 = 0.1414820425416038D 02  
SUM OF SQUARE OF SUMS = 0.3565723402182890D 02  
AMONG GROUP SUM OF SQUARES = 0.5058971897285643D-01  
WITHIN GROUP SUM OF SQUARES = 0.1379163191394209D 02  
TOTAL SUM OF SQUARES = 0.1384222163291495D 02  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.3890854828465455D 00

F-STATISTICS FOR THE 33 COMPONENT

SUM OF SUMS = 0.1732004830070321D 02  
SUM OF SUMS\*\*2 = 0.1115582097066481D 02  
SUM OF SQUARE OF SUMS = 0.2312899538600082D 02  
AMONG GROUP SUM OF SQUARES = 0.3130057176754676D-01  
WITHIN GROUP SUM OF SQUARES = 0.1092453101680480D 02  
TOTAL SUM OF SQUARES = 0.1095583158857235D 02  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.1485000000000000D 04  
F-STATISTIC = 0.3039120267386334D 00

F-STATISTICS OF MEAN VARIANCE, SKEWNESS AND KURTOSIS

F-STATISTICS FOR THE 1 COMPONENT

SUM OF SUMS = 0.1111617439422250D 05  
SUM OF SUMS\*\*2 = 0.4319425777749392D 07  
SUM OF SQUARE OF SUMS = 0.1861260733960752D 08  
AMONG GROUP SUM OF SQUARES = 0.3143833877400904D 06  
WITHIN GROUP SUM OF SQUARES = 0.3755407373518861D 07  
TOTAL SUM OF SQUARES = 0.4069790761258952D 07  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.4800000000000000D 03  
F-STATISTIC = 0.2870223636519026D 01

F-STATISTICS FOR THE 2 COMPONENT

SUM OF SUMS = -0.3062348822555926D 07  
SUM OF SUMS\*\*2 = 0.7751235739069674D 12  
SUM OF SQUARE OF SUMS = 0.2376364181020636D 13  
AMONG GROUP SUM OF SQUARES = 0.5306562101878761D 11  
WITHIN GROUP SUM OF SQUARES = 0.7031125381184634D 12  
TOTAL SUM OF SQUARES = 0.7561781591372509D 12  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.4800000000000000D 03  
F-STATISTIC = 0.2587626620217669D 01

F-STATISTICS FOR THE 3 COMPONENT

SUM OF SUMS = 0.1087692759875071D 08  
SUM OF SUMS\*\*2 = 0.1368235571001090D 14  
SUM OF SQUARE OF SUMS = 0.3521258942662316D 14  
AMONG GROUP SUM OF SQUARES = 0.8280430048704694D 12  
WITHIN GROUP SUM OF SQUARES = 0.1261530754556778D 14  
TOTAL SUM OF SQUARES = 0.1344335055043825D 14  
DEGREES OF FREEDOM IN NUMERATOR = 0.1400000000000000D 02  
DEGREES OF FREEDOM IN DENOMINATOR = 0.4800000000000000D 03  
F-STATISTIC = 0.2250444214596077D 01

## LIST OF REFERENCES

1. CLARKE, A.C., "Extra-Terrestrial Relays", Wireless World, vol. 51, no. 10, p. 305, Oct. 1945.
2. Pierce, J.R., "Orbital Radio Relays", Jet Propulsion, vol. 25, p. 153, Apr. 1955.
3. "Communication Satellites", Research Report, Encyclopedia Britannica, Inc., Chicago, Ill., 1978.
4. Jaffe, L., "NASA Communications Satellite Developments", Astronautics and Aerospace Engineering, vol. 1, no. 8, p. 48, Sep. 1963.
5. Edelson, B.I., Strauss, R., and Bargellini, P.L., "INTELSAT System Reliability", Acta Astronautica, vol. 2, p. 691, 1975.
6. Gagliardi, R.M., Satellite Communications, Lifetime Learning Publications, Belmont, Ca., 1984.
7. Bhargava, V.K., et al., Digital Communications by Satellite, John Wiley and Sons, Inc., 1981.
8. Feher, K., Digital Communications-Satellite/Earth Station Engineering, Prentice-Hall, Inc., Englewood Cliffs, New Jersey, 1984.
9. Nyquist, "Certain Topics in Telegraph Transmission Theory", Transactions of the AIEE, Apr., 1928.
10. Shannon, et al., "The Philosophy of PCM", Proceedings of the IERE, Nov. 1948.
11. "Digital Signal Modulation Techniques in Satellite Communications", Research Report, Encyclopedia Britannica, Inc., Chicago, Ill., 1984.
12. Lender, A., "The Duobinary Technique for High Speed Data Transmission", IEEE Transactions on Communications and Electronics, vol. 82, p. 274, May, 1963.
13. Kabal, P. and Pasupathy, S., "Partial Response Signalling", IEEE Transactions on Communications, vol. COM-23, no. 9, Sep. 1975.



14. Oshita, S. and Feher, K., "Combined Effect of the Carrier Recovery and Symbol Timing Error on the Pe Performance of QPR and Offset QPR Systems", IEEE Transactions on Communications, vol. COM-30, no. 12, Dec. 1982.
15. Harmuth, H.F., Sequency Theory, Foundations and Applications, vol. 9, Advances in Electronics and Electron Physics, Academic Press, 1977.
16. Hahn, G.J. and Shapiro, S.S., Statistical Models in Engineering, John Wiley and Sons, Inc., 1967.
17. Hickman, E.P. and Hilton, J.G., Probability and Statistical Analysis, Intex Educational Publishers, 1971.

## BIBLIOGRAPHY

Ahmed, N. and Rao, K.R., Orthogonal Transforms for Digital Signal Processing, Springer-Verlag, New York, 1975.

Boutin, N. and Morisette, S., "Discrete Finite Duration Partial Response Signaling Pulse with Minimum Out of Band Power", IEEE Globecom, vol. 2, 1983.

Brigham, E.O., The Fast Fourier Transform, Prentice-Hall, Inc., Englewood Cliffs, New Jersey, 1977.

Feher, K., Digital Communications Microwave Applications, Prentice-Hall, Inc., Englewood Cliffs, New Jersey, 1981.

Herman, R., et al, "A Modular Program for Simulation of Digital Systems and Its Application to Satellite System Optimization", Proceedings on the Joint Conference on Digital Processing of Signals in Communications, IERE, Apr. 1972.

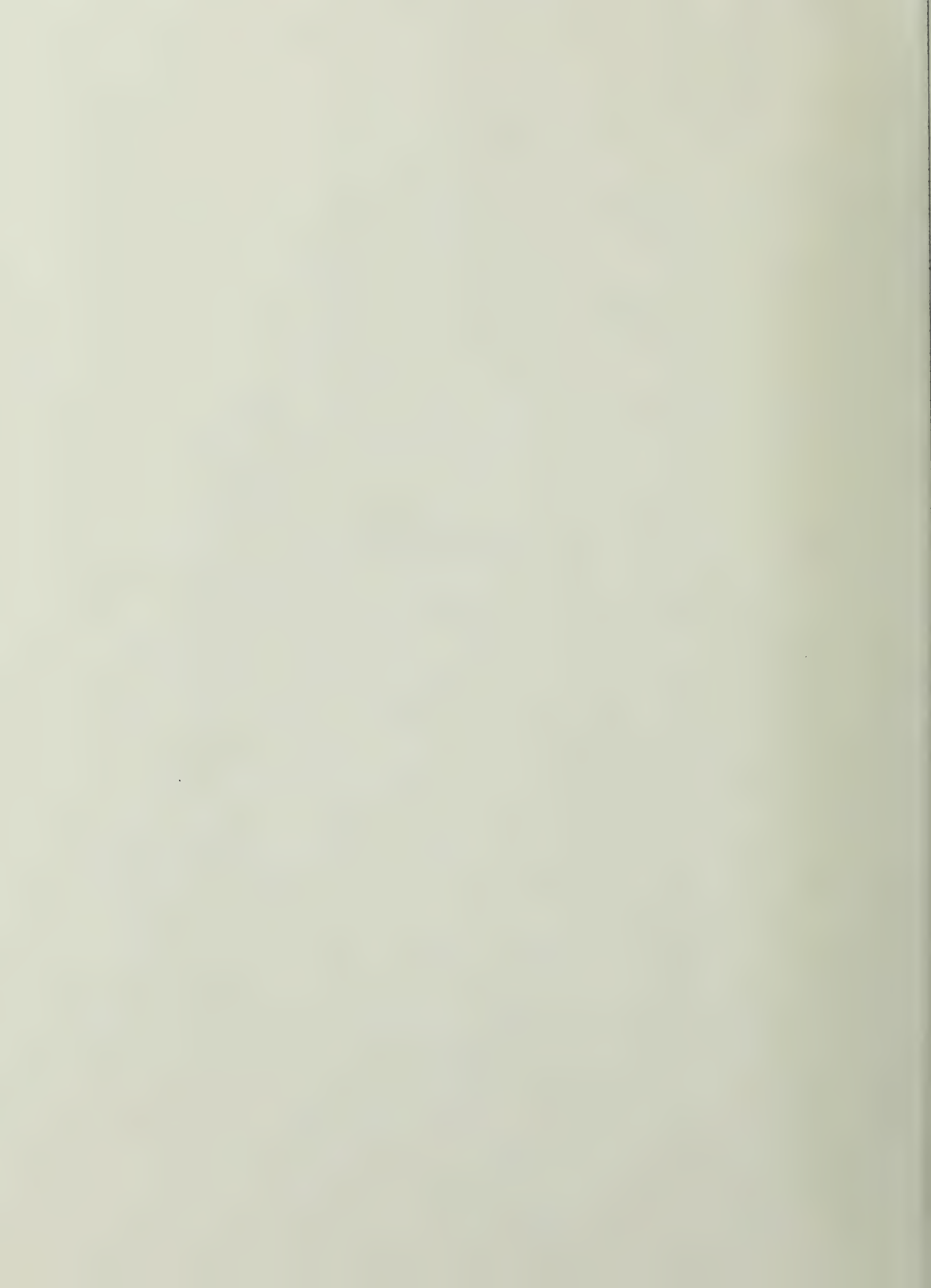
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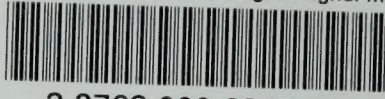
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